

## **USER GUIDE**

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For more information, please contact Ridley Engineering directly for assistance. We are here to help you.

E-Mail: [info@ridleyengineering.com](mailto:info@ridleyengineering.com)

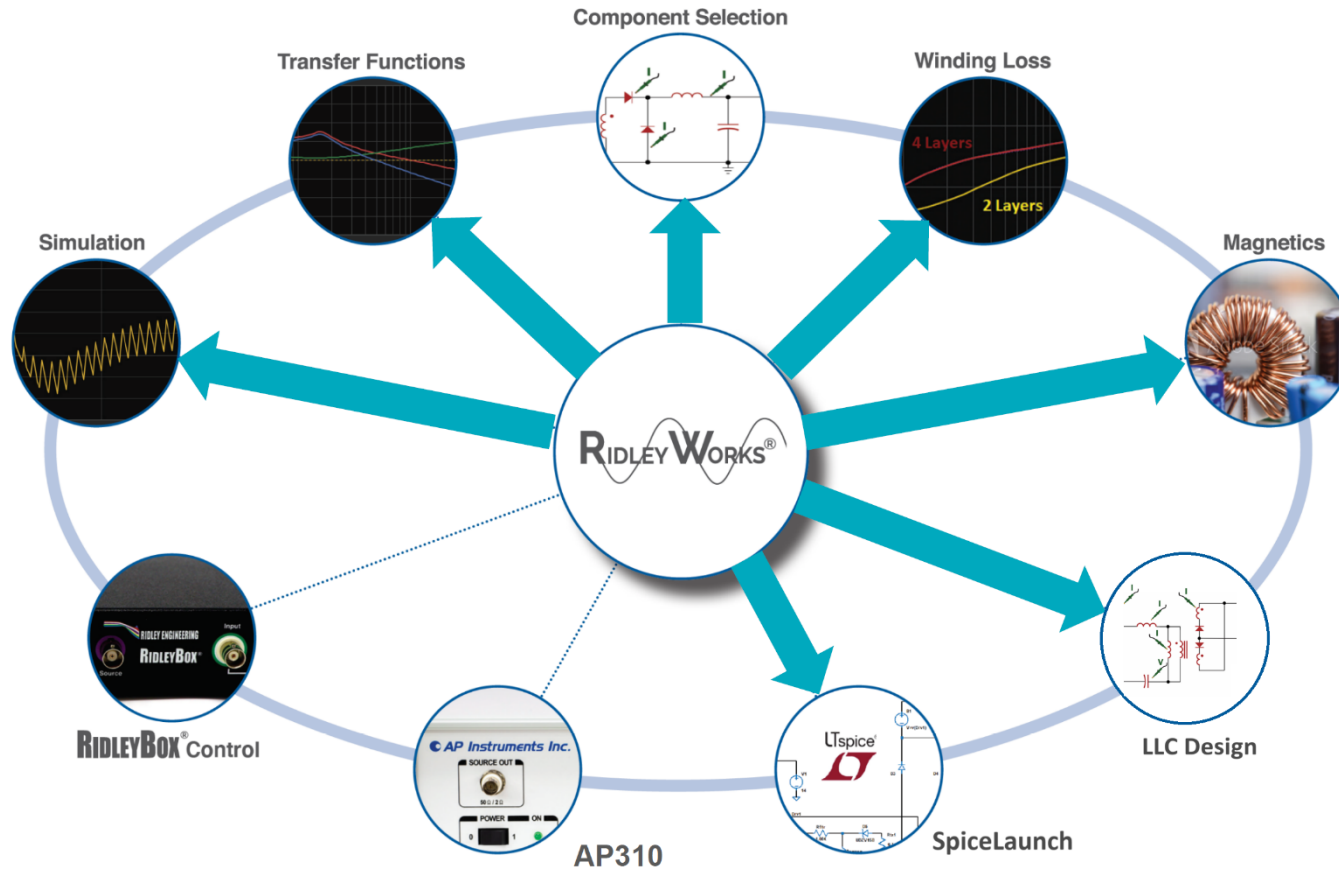
<http://www.ridleyengineering.com>

# Table of Contents

<b>1 Introduction to RidleyWorks®</b>	<b>4</b>	<b>3 Control Analysis Transfer Functions</b>	<b>27</b>
1.1 Computer System Requirements (PC Based)	7	3.1 Turning On Control Options	27
1.2 Computer System Requirements (MAC Based)	7	3.2 Voltage-Mode and Current-Mode Control	28
1.3 RidleyWorks® Software Download	7	3.3 Interactive Compensation Adjustment	30
1.4 RidleyWorks® Software Installation	8	3.4 Voltage Feedforward Control	32
1.5 RidleyWorks® Software Upgrade	10	<b>4 Magnetics Design</b>	<b>33</b>
1.6 Loading Workbooks	10	4.1 Turning On Magnetics Options	33
1.7 Saving Workbooks	11	4.2 Inductor Design Page	34
1.8 Saving Data Only and Transferring to New Version	11	4.3 Choosing the Inductor Core Size	35
1.9 Which Version Do I have?	12	4.4 Choosing the Inductor Core Material	36
1.10 Help Notes	12	4.5 Inductor Winding Design	37
1.11 Software Support	13	4.6 Transformer Design	38
1.12 Facebook Support Groups	13	<b>5 Exporting to LTspice® with SPICELaunch®</b>	<b>39</b>
<b>2 Design and Waveform Simulation</b>	<b>14</b>	5.1 Preparing an LTspice® file	39
2.1 Entering Specifications and Topology	15	5.2 Simulation Settings in LTspice®	40
2.2 Running Waveform Simulations	18	5.3 Simulating Transients in LTspice®	42
2.3 Startup Simulation	20	5.4 Generating Swept Bode Plots in LTspice®	43
2.4 Step-Load Transient Simulation	22	5.5 Generating Small-Signal Bode Plots in LTspice®	48
2.5 Adding Input Voltage Modulation	23		
2.6 AC Input Simulation	24		
2.7 Exploring Further Simulation Options	26		

<b>6 Advanced Proximity Loss and Core Loss</b>	<b>55</b>		
6.1 Estimating Proximity Losses in RidleyWorks®	56		
6.2 Advanced Proximity Losses with LTspice® Modeling	60		
6.3 Running LTspice® for Proximity Loss	61		
6.4 Advanced Core Losses with LTspice® Modeling	62		
<b>7 Using RidleyWorks® with the AP310 Analyzer</b>	<b>65</b>		
7.1 Connecting RidleyWorks® to the AP Analyzer	65		
7.2 Measuring Transfer Functions	68		
<b>8 Using RidleyWorks® with the RidleyBox®</b>	<b>69</b>		
8.1 Connecting RidleyWorks® to the RidleyBox®	69		
8.2 RidleyBox® Interface	70		
<b>9 Designing LLC Converters with RidleyWorks®</b>	<b>71</b>		
9.1 Why LLC?	71		
9.2 RidleyWorks® LLC Design Overview	72		
9.3 Entering Specifications	73		
9.4 Frequency Selection	74		
9.5 LLC Topology Choices	75		
9.6 LLC Circuit Optimization	76		
9.7 LLC Circuit Simulation	77		
9.8 High-Line Operation	78		
9.9 Exporting Waveforms from LTspice®	79		
9.10 Importing LLC Waveforms	80		
9.11 Viewing LLC Waveforms in RidleyWorks®	81		
9.12 LLC Transformer Design	82		
9.13 LLC Transformer Winding Design	83		
		9.14 LLC Inductor Design	84
		9.15 LLC Inductor Winding Design	85
		9.16 LLC Gain Curve Schematic	86
		9.17 LLC Gain Curves	87
		<b>Appendix A Error Messages</b>	<b>100</b>
		<b>Appendix B License Agreement</b>	<b>110</b>

# 1 Introduction to RidleyWorks®



RidleyWorks® is the only switching power supply design program which provides component design, large-signal simulation, feedback control design, and small signal analysis in one easy-to-use package. Features of **RidleyWorks®**, Release 14 include:

- **Power stage designer**
- **Magnetics designer** (included in full version, not in Control and Simulation version.) Library of cores.
- **12 most popular topologies** used in the industry. All topologies in one program allow you to compare quickly and easily
- Industry's fastest cycle-by-cycle simulation shows true large-signal performance
- **SPICELaunch™** instantly creates **LTspice®** files to continue design and more detailed simulation with all components selected.
- **Unique LTspice® models** for proximity loss and core loss.
- **Control loop designer** suitable for constant-frequency PWM
- **Current-mode control** using the latest and most accurate modeling techniques
- **Voltage-mode control** using the results of the PWM switch model
- **CCM and DCM** converter simulation and analysis
- **Small-signal analysis** of control system, including loop gain
- **Digital compensation** delay calculation
- Two-stage **input filter design** and analysis
- **Output impedance** analysis
- Audiosusceptibility or **PSRR analysis**
- **Second-stage LC** output filter analysis and design
- **Snubber** design and analysis
- Switching loss analysis for power FETs and IGBTs

There are three levels of application for this software. These levels are commensurate with experience to help you achieve the greatest value from RidleyWorks® without the burden of struggling with the **5800+** design equations involved in arriving at the solution.

## Level 1: Basic Design

At this level, you can simply input the power requirements for your circuit. When you click “OK”, the entire converter, including control loop, is designed for you. The components are chosen for you based on your data, and these are working values which you can put into hardware with confidence. You can then examine the design with the simulations of RidleyWorks® or convert designs to LTspice® files with our [SPICELaunch®](#) feature.

## Level 2: Intermediate Design

Level two designers have design experience and want to input their knowledge as the design progresses. They want to understand and further optimize their design. At this level, you will look carefully at the selected values from RidleyWorks® and change some of them to improve the design.

## Level 3: Advanced Design

Level three designers not only know how to design an effective power supply— they know how to get maximum performance out of their power system for their very specific applications. They have a library of preferred parts that work well for their applications and use these to fully optimize the design.

The magnetics design section works similarly in levels. You can view the transformer or inductor on the main design schematic and

1. Choose from simplified parameters and refer to a magnetics design house for completion.
2. Choose your own cores, wire, and other components and give the completed design to a manufacturer for production; or
3. Design the magnetics from the ground up using your own parameters and knowledge.

Regardless of your design experience, RidleyWorks® allows you to check and cross-check all parameters and results. You can stop at any time during the process and view resulting waveforms and control parameters.

We hope you enjoy the benefits of this interactive design tool. To learn more about the powerful capabilities of the program, we highly recommend attending one of our hands-on [Five-Day Power Supply Design Courses](#).

## 1.1 Computer System Requirements (PC-Based)

To run RidleyWorks® effectively, any modern computer with at least **Excel 2019 or later** installed will work. The RidleyWorks® code is extremely efficient, and it does not need a high-powered computer. It is also very compact, taking up only 16 MB of disk space.

Make sure you select the **64-bit version** of Excel when installing.

RidleyWorks® is recommended for use with Windows 10 or later.

## 1.2 Computer System Requirements (MAC-Based)

Microsoft Office is available to run on the Apple, but Excel for MAC does not work as it should. You will be able to load RidleyWorks®, but the formatting will not show everything properly. To use RidleyWorks® on the MAC, you will need to install a Windows partition and the normal PC-based version of Excel (**64-bit**).

## 1.3 RidleyWorks® Software Download

The latest version of RidleyWorks® is available for [download](http://ridleyengineering.com/software-ridley/download-RidleyWorks.html) at <http://ridleyengineering.com/software-ridley/download-RidleyWorks.html>

If you are installing for the first time, you should download the full install version, and if you are upgrading, download the update version.

To download the software, you will need a software product key which looks like

**RIDLEYWORKS-XXXX-XXXX-XX**



## 1.4 RidleyWorks® Software Installation

### Overview

Install **RidleyWorks®** in **C:\RidleyWorks** directory. All files will be placed here, allowing multiple connections between files.

Install **LTspice®** in the default directory during install. The subroutines and symbols needed should be found in the appropriate folder.

### Detailed Procedure

1. Earlier version of RidleyWorks® should first be removed using the Windows Uninstall utility.
2. You should have downloaded a zip file RIDLEYWORKSINSTALL on your computer. Inside this zip file, you will see the installation file:

#### **RIDLEYWORKSINSTALL.EXE**

There will also be RidleyWorks16.xx file in the zip file. You may be able to just copy this file into your RidleyWorks folder instead of doing a full upgrade. Instructions will be on the download page.

3. Double click on the RIDLEYWORKSINSTALL application to install in your Documents folder.
- 3.2** In the target **C:\RidleyWorks** directory, you should see the following files when the installation program is done:

**RidleyWorks16xx.xlsm**

Transfer.xlsm

RidleyWorks\_to\_LTspice®

There is also a permission code library directory located at C:\ProgramData\AFY\RIDLEYWORKS\cm which contains the license files:

RIDLEYWORKS64.dll  
RIDLEYWORKS64.dll.cm  
RIDLEYWORKS64 dll.cm.ini  
SIMULATIONLOAD.dll  
CMInstall.exe

The blue .xslm filename is the main RidleyWorks® program which will run under Excel. The six files shown in green contain the permission code to work with your specific password for the program, provided to you when you purchased RidleyWorks®. If you receive an error message that Excel cannot find the .dll files, please refer to section 101 of this manual.

4. To use the SPICELaunch® feature of RidleyWorks®, you will need to install LTspice®. You can download that from [Linear Technology's site](#). Please install LTspice® xvii in the default directory.) Additional LTspice® files, symbols, and subcircuits will be placed in the appropriate folders on your computer when you install RidleyWorks® in step 3 above.
5. Double click on the blue file name to launch RidleyWorks®. When the program begins to load, you may get the message that it contains Macros. Click **Enable Macros**. The first time that you load RidleyWorks®, there will be a registration process. You can make copies of this file as you wish on your computer. It is often useful to keep the RidleyWorks file on the desktop or in a specific project file.



Select the file to be configured as a standalone program and enter the password which you should have received upon purchase. After this, you will be asked to enter registration information, including address and contact information.

(If you receive an error about RidleyWorks.dll files not being found, please refer to page 78 of this manual.)

## 1.5 RidleyWorks® Software Upgrades

If you have already installed RidleyWorks® Release 15 or later, please uninstall it before installing the latest version on your computer.

## 1.6 Loading Workbooks



After registration, Excel will then proceed to open the specified workbook, and a page like that shown in Fig. 1.1 will appear. Once you see this the words “License Verified” in the bottom left corner, your product is fully registered.

To make sure sizing is correct for your display, click on [Screen Size](#), and select the option that fit your system. The pages will fit optimally on the screen when you select the toolbars option. You can also use the zoom slide bar at the bottom right to adjust individual worksheets.

If you have a RidleyBox or an AP310, you can directly link to them by clicking on the picture.

**Fig. 1.1: RidleyWorks® Guide Page**

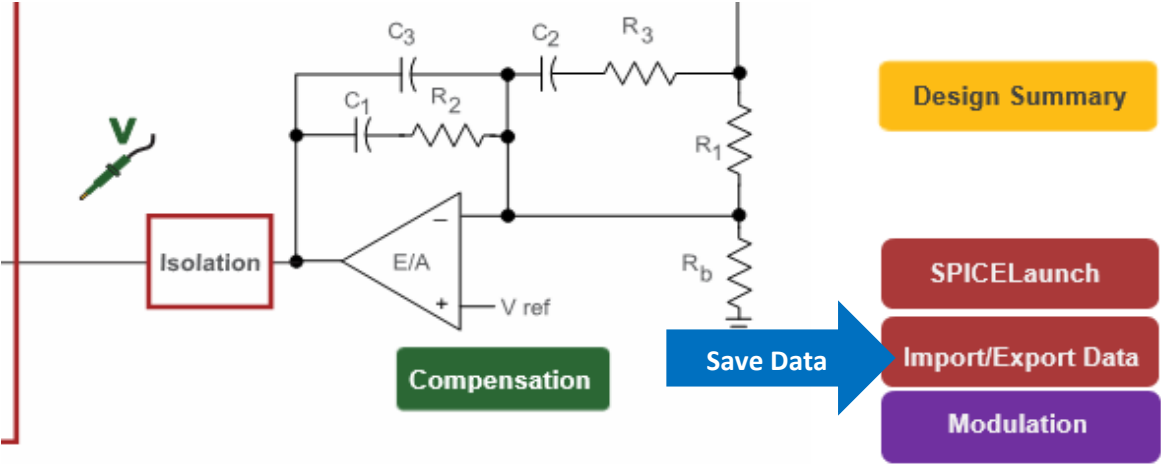
# 1.7 Saving Workbooks

Your design in RidleyWorks® can be saved just like any Excel workbook. It is recommended that you save a project under a different name from the original file. To do this, click on File Save As, and specify a folder and a name for your project.

Every time you do this, you will be saving a complete copy of RidleyWorks® together with the simulator, designer, and all the features of the program. It is a live copy of your design, constantly analyzing and simulating any changes you make, not just the data.

# 1.8 Saving Data Only and Transferring Data to New Version

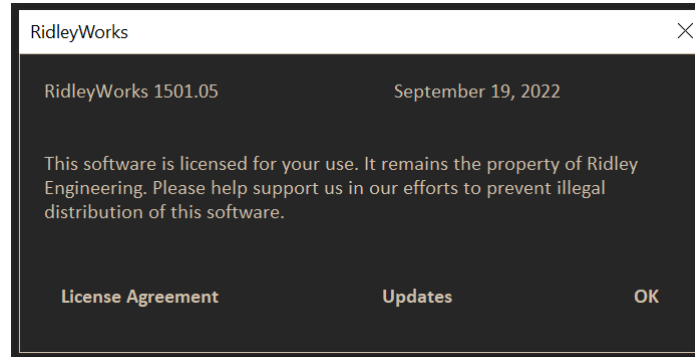
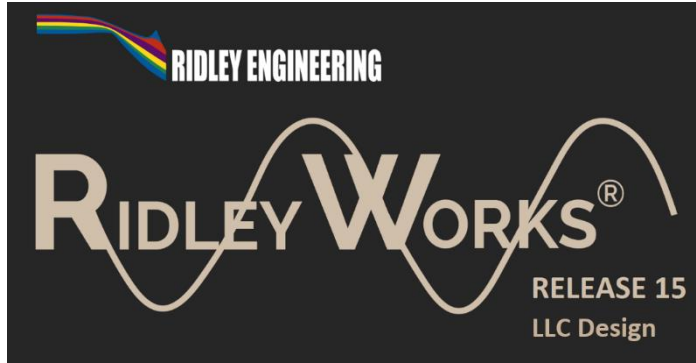
It is possible to save just the data from your design in a much smaller file. This feature has been added to allow you to easily transition your design to a new version of the program without having to re-enter all your design choices. You can also send a design to your colleagues in a very small file.



To take advantage of this feature, it is necessary to have two files open – [RidleyWorks14.xlsm](#) and [TRANSFER.xlsm](#). (the Transfer file should automatically open.) Once you have opened these files, go to the Power Stage page of RidleyWorks® and click on the Export/Import Data button. When you then click the [Export](#) button, your data that you have entered will be placed in the TRANSFER.xlsm file. Clicking on [Import](#) will load data from the TRANSFER.xlsm file into your RidleyWorks® worksheet.

You can then open the latest version of RidleyWorks® and follow the process above to reload your data into the new version.

## 1.9 Which Version Do I Have?



**Fig. 1.2: RidleyWorks® Version Number and Recent Updates.**

To find out which version you have installed, click on the RidleyWorks® name on the opening screen. You can also see the recent changes and upgrades by clicking on the **Updates** button.

## 1.10 Help Notes

RidleyWorks® has extensive on-line help inside the program. Click on any of the help buttons to access notes about a particular feature of RidleyWorks®. A wealth of design information is contained in the program, together with power supply design tips.

## 1.11 Software Support

If you are having problems installing or running RidleyWorks®, please call or e-mail for help. If no one is immediately available, please leave a message, including your name and phone number, and your call will be returned as soon as possible. You can also choose to attach your RidleyWorks® file to an email for assistance.

Phone: (US) 805 504 2212

E-Mail: [info@ridleyengineering.com](mailto:info@ridleyengineering.com)

(UK) +44 (0)1509 276 245

## 1.12 Facebook Support Group

If you are a part of Facebook, there is a group dedicated to power supply design with over 6000 industry members. Whenever a new version of the software is available, a posting will be placed on this site, immediately notifying you. You can also ask questions about the software at this site. The group is called

**[POWER SUPPLY DESIGN CENTER](#)**

## 2 Design and Simulation

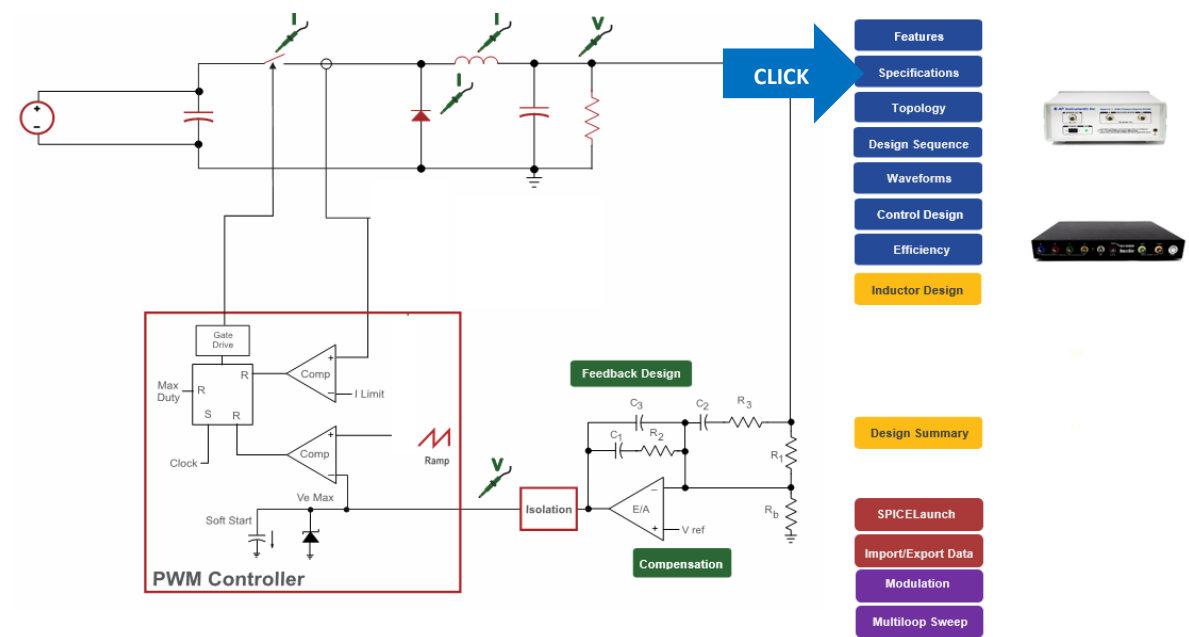
RidleyWorks® has the fastest closed-loop power supply simulator available. It accurately simulates 800 cycles of closed-loop operation of a converter with instantaneous results. Change parameter values, control design, or power stage elements and see the results on true large-signal waveforms instantly. To begin your first design and simulation, click on [Begin Design](#), and you will see be taken to the schematic page to select which features you want to use.



**Fig. 2.1:** RidleyWorks® Guide Page with Start Menu

# 2.1 Entering Specifications and Topology

You will start working on the schematics page as shown below. This will let you see waveforms and small-signal measurements while interactively changing power parts and control parts. An automatic expert-system design routine will help you with all component choices. If you are a new designer, you may choose to let RidleyWorks® choose every parameter for you and you can expect a good working design. On the schematic page, you will see a power topology (a buck converter in the figure below), a controller with all the settings needed to implement most constant-frequency options, and a feedback compensator. The compensator and controller may all be in one chip and you have full control of all the parameters.



Schematic components in red are clickable, and you can edit their values and see components stresses. This will be demonstrated in more detail later. First, though, you will want to enter your specification for your power system. To enter your power specifications, click on the top blue button as shown.

Fig. 2.2: RidleyWorks® Schematic Page



Input and Output Specifications

RidleyWorksBuck 1A Clear Design

**INPUT VOLTAGE RANGE**

☐ 120 VAC
 ☐ 240 VAC
 ☐ 120-240 VAC
 ☒ DC Input

Low Line Voltage   
 Nominal Input Voltage   
 High Line Voltage

**OUTPUT VOLTAGE AND CURRENT**

	Main	Aux 1	Aux 2	Aux 3	Aux 4
Output Voltage	<input type="text" value="12"/>	<input type="text" value="1"/>	<input type="text"/>	<input type="text"/>	<input type="text"/>
Output Current	<input type="text" value="1"/>	<input type="text"/>	<input type="text"/>	<input type="text"/>	<input type="text"/>

Clear Aux

**POWER SUPPLY COOLING**

☐ High Airflow
 ☐ Moderate
 ☐ Low Airflow
 ☒ None

**FEEDBACK ISOLATION**

☒ Non-Isolated
 ☐ TL431/Opto E/A
 ☐ TL431/Opto Bias E/A  
☐ TL431/Opto I FB
 ☐ TL431/Opto Bias I FB

Help OK

The first area of the Specifications form lets you choose either DC or AC input. You can select the pre-programmed ranges of ac, or enter specific voltages that you want to use. One of the unique advantages of RidleyWorks is that there is no limit on the voltage ranges. The design process is not restricted to any specific controller that may limit the operating ranges. Control chip choice, and semiconductor choice, will come later.

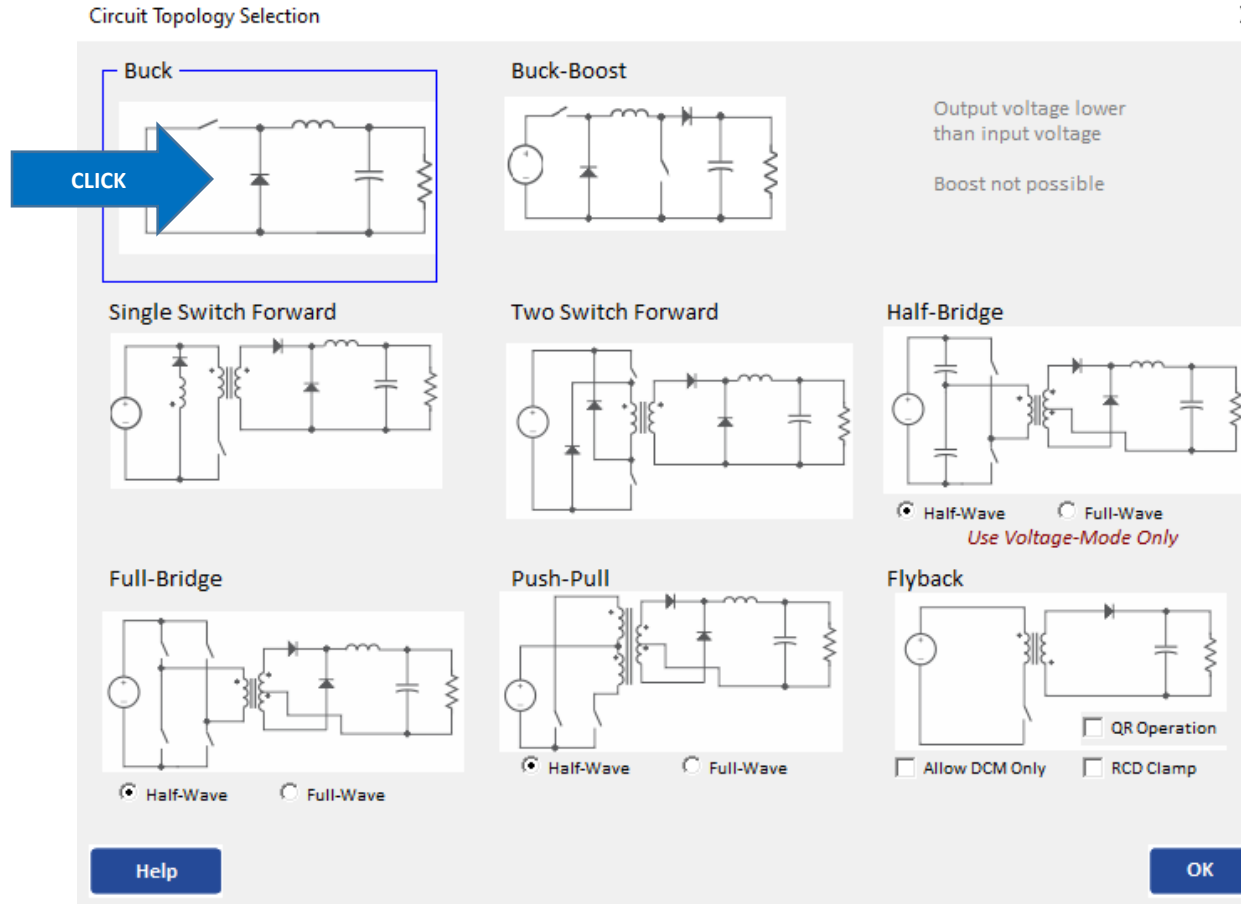
If you plan on using a flyback converter, you can specify up to 5 outputs. Other converters are restricted to single outputs in the present release, although this does not limit the experienced designer.

Airflow and temperature ranges will impact magnetics design. The higher the temperature, the larger the magnetics size estimates will become.

The feedback can be configured as either nonisolated, or with a TL431 optocoupler.

Once you click OK on this form you will see a summary of your inputs to check, then you will be presented with a set of choices for the topology of the power stage.

**Fig. 2.3: RidleyWorks® Specifications Entry Form**



× Figure 2.4 shows the set of topologies available for the input and output voltage specifications. Notice that the boost topology is not possible since the output voltage is lower than part of the input range of the converter.

Once you have selected a new topology, RidleyWorks® will proceed to do an amazing amount of work. First, all the power components will be selected. A controller will then be optimized around the power stage to give good transient performance and stability. Loop gains will be plotted and analyzed. Finally, 400 cycles of large-signal operation will be simulated. All of this is done instantly. This will total change your design and simulation experience. Your design procedure will become much more interactive, trying far more options and repetitively simulating as if you had a live breadboard.

Before digging deeper into design values, it is very valuable to look at circuit waveforms to see how the converter is operating. At this point, you have had no input to design values and RidleyWorks® will provide you with fully automated results for a realistic converter. You can verify this by looking at the waveforms of the circuit.

Fig. 2.4: RidleyWorks® Topology Choices

## 2.2 Running Waveform Simulations

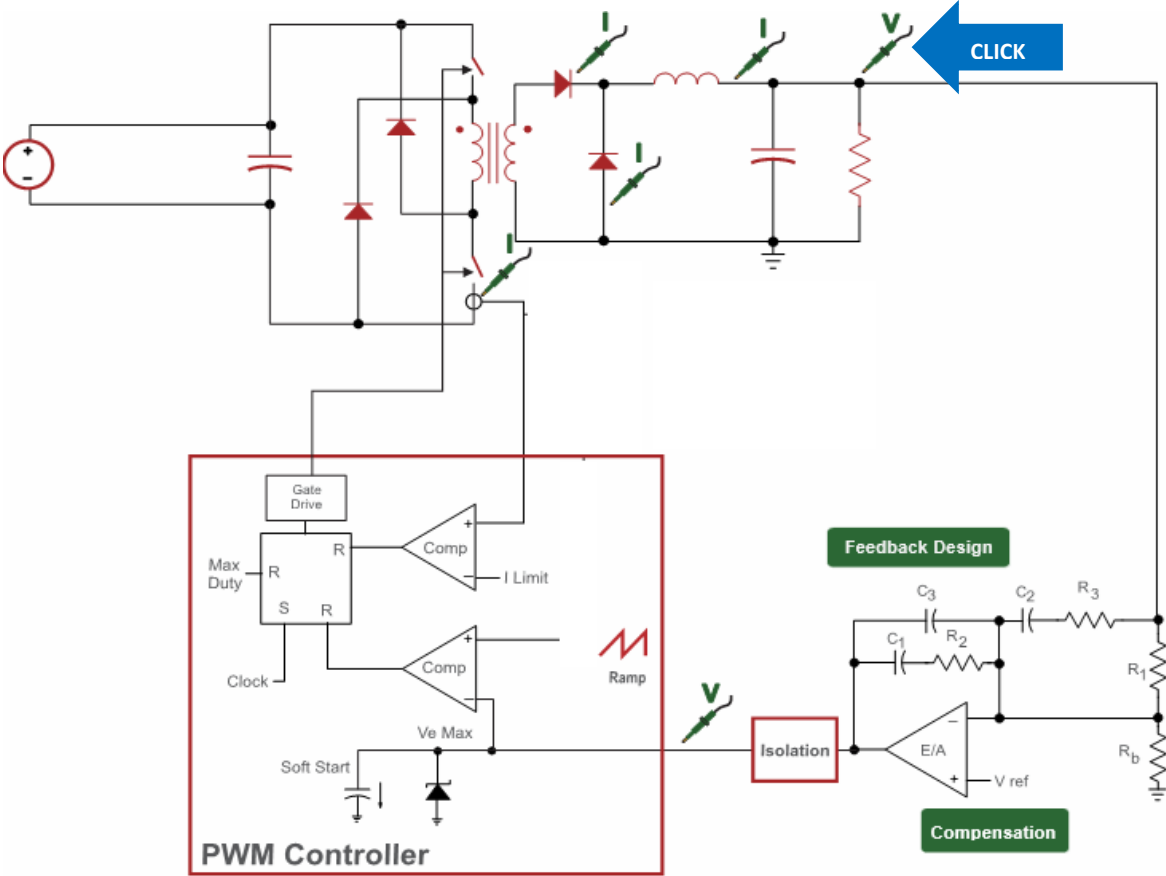


Figure 2.5 shows the selection of voltages and currents that can be probed with RidleyWorks®. The simulation algorithm inside RidleyWorks® is so fast that you will routinely run hundreds of simulations during a design session, never having to wait for the results. Every time a component is changed, another 400-cycle simulation is run to confirm that the converter is operating properly.

To begin doing this click on the blue button labeled Waveforms, then click on the scope probe at the output of the converter.

Waveforms

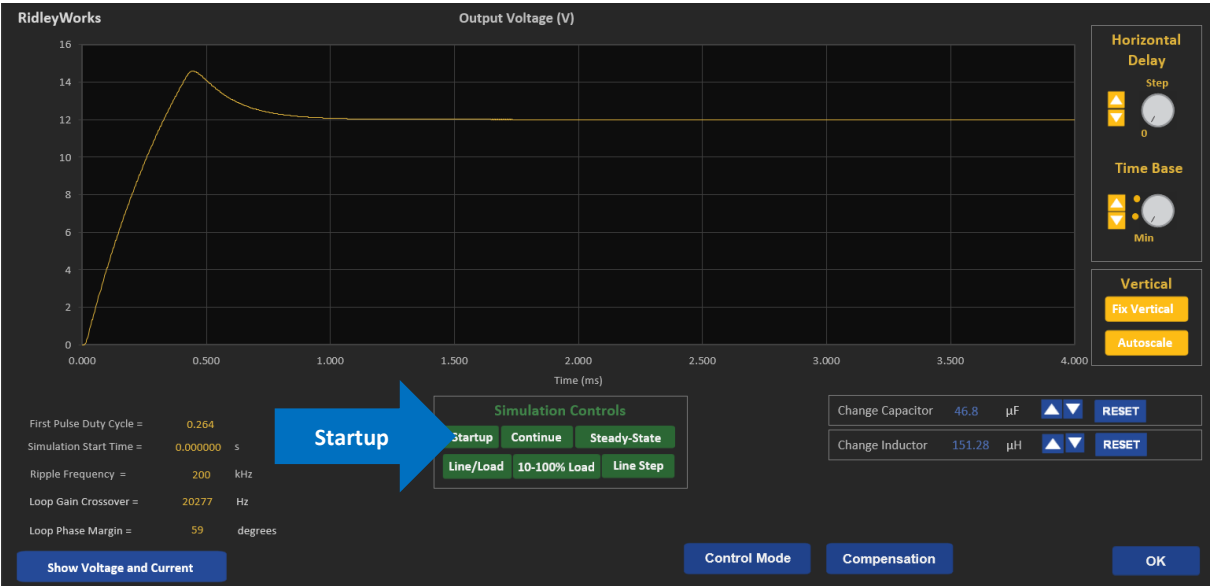
Fig. 2.5: RidleyWorks® Waveform Choices



**Fig. 2.6: RidleyWorks® Output Voltage Waveform**

RidleyWorks® automatically simulates 800 cycles at a time. To zoom in on this waveform, treat the controls of RidleyWorks® just like a scope. Turn the **Time Base** dial clockwise to zoom in by clicking on the up arrow next to the dial. Each time you click on the up arrow, the number of complete waveforms will be cut in half. You can quickly turn the dial back to the original setting by clicking on the Min setting.

## 2.3 Startup Simulation



**Fig. 2.8: RidleyWorks® Start Up Simulation Voltage Waveform**

Startup of your converter can be simulated by clicking the green button as shown above. This will start with zero output voltage, and zero inductor current, and simulate the converter into regulation as shown. It takes over 200 cycles to get the output voltage into regulation in the example shown. It can be instructive to click on button to add the current waveform of the converter to show what effect the current limiting has on the performance of the startup.

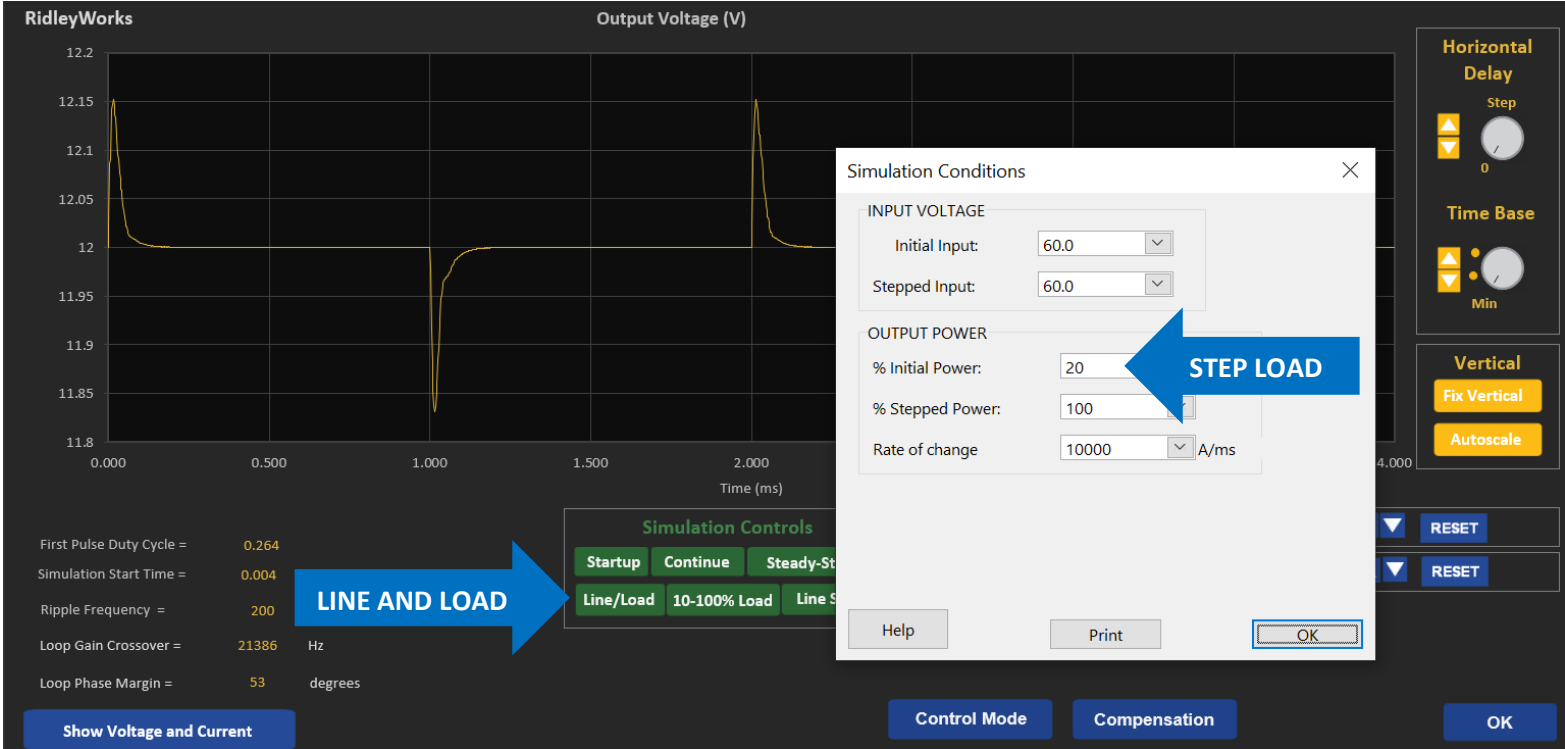


**Fig. 2.9: RidleyWorks® Start Up Simulation Voltage and Current Waveforms**

In the red waveform of Figure 2.9, you can see that the converter enters current limiting during startup.

Determining the proper values of parameters such as current limit is crucial at the very early stages of design. The peak current will impact the design of magnetics and sense resistors, and you don't want to discover this kind of effect later in the design cycle. RidleyWorks® is ideal for this kind of work since it automatically designs all compensation parameters at the very beginning of a design, and this is essential for proper simulation.

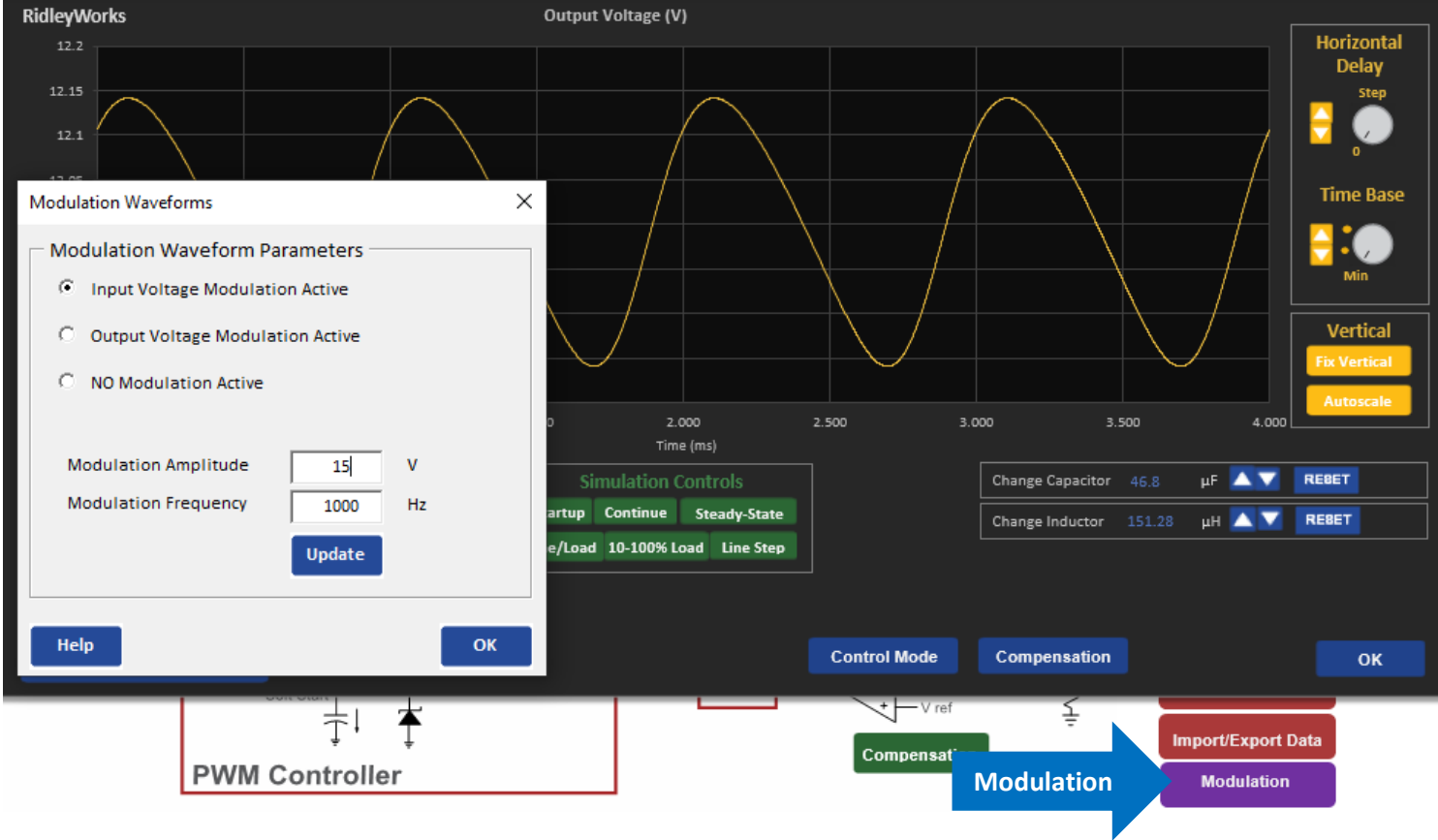
## 2.4 Step-Load Transient Simulation



**Fig. 2.11: RidleyWorks® Step-Load Transient Simulation**

The line and load conditions can be set as shown. Two values of load are shown. The first 200 cycles are run with 100% load, the second 200 cycles with 20% load. Simulation of both the voltage and current are shown, with the inductor current entering DCM on the second transient when the load falls to 20%.

# 2.5 Adding Input Voltage Modulation



RidleyWorks® has the capability to add sinusoidal modulation on top of the input dc voltage. It can also be used to add modulation to the reference voltage to drive the output.

Fig. 2.12: RidleyWorks® Input Line Modulation



## 2.6 AC Input Simulation

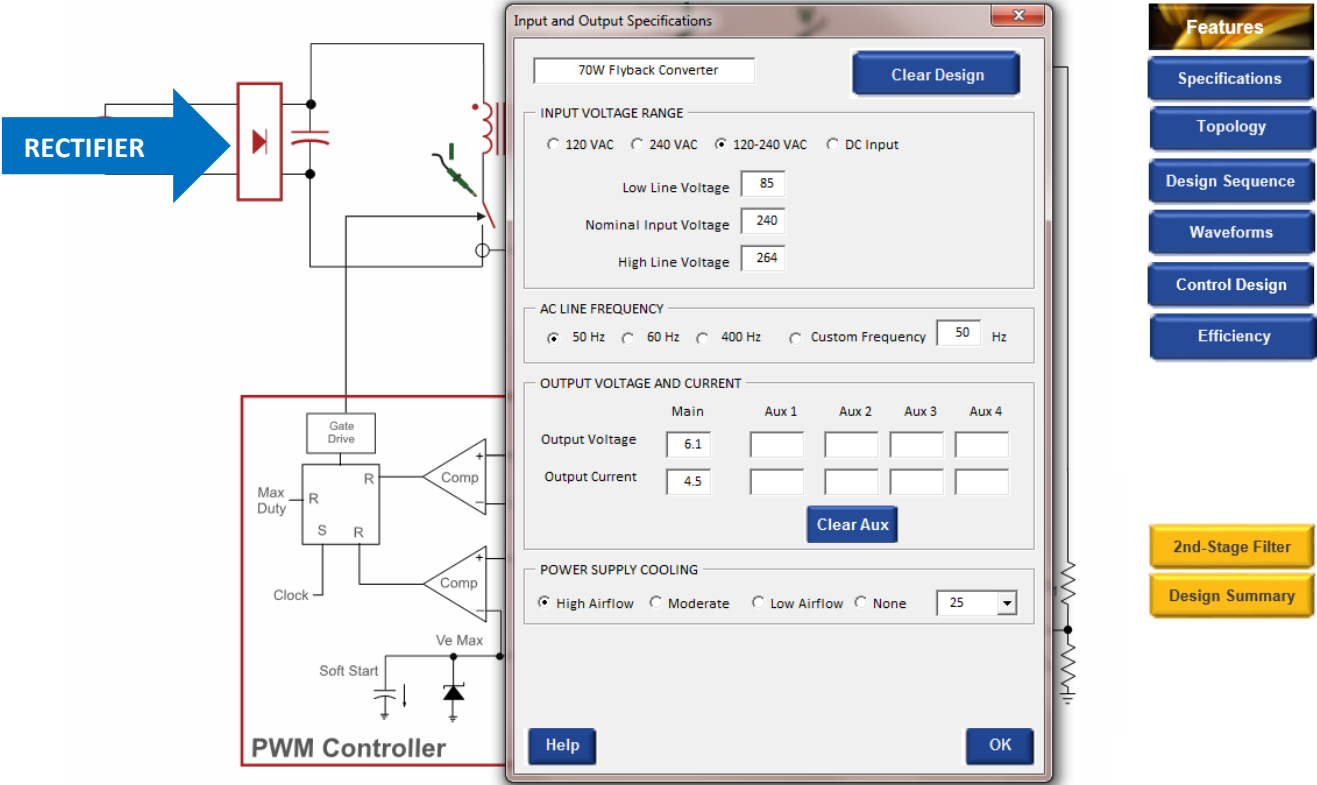


Fig. 2.13: Choosing AC Input on the Specifications Form (not available in Demo Version)

With the full version of RidleyWorks® you can select an AC input and perform a simulation of the input rectifier. Three choices of range are available to you. When you check one of these buttons, as shown below, the expected low, nominal, and high-line voltages will be filled in for you. You can change these values if you wish. Once you have checked an AC input, the schematic will change on the input, showing you the input rectifier block.

After clicking OK, click on the rectifier at the input to see the simulated input waveforms.



The simulation of input line current and input capacitor voltage are shown in the figure above. This helps you size the input capacitor appropriately for the desired amount of ripple at the input of your converter.

**Fig. 2.14: AC Input Waveforms for a 27 W Flyback Converter**

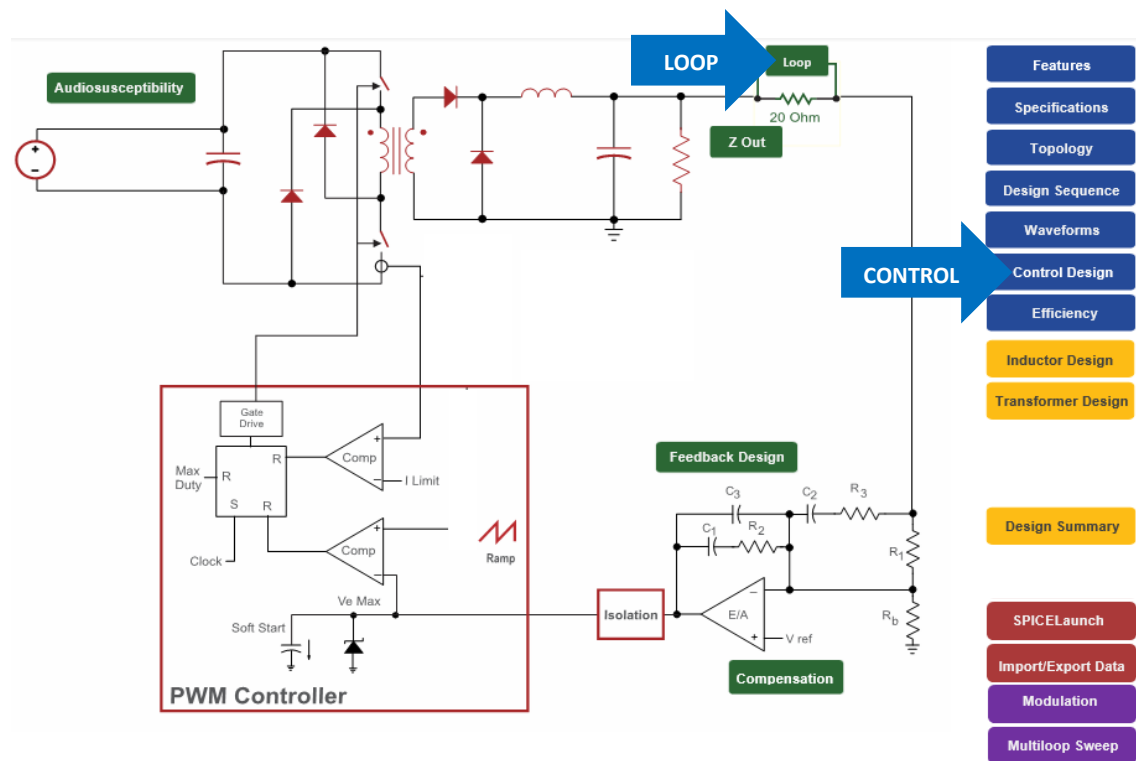
## 2.7 Exploring Further Simulation Options

Many more simulation options are possible within RidleyWorks® and you are encouraged to try them. For example, with the transient load simulation in Fig. 2.11, you can interactively see the change in transient performance with different value inductors and capacitors. You can move these values up and down while watching the change in the waveforms. This is a truly unique feature of RidleyWorks® that will change the way that you do your designs and greatly speed up the process.

You are encouraged to try all these options in RidleyWorks®. Let us know what you see and tell us if there are additional features that you would like to see included in the program in the future. Once you have a license for the software, all upgrades are free if your license is current.

Please contact us to let us know what you would like to see included in the program in the future. We are constantly upgrading the software to respond to user's needs.

# 3 Control Analysis Transfer Functions



RidleyWorks® has unsurpassed techniques for designing and analyzing the control system for a switching power supply. The analysis techniques and design routines have evolved over a period of 35 years to give you all the control information that you need to optimize the performance of your power supply.

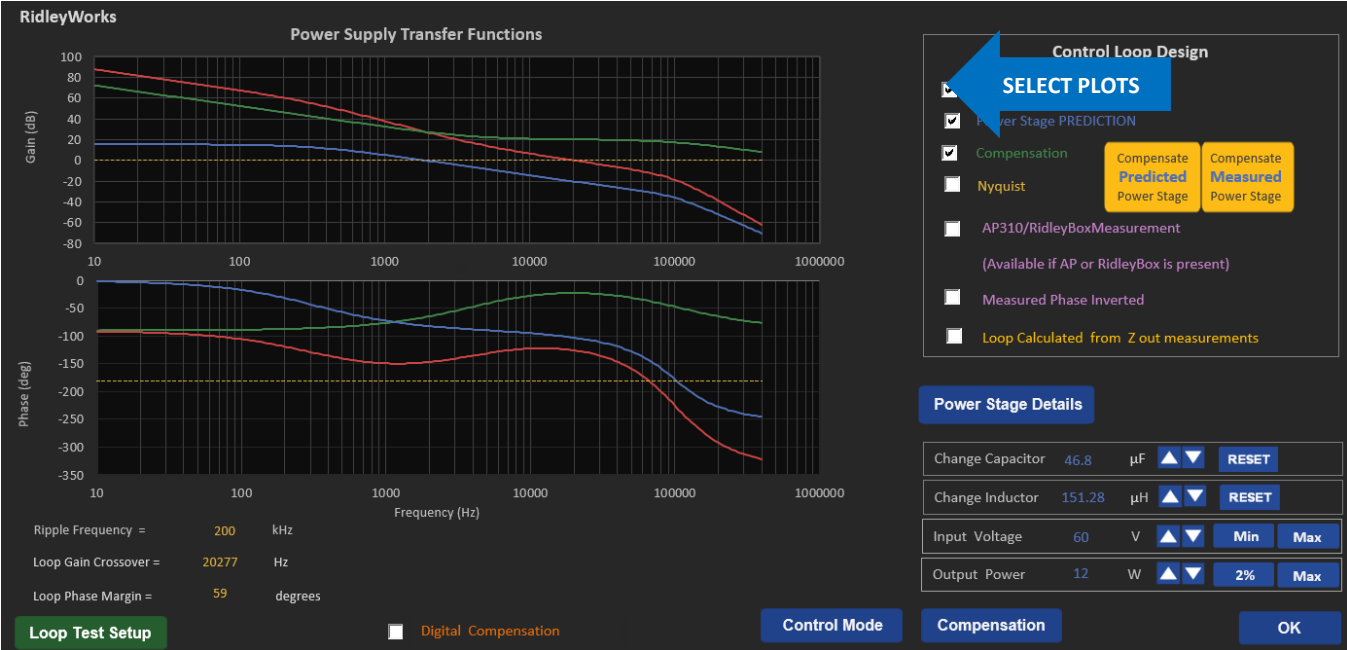
## 3.1 Turning On Control Options

When you click on the **Control Design** blue button you will see the green options buttons below appear on the power stage schematic. (If you don't see the **Z Out** and the **Audiosusceptibility** buttons, make sure these features are checked when you click on the Features button.)

Once you see the green control buttons, click on the **Loop** button to see the main control transfer functions.

Fig. 3.1: Control Design Options in RidleyWorks®

## 3.2 Current-Mode and Voltage-Mode Control



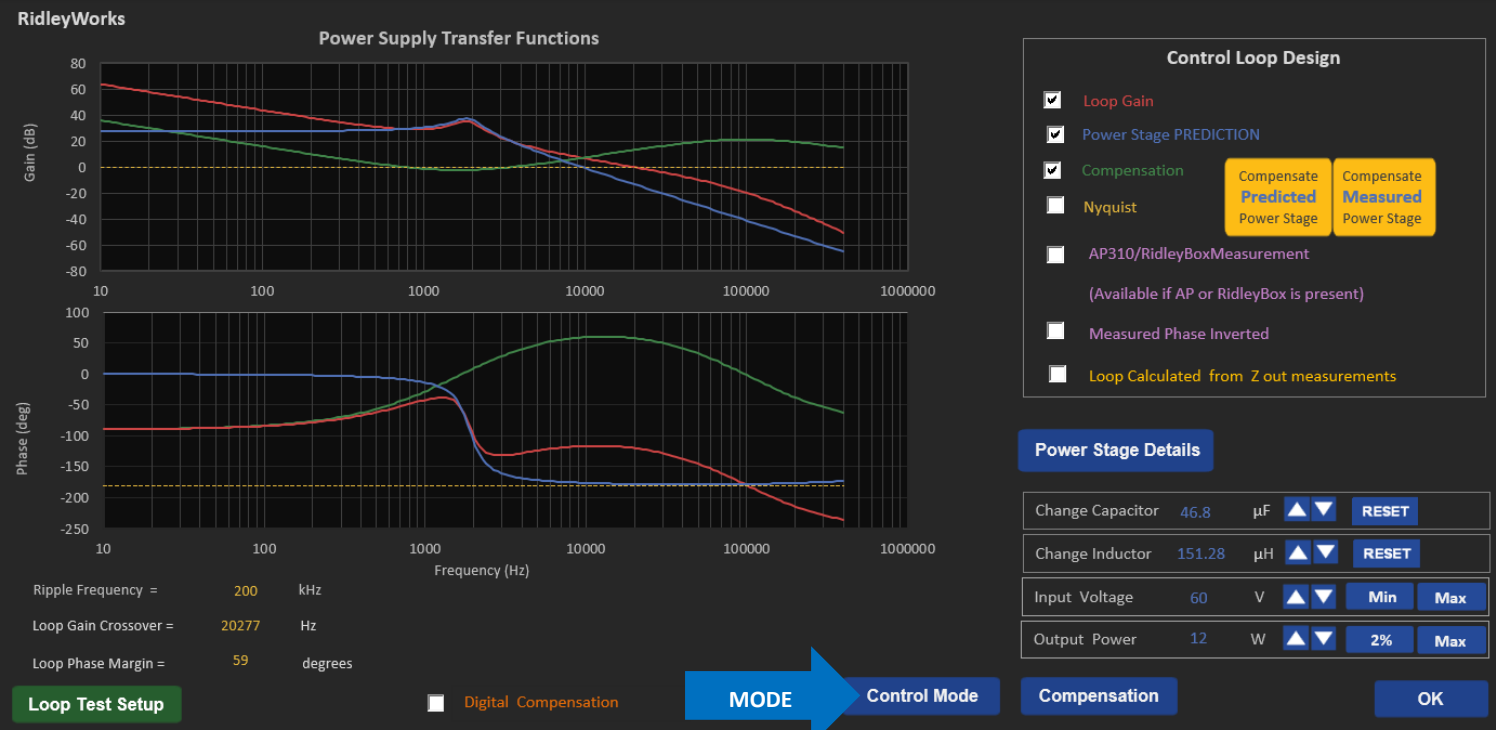
**Fig. 3.2: Current-Mode Control Transfer Function Bode Plots in RidleyWorks®**

Note that when you change the inductor and capacitor values, each time you click on the button, the control loop is optimized for the new power stage and the transfer function plotted for the updated design. RidleyWorks® will automatically detect whether the converter is operating in CCM or DCM and provide the appropriate transfer function. If you wish to see details of the converter you can click on the [Power Stage Details](#) button and it will give you the poles and zeros.

The [Compensation](#) button is a feature of RidleyWorks® that lets you interactively move the shape of the compensation values while looking at the shape of the loop at the same time. Figure 3.2 shows the transfer functions for current-mode control, the default control scheme chosen for all converters except the half-bridge.

Figure 3.2 shows the control transfer functions which are available in RidleyWorks®. You can select multiple plots to be shown all at the same time as shown in the box above. Measurements from the AP300 Frequency Response Analyzer can also be overlaid on the graph to compare with predictions.

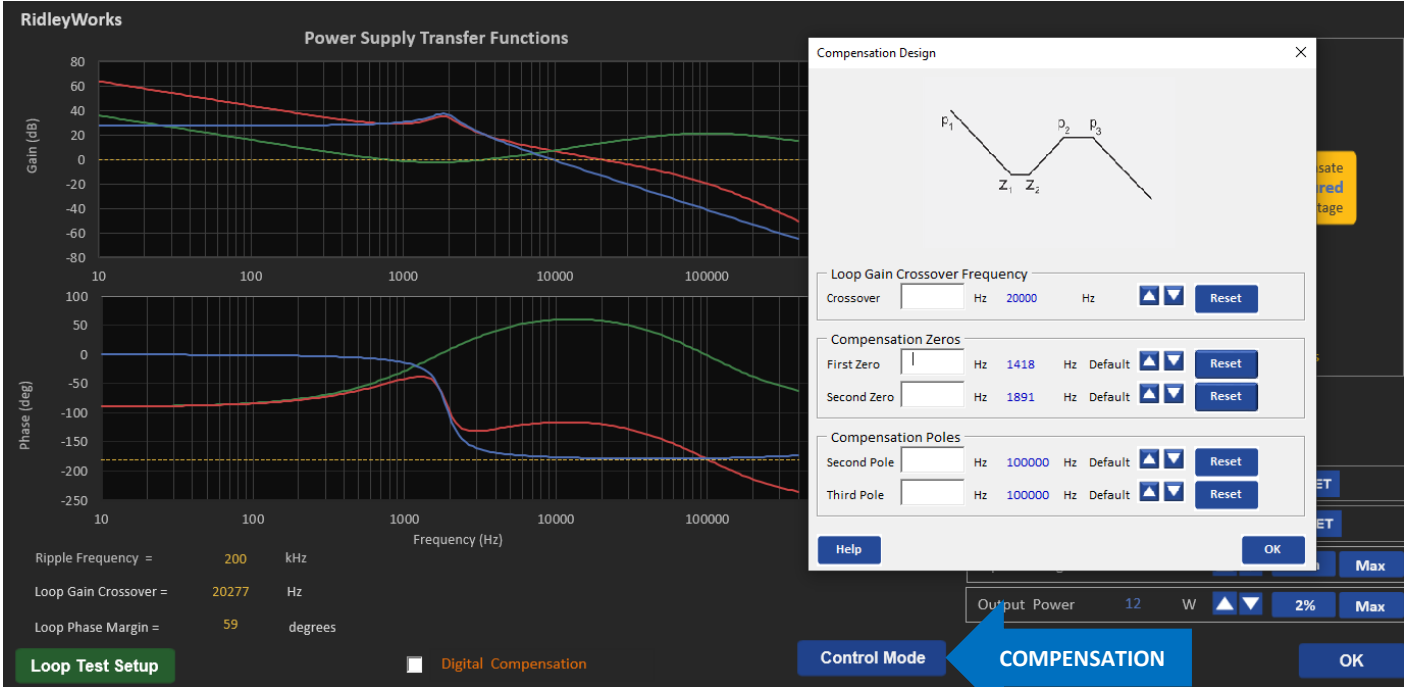
You can do many things on these graphs, including moving the line and load up and down, and changing the values of the inductor and capacitor in the converter. As you change these values, you can see the immediate effect on the power stage, compensation and loop gain.



If you click on the Control Mode button, you can change your selection to voltage-mode control. Once you do this, the compensator will be re-optimized and the resulting transfer functions plotted.

Fig. 3.3: Voltage-Mode Control Transfer Function Bode Plots in RidleyWorks®

### 3.3 Interactive Compensation Adjustment



It is possible to look at the control transfer functions of RidleyWorks® while adjusting the pole and zero locations of the compensation, as shown in the figure above. Each time you click the **Adjust** button the zero or pole will be moved, new compensation components selected, and the curves replotted.

**Fig. 3.4: Interactive Adjustment of Compensation Poles and Zeros with Transfer Functions**



You can also interactively change the compensation parameters while looking at the simulation waveforms. This is a powerful and completely unique feature of RidleyWorks®. No other design or simulation program has this capability.

**Fig. 3.5: Interactive Adjustment of Compensation Poles and Zeros with Simulation Waveforms**



### 3.4 Voltage Feedforward Control

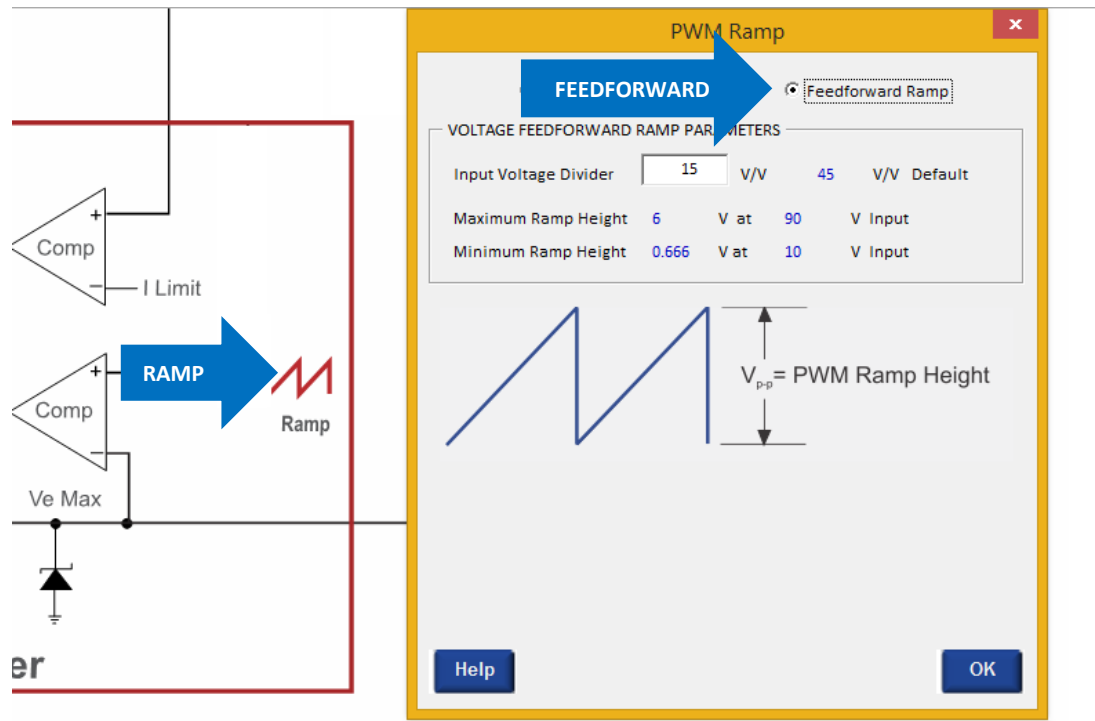


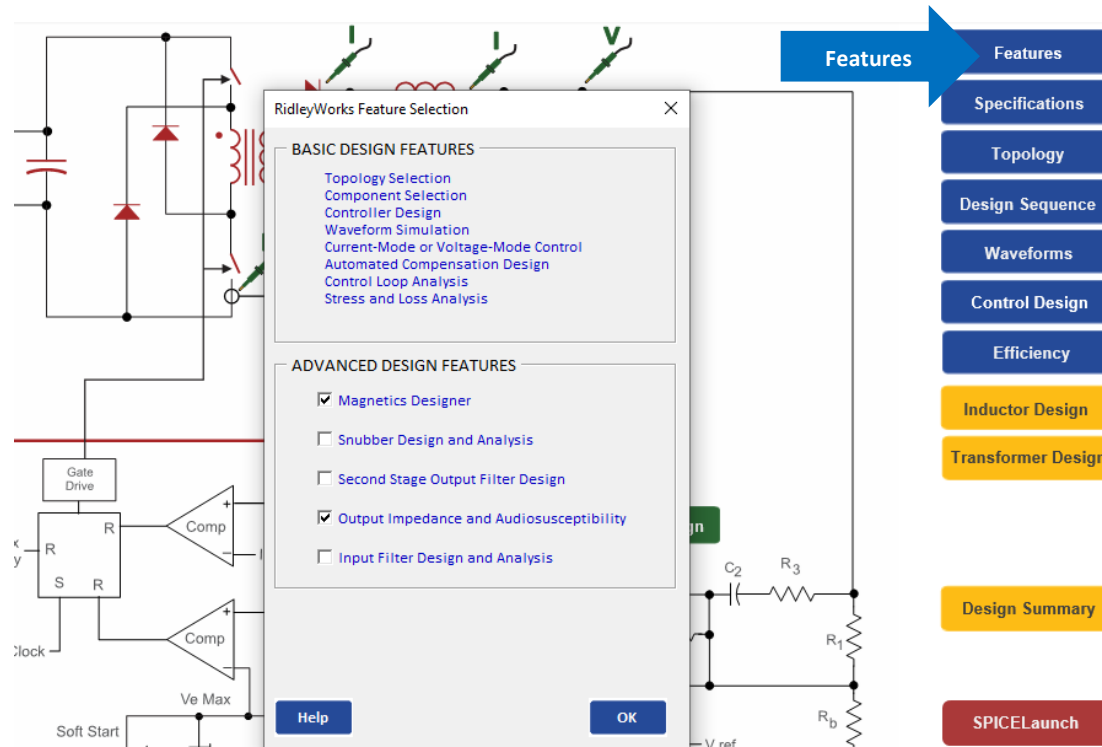
Fig. 3.5: Selecting Voltage-Mode with Feedforward

It is possible to implement voltage-mode control with feedforward. The normal implementation of this technique in the control chip is to make the control ramp proportional to the input voltage, scaled by an input voltage divider constant,  $k$ .

When the input voltage is increased, the ramp size will increase, and there will be an immediate change in duty cycle. This can greatly improve the transient response with changes in input voltage. Normally it will be at least an order of magnitude better, especially for buck-derived converters.

In addition, the changing ramp will make sure that the loop gain of the system stays fixed for different input voltages. This can be very important for large ranges of input voltage, and compensation will remain optimal across the entire range. The converter will regulate much better, and step load response will be improved at low line.

# 4 Magnetics Designer



## 4.1 Turning on the Magnetics Designer

On the Power Stage design page, clicking on the Features button allows you to enable the magnetics design features of RidleyWorks®. You will then see the Inductor Design and Transformer Design button appear on the screen.

Click on either of these to start your magnetics design process.

**Fig. 4.1: Selecting the Magnetics Design Option Enables the Inductor and Transformer Designers**

# 4.2 Inductor Design Page



Fig. 4.2: Inductor Design Page Guides You Through Inductor Details

### 4.3 Choosing the Inductor Core Size

Inductor Design

INDUCTOR CORE SELECTION

Approximate Core Area 0.28 sq. cm

Generic Core

Select Core

Use Selected Core

CORE TYPE AND AREA

Core Type Generic Inductance 151.28  $\mu$ H

Minimum Area 0.28 sq cm

MAXIMUM FLUX LEVEL SETTING

0.3 T

Reset

$\Sigma$

This value will determine the number of turns needed

INDUCTOR TURNS REQUIRED

Turns 44 Update

Maximum Flux Density at Peak Current Limit: 0.2991 T

CORE PHYSICAL DETAILS

Core Volume 0.88 cu. cm

Window Width 0.79 cm

Window Build 0.26 cm

Inner Turn Length 2.46 cm

Outer Turn Length 4.09 cm

CORE GAPPING

0.45 mm 18 mills

Note that gap length is a rough guide only. Manufacturing design is specified by AI, and the core is gapped to achieve the correct value

AI value 78.14 nH/n2

Help

OK

Core, turns

Turns Core Gap

Clear Inductor

Core Material

Winding

Winding AC Model

Inductor Impedance

Test Setup

Return to Schematic

Pushing the top button on the inductor design page will bring up the form for selecting the design parameters of the inductor. By default, a generic core will be selected with a recommended core area. This lets you choose any specific core that you or your company likes to use that has approximately that area. A larger core area will run a little cooler, and a smaller core size will have higher dissipation. Your actual selection will depend on your cooling situation, and your need for efficiency versus size.

If you click on the Select Core button, you will be presented with an array of standard cores to use, and you can also add your own favorite cores very easily to the program.

Fig. 4.3: Inductor Core and Turns Selection

35 | Ridley Works Manual

# 4.4 Choosing the Inductor Core Material

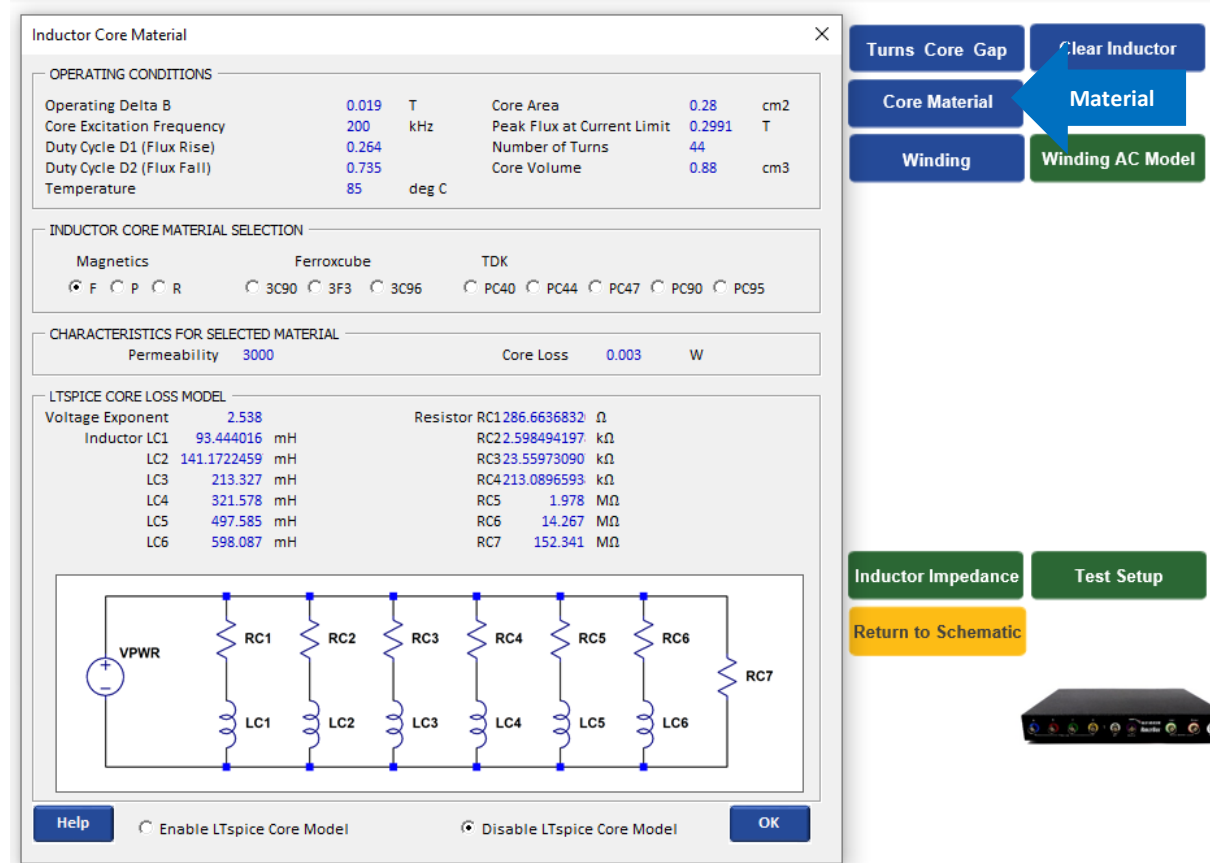


Fig. 4.4: Inductor Core Material Selection

You can use any core selection with a variety of core materials. RidleyWorks® incorporates a unique core loss modeling techniques that accurately give the loss for different materials. Variable Steinmetz equation coefficients are used to produce a much more accurate core loss estimate than is available from the manufacturers. Details of the techniques used to calculate the loss are given in the Ridley Engineering Design Center (<http://www.ridleyengineering.com/design-center.html>) in paper **[A03] Modeling Ferrite Core Losses**.

If you are using a different core material to those shown, RidleyWorks® gives you the proper frequency, flux level, and Delta B to be used with core loss curves.

RidleyWorks® also selects the values of an RL network to model the losses in LTspice®. This is a nonlinear frequency-dependent loss model that allows you to run an LTspice® simulate and click on the core to see the losses. There is no need to use core-loss curves.

# 4.5 Designing Inductor Windings

Inductor Winding Design

INDUCTOR DESIGN

Inductor Value

152.95

uH

Number of Turns

25

AVAILABLE WINDOW

Margin Required

0

mm each end

Insulation Required

0

mm total

2.63

cm

winding length

0.87

cm

build

WINDING STRUCTURE

Use Magnet Wire

Use Foil

Helical Foil

Number of Layers

1

Minimize Loss

Number of Parallel Wires

1

Maximum Conductor Size That Will Fit Exactly is

19

awg

Your Choice of Conductor Size

19

awg

Update

Minimize Loss

WINDING LOSS DETAILS

RMS Current	1.00148	A	Winding Loss (Including Proximity)	0.049	W
DC Current	0.99915	A	Winding Loss (without Proximity Loss)	0.048	W
AC Current	0.06827	A	Winding Surface Area	20.42	sq.cm
DC Resistance	4.79E-02	Ω			
AC Resistance	2.84E-01	Ω			

Proximity Loss

Sweep AC Resistance

Σ

Help

OK

Gap

Clear Inductor

Material

Winding

Model

Inductance

Test Setup

Schematic

Pushing the Winding button will show you the details of the inductor winding. You can build inductors with foil, helical windings, or with wire, and RidleyWorks® will help you to optimize each of these. The foil winding thickness can be optimized with proper consideration for proximity losses, by clicking the Minimize Loss button.

Details of the sophisticated techniques used to calculate the proximity loss are given in the Ridley Engineering Design Center ([www.ridleyengineering.com/design-center.html](http://www.ridleyengineering.com/design-center.html)) in numerous papers on magnetics design and analysis.

Fig. 4.5: Inductor Winding Design

37 | Ridley Works Manual

## 4.6 Designing Transformers

Transformer Design

TRANSFORMER CORE SELECTION

Approximate Core Area0.631sq. cm

☒ Generic Core

Select Core

☐ Use Selected Core

CORE TYPE AND AREA

Core TypeGeneric

Minimum Area0.631sq. cm

MAXIMUM FLUX LEVEL SETTING

▲▼

0.3T

Reset

This value will determine the number of turns needed

TRANSFORMER TURNS

Primary Turns49

Update

Reset

Integer Turns

Secondary Turns6.69

Magnetizing Inductance3.5114mH

Worst Case Flux Density0.2958T (Maximum input Voltage and Maximum Duty)

CORE PHYSICAL DETAILS

Core Volume3cu. cm

Window Width1.19cm

Window Build0.397cm

Inner Turn Length3.419cm

Outer Turn Length6.59cm

Help

Once you select OK, values appearing on this form will be used in the converter simulation.

OK

Turns and Core

Core Material

Clear Transformer

Primary Winding

Primary ACR

Secondary Winding

Secondary ACR

Primary Impedance

Test Setup

Return to Schematic

Fig. 4.6: Transformer Design Page

RidleyWorks® has a powerful algorithm for designing transformers, incorporating advanced core loss calculation, proximity loss analysis, core and winding selection, all in an easy-to-use package. The design program interfaces closely with simulation waveforms to provide fast, efficient, and reliable designs.

The sequence of design is similar to that of the inductor. Start by clicking the Turns and Core button to begin the design.

38 | Ridley Works Manual

# 5 SPICELaunch™

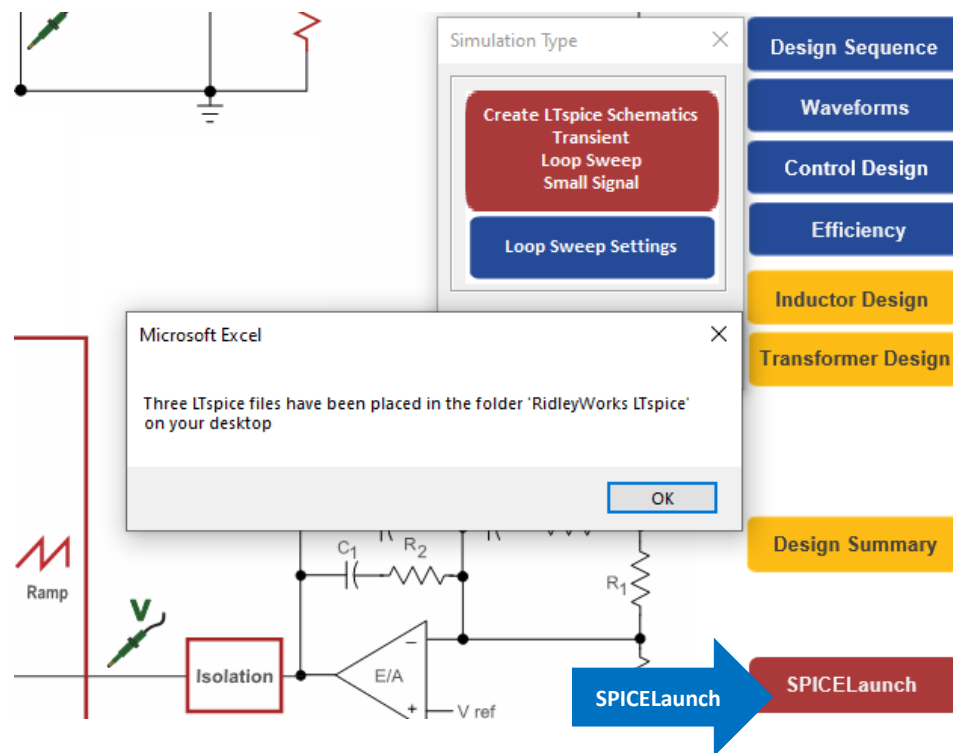


Fig. 5.1: Running SPICELaunch™

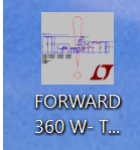
## 5.1 Preparing an LTspice® File

Designers often spend a lot of their development time in creating LTspice® models for their converters. This is a process that requires experience with magnetics models and power controllers. A powerful feature of RidleyWorks® is the ability to automatically generate LTspice® schematic files with just the click of a button.

When you click on the SPICELaunch® button, you will have the option to create LTspice® schematics, or to adjust loop sweep settings. When using this feature for the first time, click on the button to create the files, and you will see the message that three LTspice® files have been created. These can be found in the folder RidleyWorks LTspice that can be found in the C:\Ridleyworks directory.



# 5.2 Simulation Settings in LTspice®



If you double click on the LTspice® TRAN file that is in the folder on your desktop, it will automatically open LTspice® .

Before running a simulation, it is recommended that you set some parameters of LTspice® for better simulation accuracy. Select [Tools Control Panel Hacks!](#) and set the check boxes as shown below.

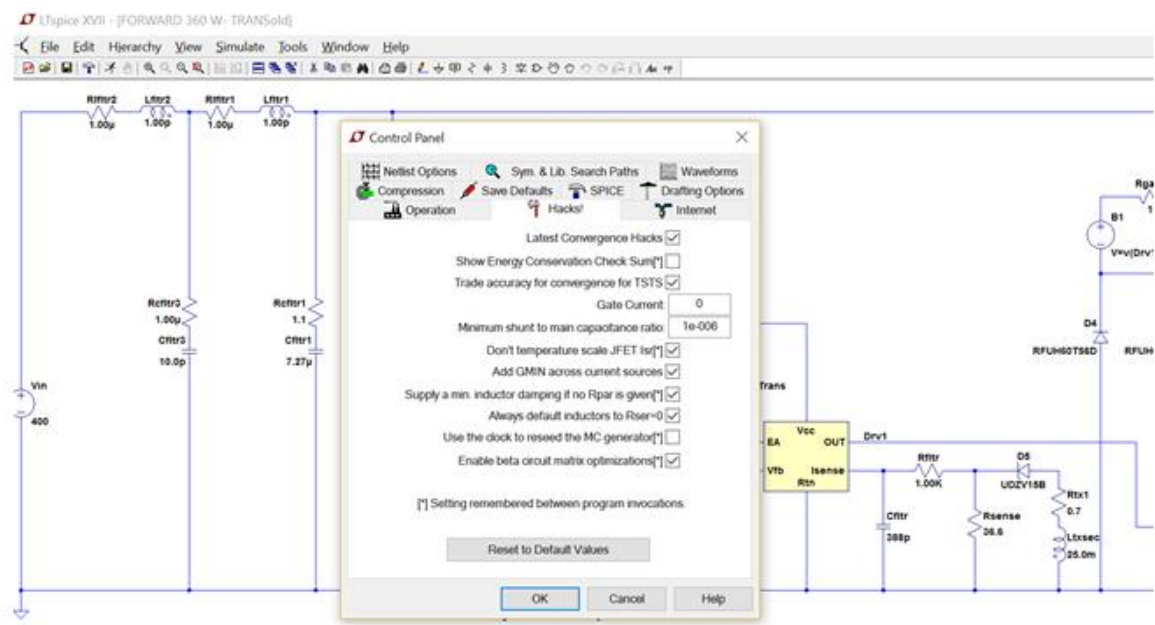
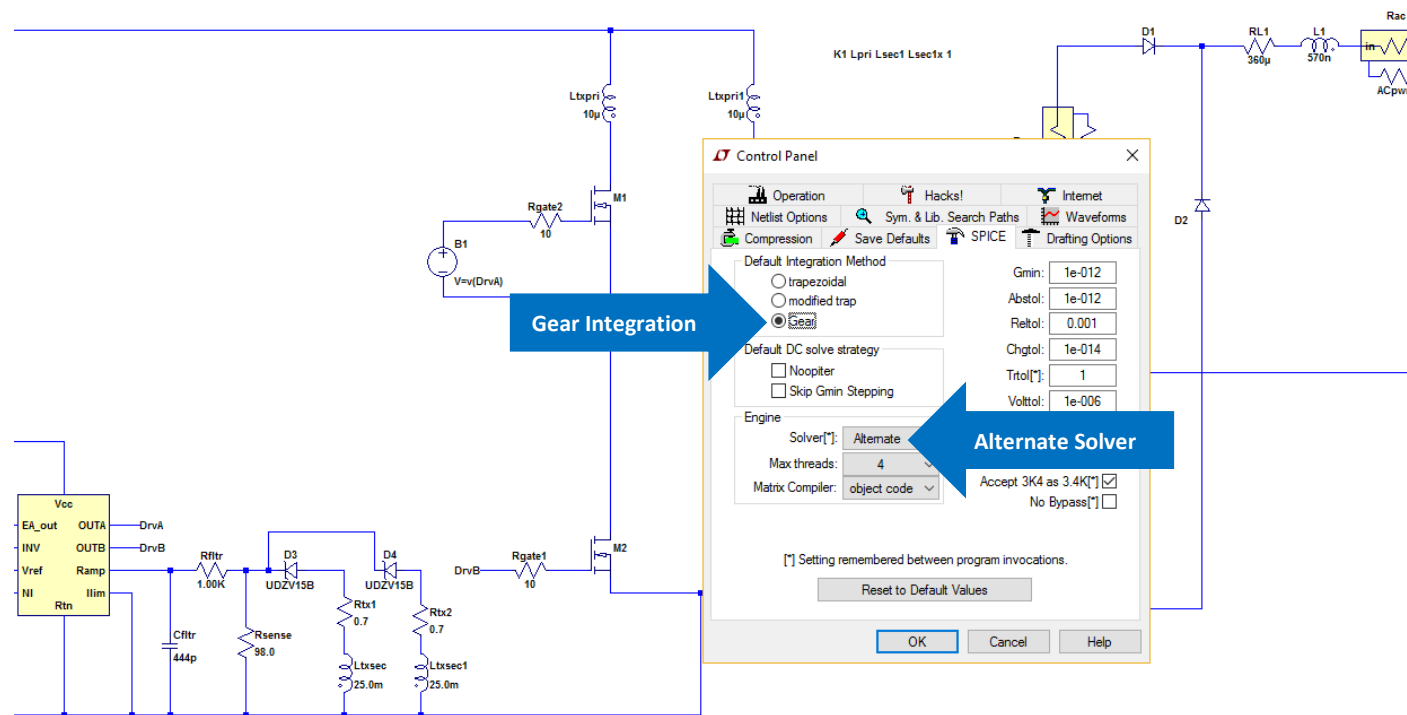


Fig. 5.2: Setting Up LTspice® Hacks!

LTspice®, like all versions of spice, may sometimes struggle to converge during a simulation, and you may need to adjust other simulation parameters to help it along. This is one of the advantages of the RidleyWorks® simulation – it never has convergence issues and simulations are almost instantaneous.



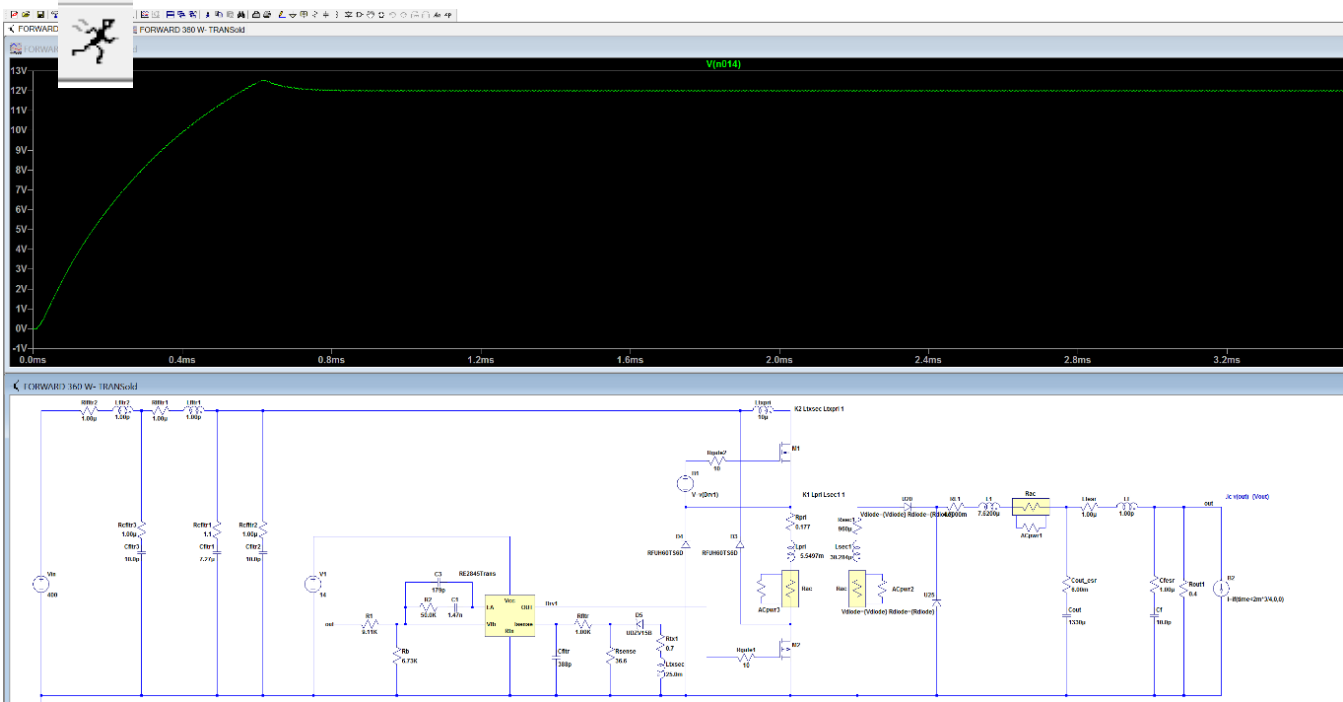
It is also recommended that your Solver Engine should be selected as **Alternate**. This will give greater simulation accuracy. (The **Normal** setting will simulate faster which may be useful when generating the Bode plots inside LTspice®.)

We also find that the **Gear** integration routine works better for transformer isolated converters, but you might want to sometimes try the default settings if you run into difficulty. The default setting is Modified Trap, and that may work better for some circuits.

The integration method settings are not retained in LTspice® each time you exit the program, so you will need to set them each time you come in.

**Fig. 5.3: Setting Up LTspice® simulation parameters**

# 5.3 Simulating Transients in LTspice®

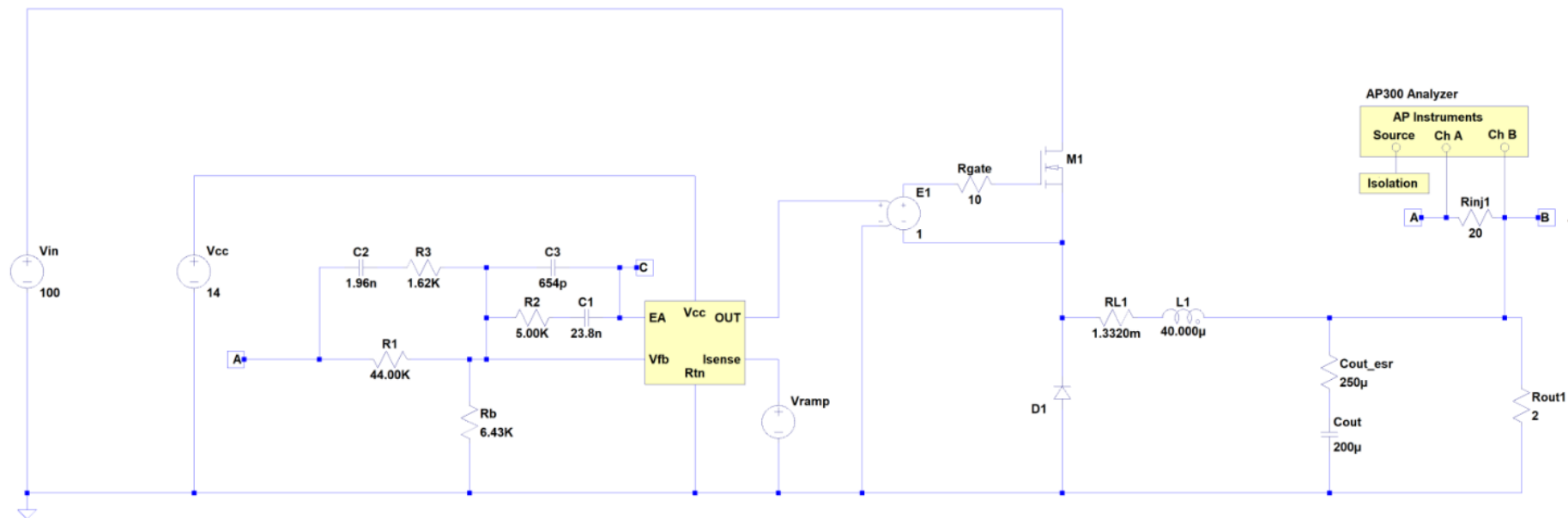


When you click on the [running man](#) icon, the spice simulation will start. You can then probe any waveform on the circuit. The simulation above is the output voltage during start-up of a circuit. Although the spice simulation may be slow, you have already done all of the design of the control circuit and power parts in RidleyWorks®, and you can now use LTspice® for the things it is good at. You can add any arbitrary circuits, any devices, and move your design much closer to the production schematic.

You will find that the combination of RidleyWorks® and the rapid generation of the schematics for LTspice® changes the way that you design. It should greatly speed up your process.

**Fig. 5.3: Running LTspice® Transient Simulations**

# 5.4 Generating Swept Bode Plots in LTspice®



**Fig. 5.4:** LTspice® Sweep Schematic for Bode Plots

A powerful new feature of RidleyWorks® is the generation of schematics which can produce Bode plots *directly* from the time-domain schematic. There is *no need to simulate small-signal equivalent circuits* which are, by definition, approximations of the real circuit. For the first time, computers (in 2020) are fast enough that the intensive simulation needed for Bode plot generation can be done in a reasonable amount of time. This kind of work was only possible up until now with specialized high-speed simulation programs.

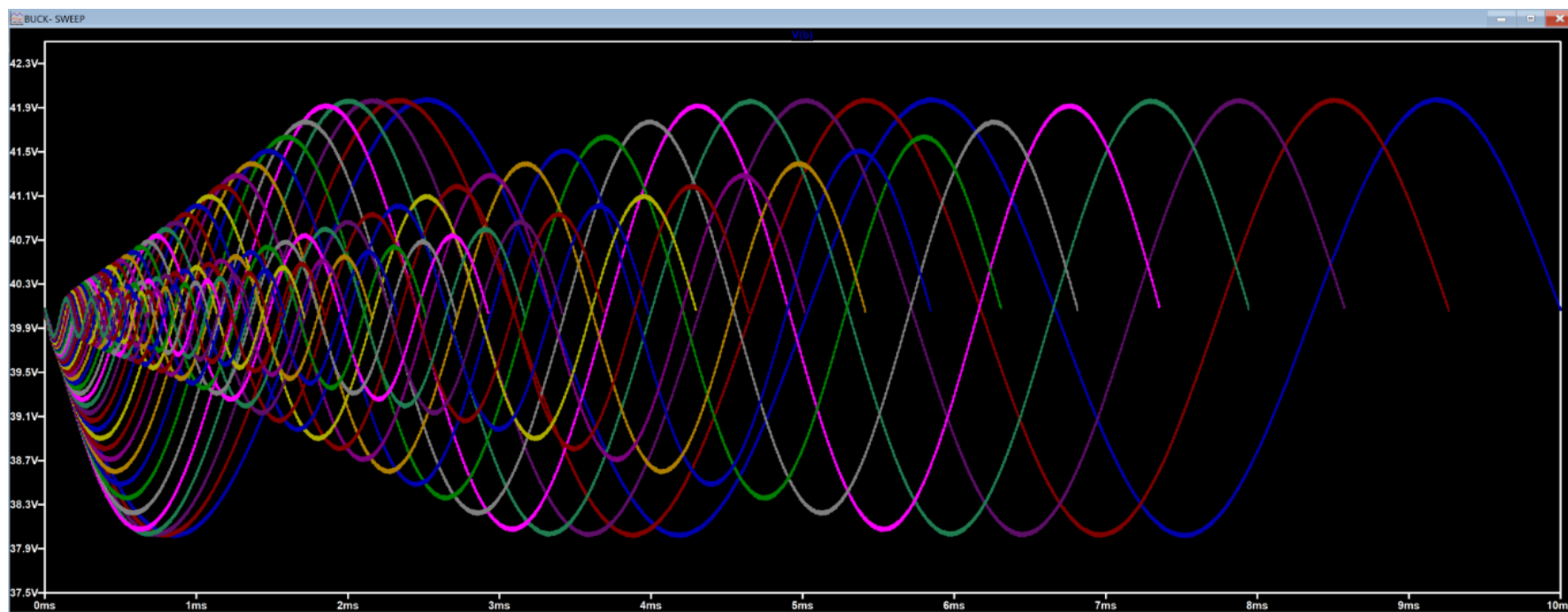
How fast can this be done? At the time of writing this manual, our record for simulation is just 24 seconds for a buck converter. This kind of speed is produced with the following features in place:

1. Automatic generation of a loop compensation that is ruggedly stable
2. Emulation of the features of the AP300 analyzer with frequency-dependent source, automatically configured to give good results.
3. Proper setup of simulation times for steady-state solution.

For the example buck schematic above, the bode generation is initiated by clicking on the running man icon. If you then click on **Point B** of the schematic, you can watch the waveforms being generated ready for measurement. This is done in a very similar way to the powerful AP300 analyzer.

Generation of the sweep schematic is a single button click in RidleyWorks®, and there is no need to change any settings from the default file that is generated. Figure 5.5 shows the repeated sweep waveforms generated in LTspice®. It is very instructional to watch this procedure to enhance your understanding of how the AP300 frequency response analyzer works.

These waveforms are generated by clicking on the Run icon, then clicking a probe on the output voltage of the converter.



**Fig. 5.5: LTspice® Sweep Waveforms at Point B of the Circuit.**

Once the sweeps shown above are done, and the data has been processed by LTspice®, click on [View](#) then [Spice Error Log](#). (If the Spice Error Log option is greyed out, LTspice® has not yet finished processing the simulation data.) A file similar to that shown in Figure 5.6 will then appear:

```

Direct Newton iteration for .op point succeeded.
.step freq=300
.step freq=323.933
.step freq=349.774
.step freq=377.678
.step freq=407.807
.step freq=440.34
.step freq=475.468
.step freq=513.398
.step freq=554.355
.step freq=598.579
.step freq=646.33
.step freq=697.892
.step freq=753.566
.step freq=813.682
.step freq=878.593
.step freq=948.683
.step freq=1024.36
.step freq=1106.08
.step freq=1194.32
.step freq=1289.6
.step freq=1392.48
.step freq=1503.56
.step freq=1623.51
.step freq=1753.02
.step freq=1892.87
.step freq=2043.88

```

Right click in the space to the right of the numbers on the screen and select the option **Plot .Step'd meas data**. Answer **Yes** to the question that pops up about complex data, and you will see Bode plot Axes appear. If you right click in the plot area, you can then add traces, and the **Loop Gain**, **Power Stage Gain**, and **Compensation Gain** are available as options.

Figure 5.6 shows the sweep for a buck power stage gain. Notice that approaching half the switching frequency, noise is apparent on the sweep. This noise characteristic will be very familiar to anyone who has used the AP300 or the RidleyBox on a switching power supply.

**Fig. 5.6: LTspice® Error Log**

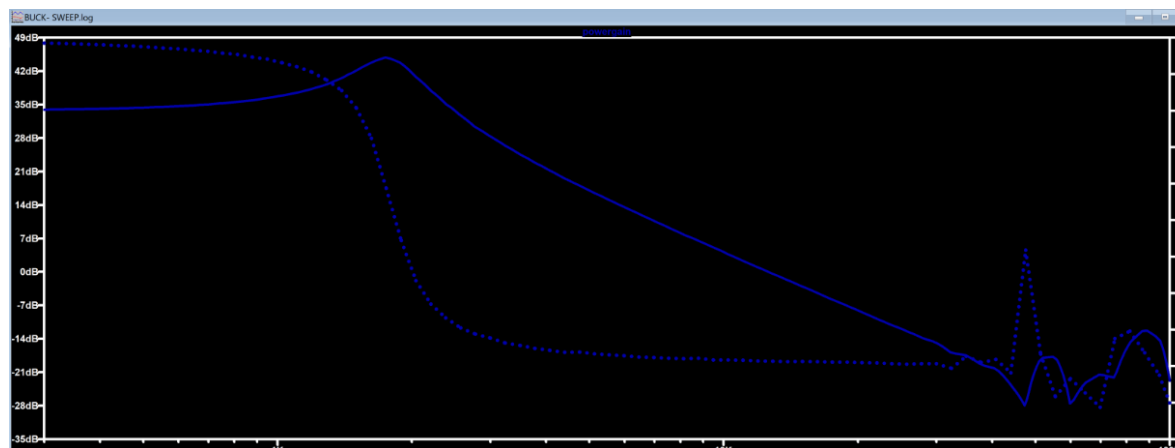


Figure 5.7 shows the sweep for a buck power stage.

**Fig. 5.7: LTspice® Bode Plot of a Buck Power Stage Gain**

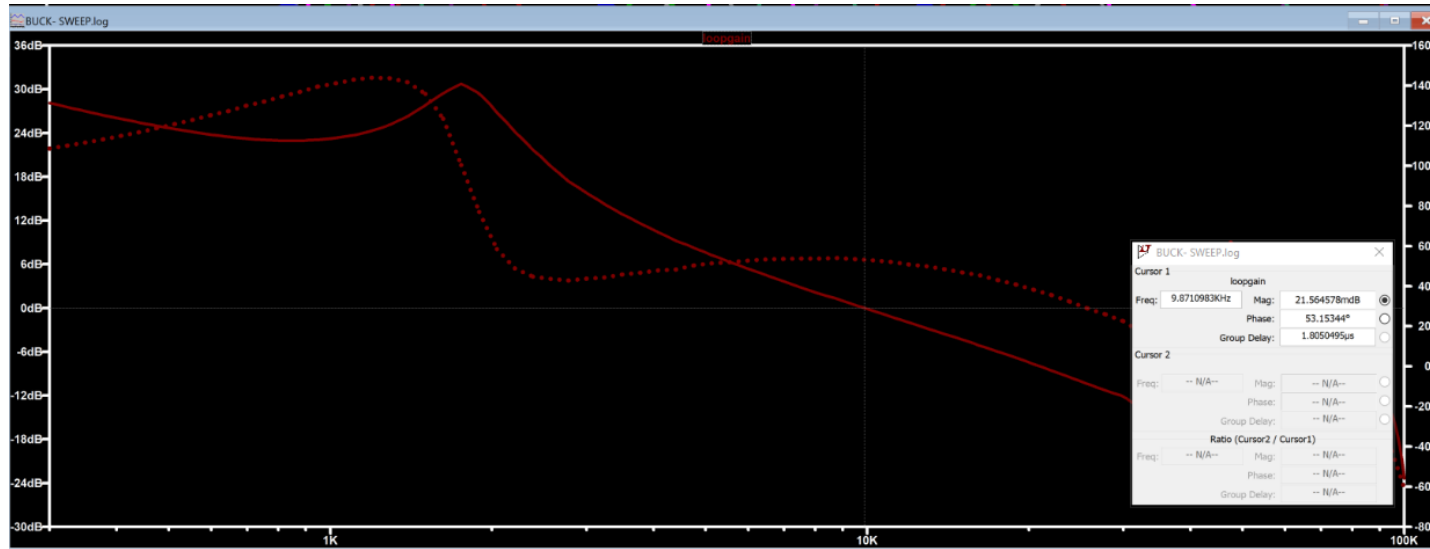


Figure 5.8 shows the sweep for a buck power loop gain. Notice that the phase of this plot is reading phase *margin*, not phase. The cursors on the loop gain show the crossover frequency and the phase margin at this crossover.

**Fig. 5.8: LTspice® Bode Plot of a Buck Loop Gain**

The ability to sweep the transfer functions on the large-signal schematic inside LTspice® or PSIM is a powerful tool for finding the bode plots of circuits that don't have small-signal models available for them and we encourage you to explore this feature of the software.

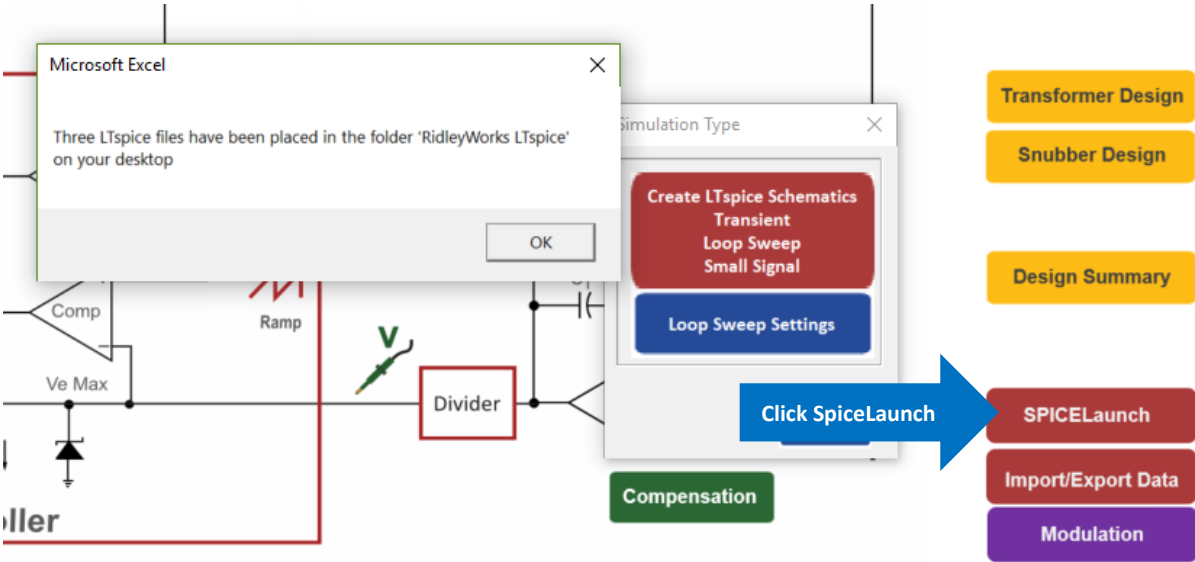
However, there are times when technique this cannot be used, such as:

- The circuit is not stable
- The desired sweep frequency is too low (10 Hz is very time consuming in LTspice®, not an issue in PSIM)
- Characteristics of the ac sweep are below the noise-floor resolution of the technique.

The next section of this manual describes another method of obtaining loop gains and other transfer functions via the small-signal models.



# 5.5 Generating Small-Signal Bode Plots in LTspice®



The major modern work on small-signal models for power converters was initiated by Dr. Vatché Vorpérian with the publication of the PWM Switch Model in 1988. Models were completed by Dr. Ray Ridley with the publication of his Current-Mode Model in 1990. If you are new to power supply design, these models can take some time to become familiar with and implement in your circuit simulator. We have fully automated this process for the first time, putting small-signal analysis at every engineer's fingertips.

Release 12.35 of RidleyWorks® exports a small-signal model for use with LTspice®. This provides a rapid high-speed design and simulation environment for the feedback networks. All characteristics of input and output filters will be seen in the loop characteristics, providing more detailed sweeps than those available in RidleyWorks®.

**Fig. 5.9: Exporting an LTspice® Small-Signal File. Three files are automatically exported: a transient simulation, a swept-sine simulation, and a small-signal model.**

RidleyWorks LTspice Schematic 032819

### BUCK FAMILY

Vin = 25.0, Vout = 30, Iout = 0.1 two-switch forward converter, voltage-mode

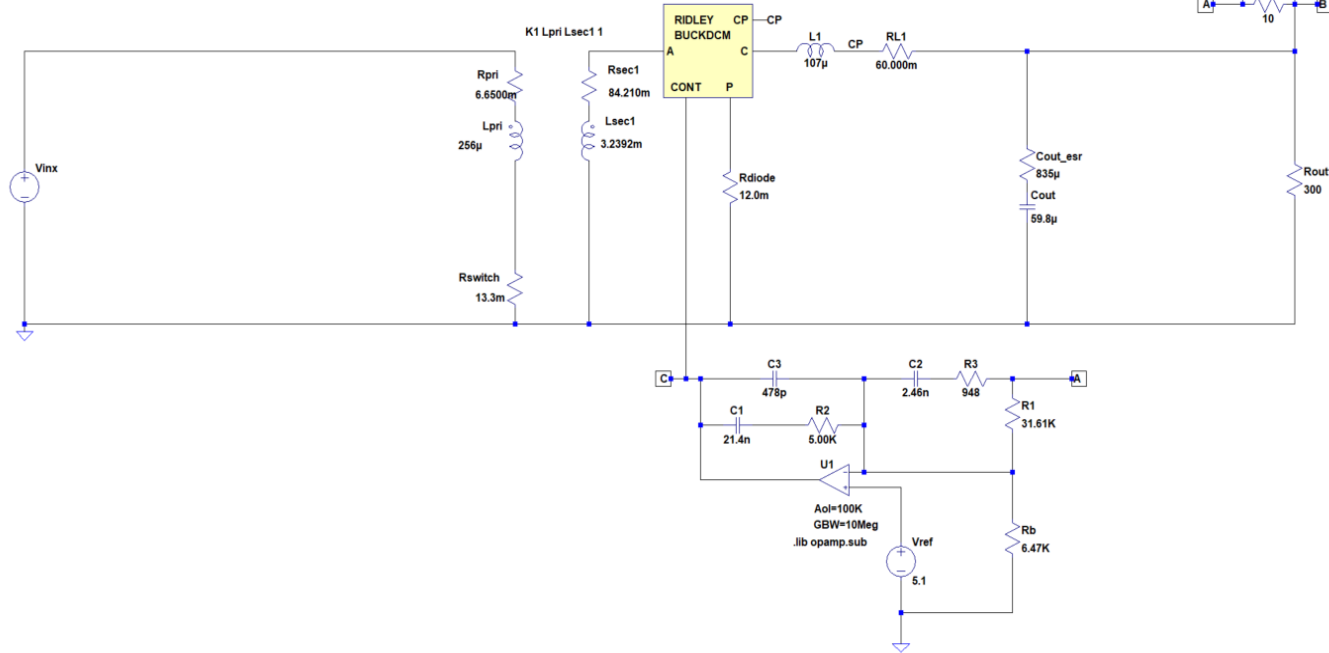


Fig. 5.9: Buck Family Small-Signal Model Exported to LTspice®

Figure 5.5 shows the buck family configuration of the small-signal model in LTspice®. There are three fundamental families that cover all of the topologies in RidleyWorks® - the buck, boost, and buck-boost. Each of these circuits contains a single small-signal element, in this case the Ridley BuckDCM. When the file is exported to LTspice®, RidleyWorks® determines whether the circuit is going to be analyzed in CCM operation or DCM operation.

Notice that the model is invariant for both voltage-mode and current-mode control. This is very important – our small-signal model is the only one that works in both regions. As the compensating ramp for current mode is increased, it is possible to get a system that operates in between voltage mode and current-mode, but the model gives the right results regardless.

RidleyWorks LTspice Schematic 032819-A  
BOOST FAMILY  
 $V_{in} = 25.$ ,  $V_{out} = 30$ ,  $I_{out} = 5$  two-switch forward converter, voltage-mode

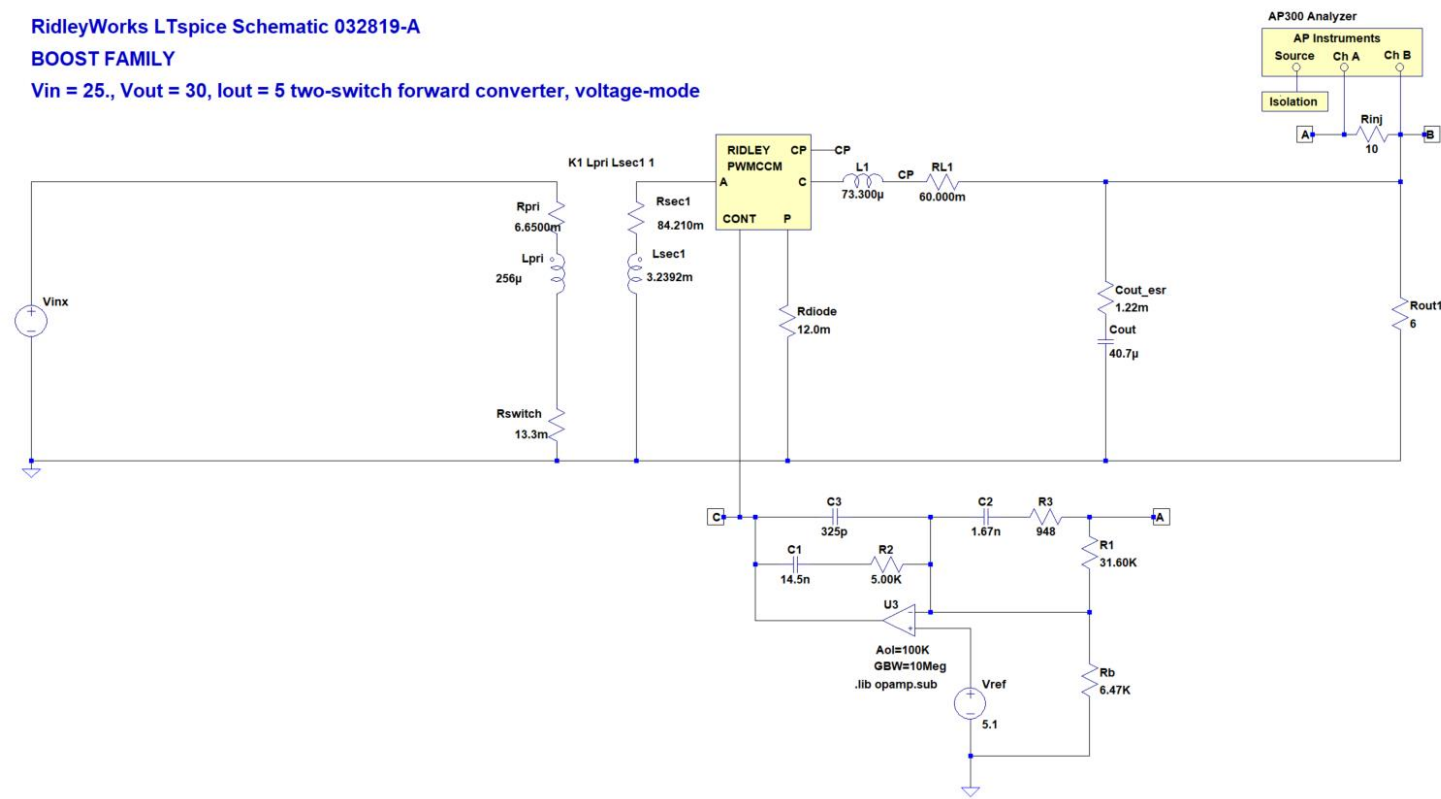


Fig. 5.10: Boost Family Small-Signal Model Exported to LTspice® . Notice the PWM block is the same subcircuit as for the buck converter family.

# RidleyWorks LTspice Schematic 032819-A

## BUCK-BOOST FAMILY

Vin = 20., Vout = 30, Iout = 5 flyback converter, voltage-mode

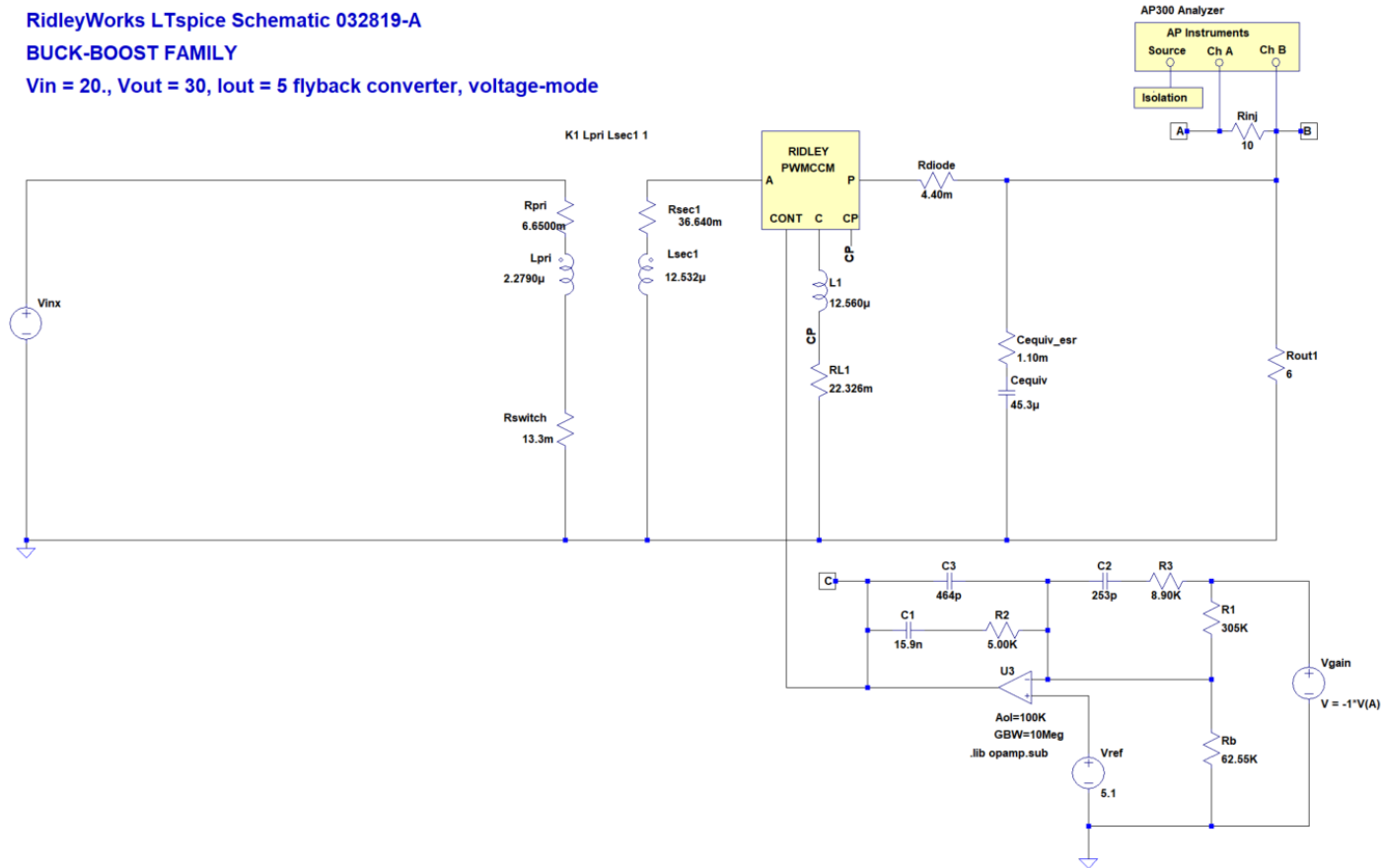


Fig. 5.11: Buck-Boost Family Small-Signal AC Model Exported to LTspice® . Notice the inversion of the output voltage signal in the feedback path.

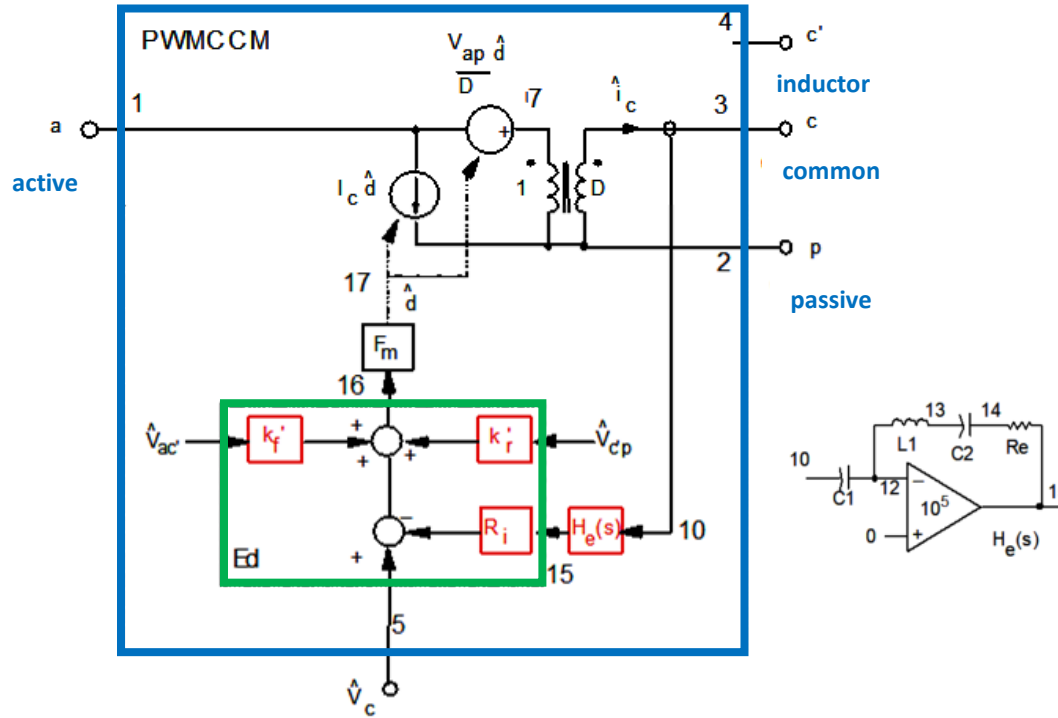
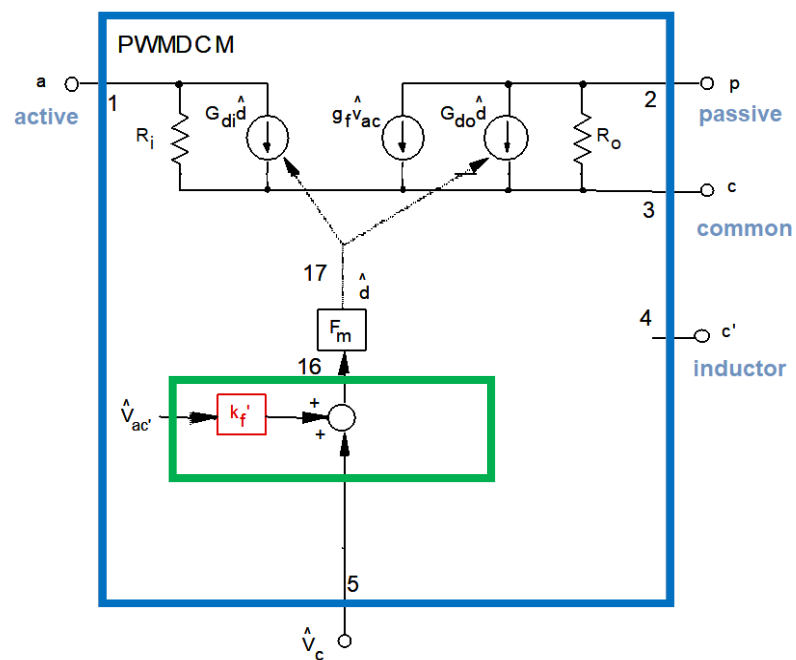
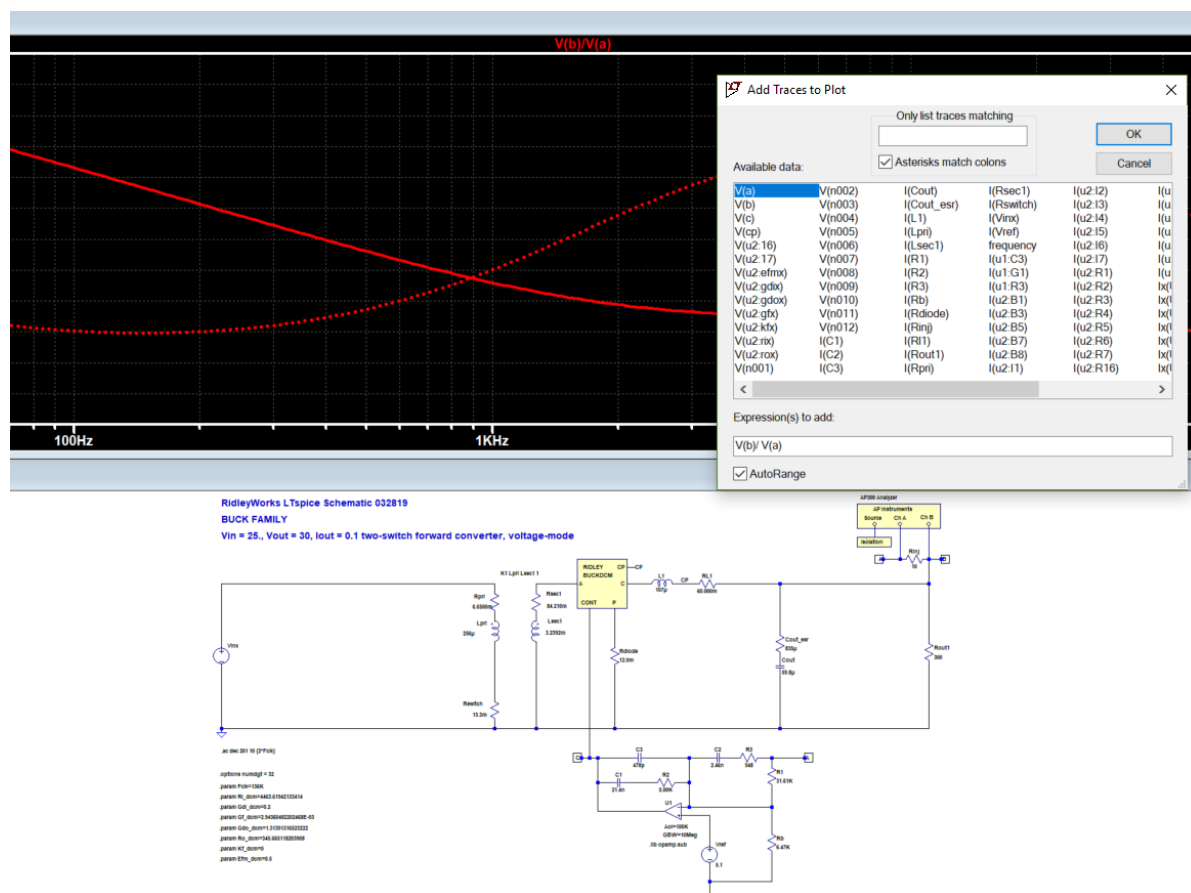


Fig. 5.12: Invariant Small-Signal Subcircuit for Current- or Voltage-Mode control in CCM. For pure voltage-mode control, the gain blocks inside the green rectangle are all zero.



**Fig. 5.13: Invariant Small-Signal Subcircuit for Current- or Voltage-Vode Control in DCM. For pure voltage-mode control, the gain block inside the green rectangle is zero.**



The figure above shows the loop gain plot for a buck-family converter. Running the sweep only takes a couple of seconds. You can then add any quantities to like to see desired transfer functions. The following are the most useful:

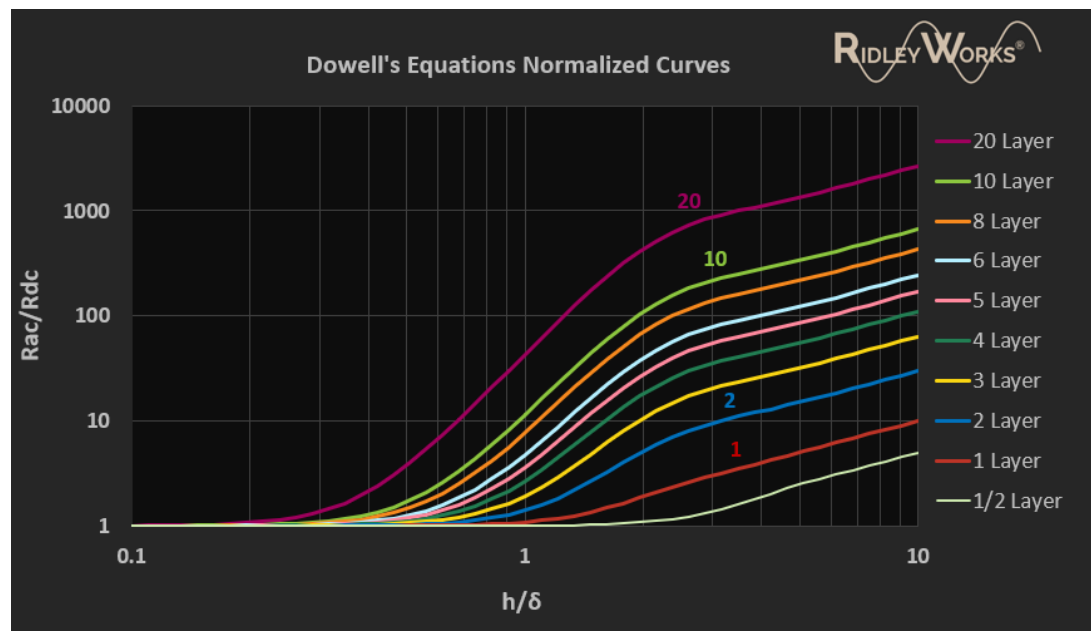
$V(b)/V(a)$  Loop Gain and Phase Margin

$V(b)/V(c)$  Power Stage Gain (specify -  
 $V(b)/V(c)$  for the buck-boost family

$-V(c)/V(a)$  Compensation Gain

**Fig. 5.14:** Plotting the loop gain of a converter consists of just clicking on run, then right click on the plot area. The quantity  $v(b)/v(a)$  will be the loop gain and phase MARGIN.

# 6 Advanced Proximity Loss and Core Loss



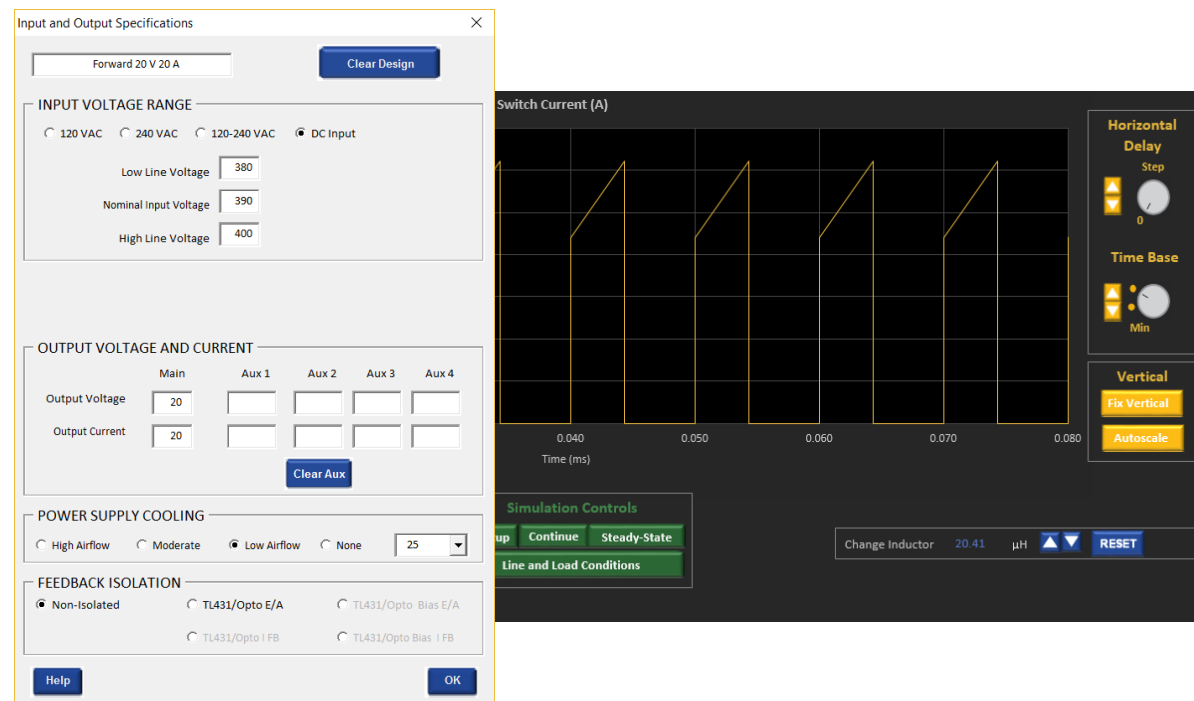
One of the biggest sources of error in designing a power supply is underestimating the losses in inductor and transformer windings. To get a reasonable estimate of how much loss there will be, it is necessary to apply Dowell's equations, a set of highly complex formulae that is usually the domain of PhD students in power electronics. It is estimated that less than 1% of practicing engineers ever use these techniques in their design work.

Fortunately, RidleyWorks® will do this for you. It does this in two ways – first by solving the Dowell's equation at the switching frequency to finding the increase in resistance. The winding current is then broken down into two parts – the dc current, and the ac current, which is the sum of all the harmonics. For a simple approximation, RidleyWorks® assumes that all the ac current is at this frequency. This will give a low estimate of the proximity loss, but it does provide good and fast guidance for how to organize the winding layers and winding size.

Secondly, to get more accurate loss in the windings due to *all* the current harmonics, RidleyWorks® first does a sweep of the Dowell's equation solutions from 1 kHz to 10 MHz. It then generates a circuit network that produces the same impedance versus frequency as the Dowell's equations. This network is exported to an LTspice® file where the proximity losses can be directly simulated in the time domain. This process is described in this chapter.



## 6.1 Estimating Proximity Losses in RidleyWorks®



The specifications for a 100 kHz two-switch forward converter are shown in the Specifications entry form for RidleyWorks®. This is for a 20 V, 20 A isolated converter running from a high-voltage input.

Figure 7.1 shows the simulated primary current for the converter. The rms primary winding loss, using the dc value of the resistance of the winding is calculated to be 0.491 W. This is the value that most designers would use for estimating temperature rise, but we will soon see that it is much too low.

**Fig. 6.1: Steady-State primary current from RidleyWorks® Simulation**

Transformer Design ×

**TRANSFORMER CORE SELECTION**

Approximate Core Area 1.229 sq. cm

☐ Generic Core Select Core ☒ Use Selected Core

**CORE TYPE AND AREA**

Core Type EC41

Minimum Area 1.06 sq. cm

**MAXIMUM FLUX LEVEL SETTING**

0.3 T ▲ ▼ RESET

This value will determine the number of turns needed

**TRANSFORMER TURNS**

Primary Turns 62 Update Reset Integer Turns

Secondary Turns 8

Magnetizing Inductance 10.6671 mH

Worst Case Flux Density 0.2982 T (Maximum input Voltage and Maximum Duty)

Figure 6.2 shows the important transformer parameters using and EC41 core. There are 62 primary turns, and 8 secondary turns.

**Fig. 6.2: Transformer parameters from RidleyWorks®.**


PRIMARY WINDING ALLOCATION

Window Allocation

0.5

Number of Turns

62



AVAILABLE WINDOW

Margin Required

0

mm each end

2.45

cm

Available Window

Insulation Required

0

mm total

0.2875

cm

WINDING STRUCTURE

☒ Magnet Wire

☐ Triple Insulated Wire

☐ Foil

Number of Layers

2

Minimize Loss

Number of Parallel Wires

1

Maximum Conductor Size That Will Fit Exactly is

22

awg

Your Choice of Conductor Size

22

awg

Update

☐ Split Primary Winding

Leakage Inductance

12.4502

μH

WINDING LOSS DETAILS

RMS Current	1.752	A	Winding Loss (Including Proximity)	2.908	W
DC Current	1.146	A	Winding Loss (without Proximity Loss)	0.491	W
AC Current	1.325	A	Winding Surface Area	16.5	sq.cm
DC Resistance	0.16005	Ω			
AC Resistance	1.53694	Ω			

Proximity Loss

Plot AC Resistance


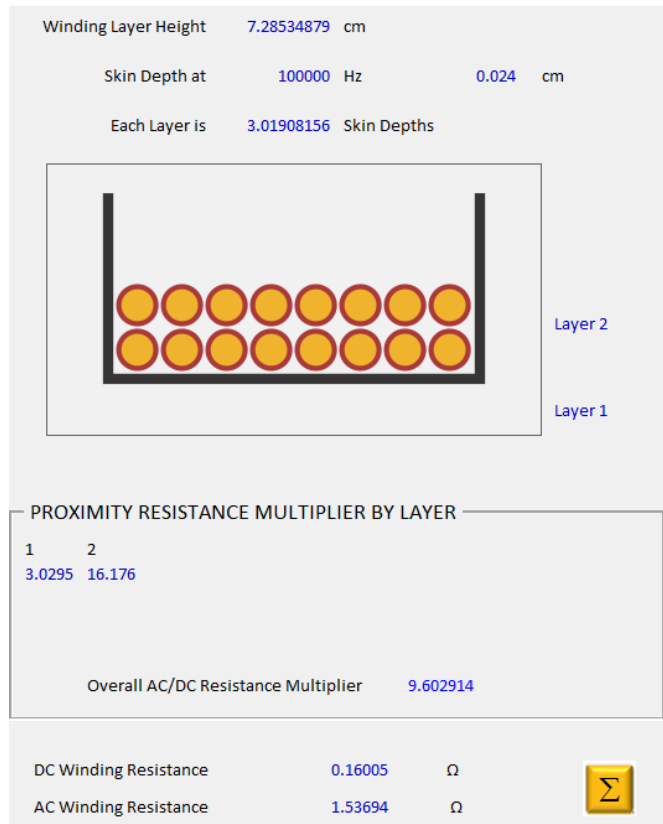


Fig. 6.3: Primary winding structure and analysis from RidleyWorks®

Figure 6.3 shows the structure and analysis of the primary winding of the transformer. Two layers of 22 awg wire are used to give a dc resistance of 160 mOhm. When this is multiplied by the square of the rms current, we see 0.491 W of loss.

However, the ac resistance of the winding, calculated at 100 kHz, is much higher at 1.54 ohms, almost 10 times the dc resistance. RidleyWorks® takes the rms of the ac current squared and multiplies by this number, adds to the dc current squared times the dc resistance, to get the total losses. You can see that they are much higher at 2.91 W. This is an increase of four times.

This more detailed and insightful information allows you to experiment with different numbers of layers, wire sizes, and parallel strands to see what gives the lowest loss combination. The **Minimize Loss** button automatically tries different layer counts to find the best solution for you.



Total power dissipation in a winding is given by :

$$P_d = b_w \sum_{i=1}^n l_i \frac{1}{h_i \eta_i \sigma} H_i^2 \left[ (1 + \alpha_i^2) G_{1_i} - 4 \alpha_i G_{2_i} \right]$$

The  $H$  field is calculated from

$$H_i = \frac{N_i I_i}{b_w} \quad b_w = \text{winding width}$$

Complex functions are needed to calculate the losses :

$$G_{1_i} = \Delta_i \frac{\sinh 2 \Delta_i + \sin 2 \Delta_i}{\cosh 2 \Delta_i - \cos 2 \Delta_i}$$

$$G_{2_i} = \Delta_i \frac{\sinh \Delta_i \cos \Delta_i + \cosh \Delta_i \sin \Delta_i}{\cosh 2 \Delta_i - \cos 2 \Delta_i}$$

The ratio of the winding layer height to skin depth is

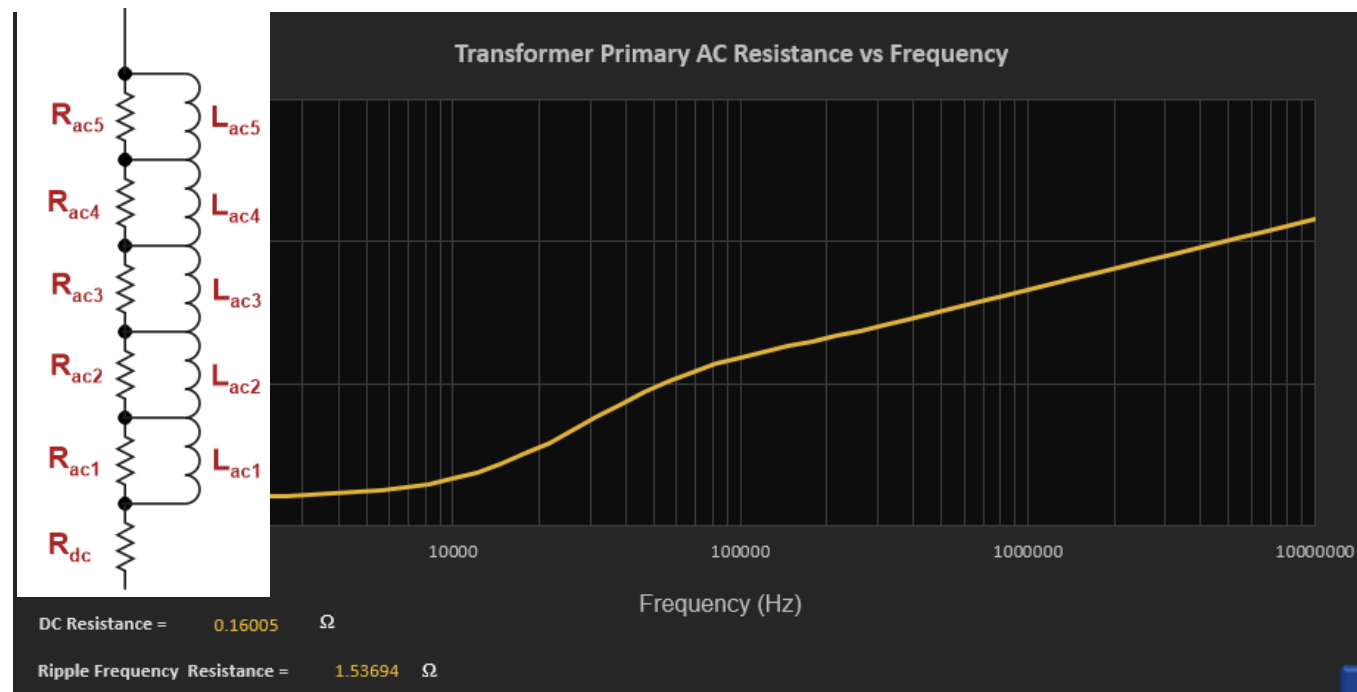
$$\Delta_i = \frac{h_{cu_i}}{\delta} \quad \text{skin depth } \delta = \sqrt{\frac{2}{\omega \mu_o \sigma \eta}}$$

$$\sigma = \text{conductivity} \quad \mu_o = 4\pi \times 10^{-7} \quad \eta = \text{porosity}$$

If you click on the blue [Proximity Loss](#) button, you can see a detailed breakdown of the resistance of the wire layer-by-layer. The complexity of Dowell's equations shows why few engineers ever attempt this analysis. Fortunately, RidleyWorks® automates the whole procedure.

Fig. 6.4: Primary loss analysis by layer together with Dowell's equations published in 1966.

## 6.2 Advanced Proximity Losses with LTspice® Circuit Modeling



Once you have defined the winding structure and are happy with the total dissipation, you can begin the process of LTspice® model generation and simulation. When you click on this button

RidleyWorks® will proceed to sweep the solutions to Dowell's equations from 1 kHz to 10 MHz and generate an equivalent circuit model to match the complex impedance.

This process must be followed for each of the magnetics windings to generate the equivalent circuit models to be used in LTspice®. You can then click on the button below to export the complete circuit model.

Fig. 6.5: Sweep of ac resistance and equivalent circuit model

### 6.3 Running LTspice® for Proximity Loss

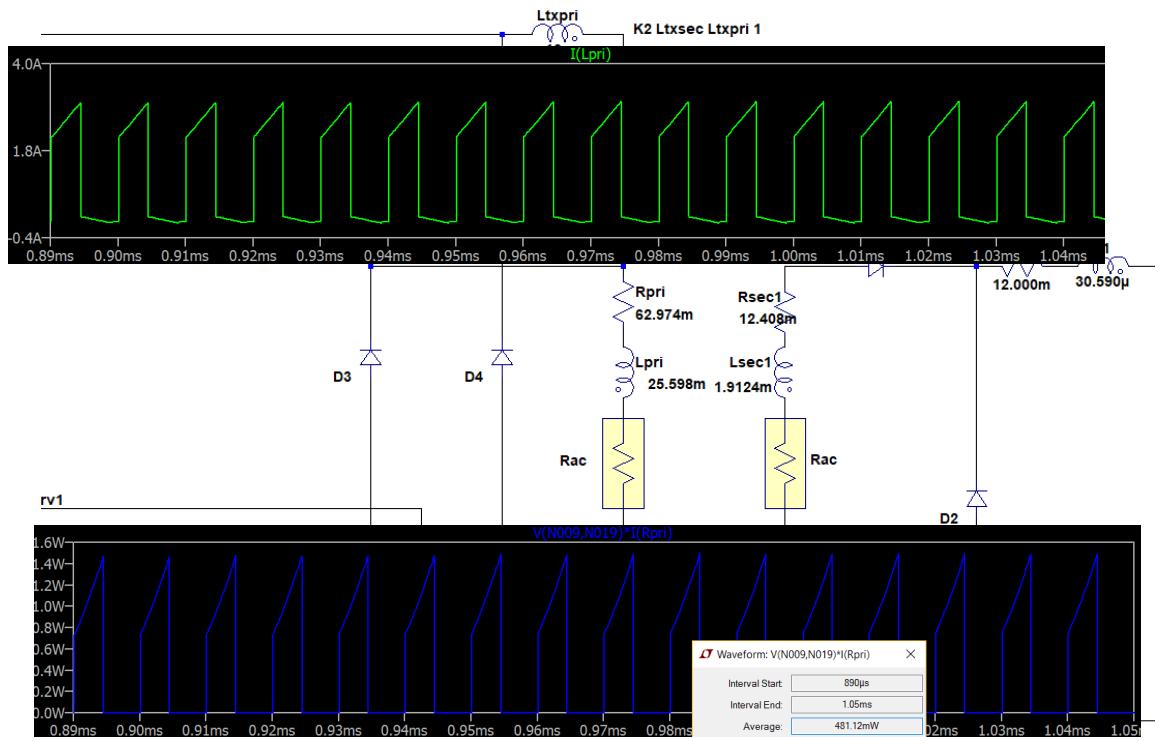
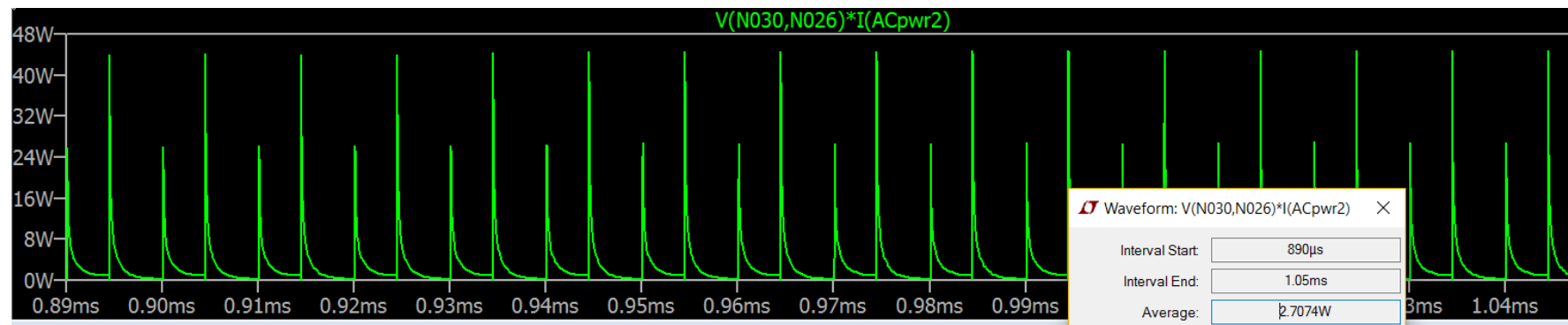


Fig. 6.6: Conventional dissipation in transformer primary (no proximity loss).

Figure 6.6 shows the two-switch forward schematic in LTspice®. The yellow resistive elements are the ac proximity models for each of the windings.

Step 1 is to simulate the converter to steady-state and plot the primary switch current  $I(L_{pri})$  as shown.

You can plot the conventional dissipation in the primary (without proximity) by holding down the **ALT** key and clicking on **Rpri**. This results in the waveform shown in Figure 6.8. If you click on the name of the waveform in blue, while holding down the **Ctrl** key, it will show the average dissipation. You can see that this number agrees with the dissipation in RidleyWorks® before proximity is added.



**Fig. 6.7: Additional proximity loss dues to ac resistance model.**

You can now plot the additional proximity loss dissipation in the primary by holding down the **ALT** key and clicking on the circuit element **Rac** in the primary. This results in the waveform shown in Figure 6.9. If you click on the name of the waveform in green, while holding down the **Ctrl** key, it will show the average dissipation. The sum of the conventional dissipation and the **Rac** dissipation is the total winding loss in the primary.

You will probably see the total dissipation predicted by LTspice® is higher than that predicted by RidleyWorks®. The LTspice® number will be more accurate since it includes the dissipation in the resistors at higher harmonic frequencies. Depending on the topology and operating point, the difference may be considerable. In the example above, the conventional losses are 0.481 W. The additional proximity losses produce an extra 2.7 W of dissipation!

Once you have mastered this technique, you will be doing proximity loss calculations that are more advanced than just about anyone in the industry. Attempting to do this work without the aid of RidleyWorks® is extremely time-consuming and difficult – you now have a powerful tool that gives you a huge design advantage.

# 6.4 Advanced Core Losses with LTspice® Circuit Modeling

Transformer Core Material Choices

OPERATING CONDITIONS

Operating Delta B	0.069	T	Core Area	0.424	cm2
Core Excitation Frequency	166	kHz	Peak Flux at Max Duty	0.284	T
Duty Cycle D1 (Flux Rise)	0.449		Number of Turns	11	
Duty Cycle D2 (Flux Fall)	0.449		Core Volume	1.65	cm3
Temperature	70	deg C			

TRANSFORMER CORE MATERIAL SELECTION

Magnetics

☒ F ☐ P ☐ R

Ferroxcube

☐ 3C90 ☐ 3F3 ☐ 3C96

TDK

☐ PC40 ☐ PC44 ☐ PC47 ☐ PC90 ☐ PC95

CHARACTERISTICS FOR SELECTED MATERIAL

Permeability	3000	Magnetizing Inductance	0.1746	mH	Core Loss	0.131	W
--------------	------	------------------------	--------	----	-----------	-------	---

LTSPICE CORE LOSS MODEL

Voltage Exponent	2.538	Resistor RC1	12	Ω
Inductor LC1	4.233	RC2	117	Ω
LC2	6.396	RC3	1.067	kΩ
LC3	9.665	RC4	9.654	kΩ
LC4	14.569	RC5	89.63	kΩ
LC5	22.543	RC6	646.402	kΩ
LC6	27.097	RC7	6.902	MΩ

Help

☐ Enable LTspice Core Model

☒ Disable LTspice Core Model

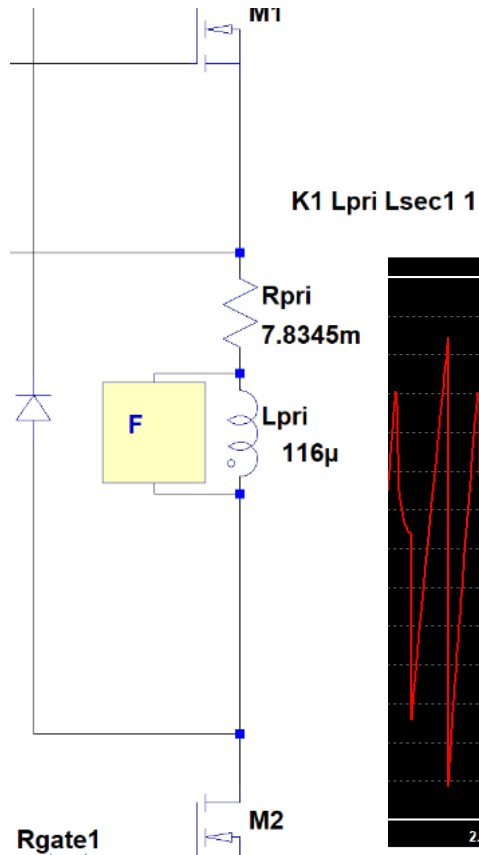
OK

With the selected core shape and material, RidleyWorks® will automatically calculate parameters for core loss simulation in LTspice® .

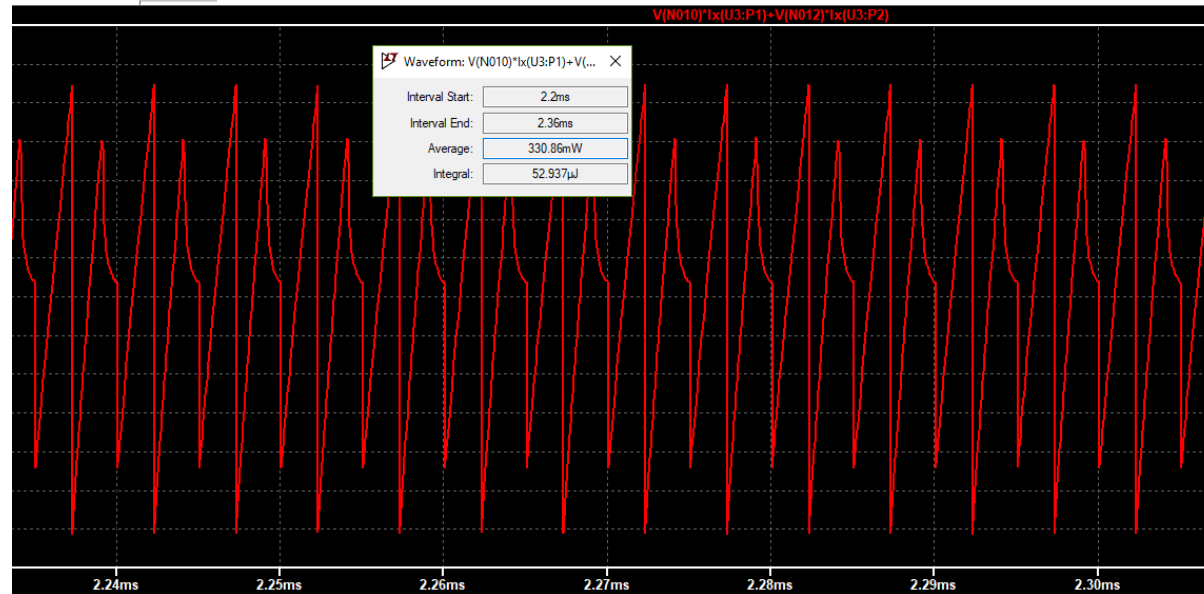
The component elements of the core loss model define how the core will behave with different drive frequencies and amplitudes. This is a unique feature of RidleyWorks®.

Fig. 6.8: Selecting a core material will produce a matching LTspice® circuit model





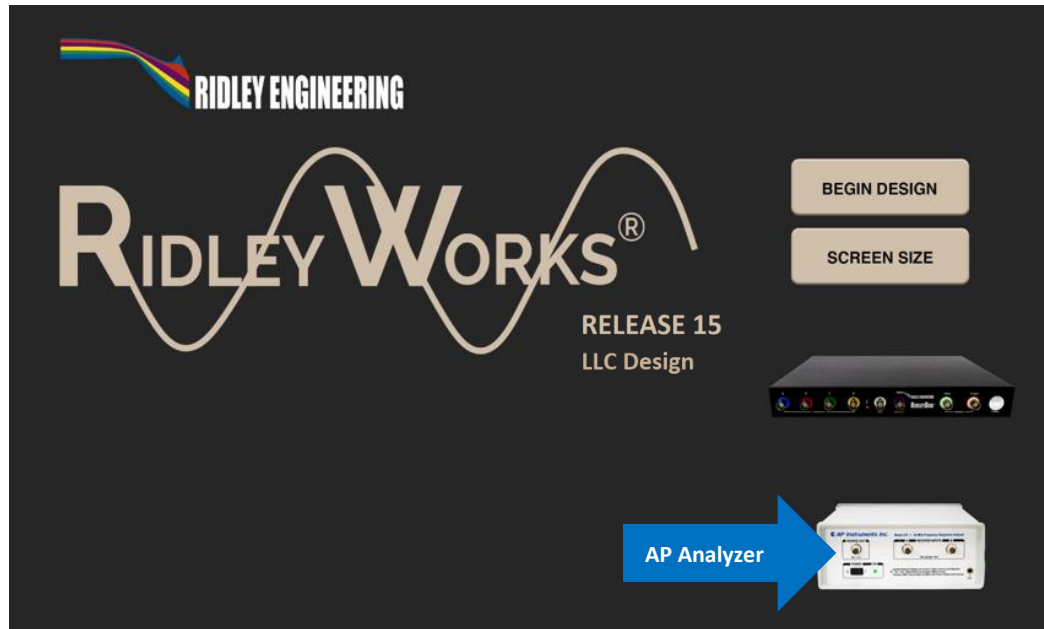
You can plot the instantaneous core losses of the block to the left of the transformer primary by holding down the **Alt** key and clicking on the block. The instantaneous core loss can be averaged by holding down the **Ctrl** key and clicking on the name of the waveform shown in red below. You will observe that for the same flux swing in a transformer, the losses increase with a shorter duty cycle, as is observed empirically. You will also see that the losses of a transformer decrease with switching frequency if the turns are left unchanged on the core.



**Fig. 6.9:** Core loss model parameters are automatically loaded into your LTspice® schematic. Instantaneous core losses are shown, together with the average value.

# 7

## Using RidleyWorks® with the AP310 Analyzer



### 7.1 Connecting RidleyWorks® to the AP

RidleyWorks® has a powerful and user-friendly interface to the AP310 frequency response analyzer. This enables you to control complex analyzer functions and settings with a single button click inside RidleyWorks®.

Open RidleyWorks® on your computer. You will then see the opening screen shown here. Click on the AP310 analyzer picture to start controlling your instrument. (You can also click on the AP analyzer on any page that you see it within RidleyWorks®.

RidleyWorks® will connect to the AP300 Measurements page shown below.

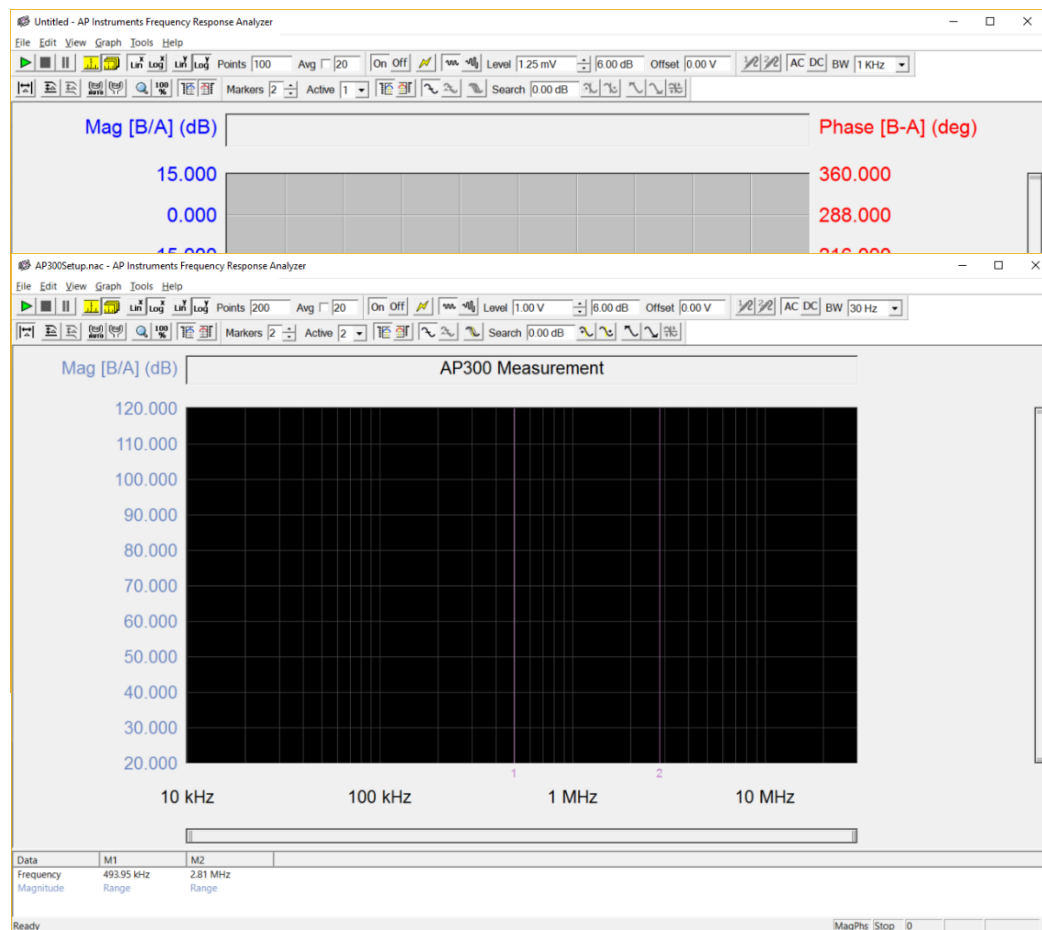
**Fig. 7.1: RidleyWorks Opening page with AP310 and RidleyBox® Icons**



This page provides a powerful interface for driving the AP300 and collecting multiple data sweeps. There are 8 preset settings on this page for magnetics measurements, control measurements, impedance measurements, and PSRR measurements.

First, turn on the power switch of the AP analyzer that you have, and click on the Connect button to automatically launch the APfra software.

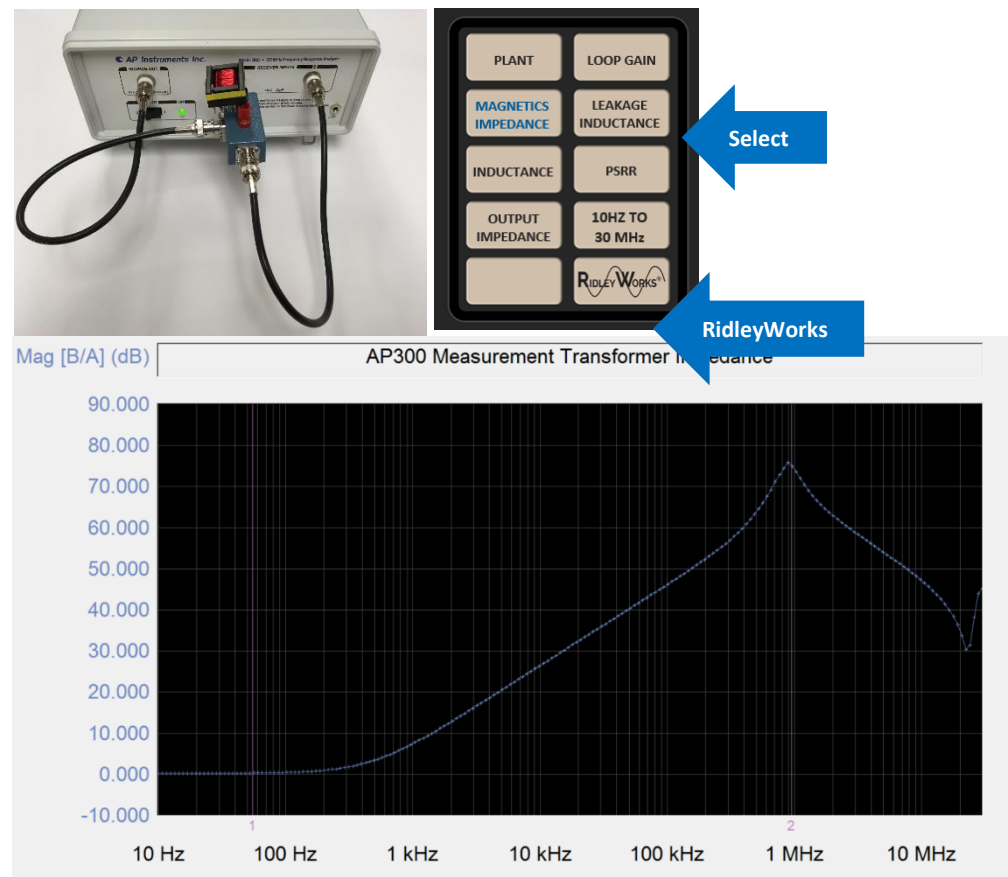
**Fig. 7.2:** AP Instruments Analyzer Interface Page



Once you see the screen to the left, Click on File>Open Setup and select AP300Setup.nac to load new default settings. Notice that the sweep is now logarithmic, and the stop frequency is 30 MHz with a drive signal of 1 V.

**Fig. 7.3: Apfra default opening screen and screen after loading AP300Setup file.**

# 7.2 Measuring Transfer Functions



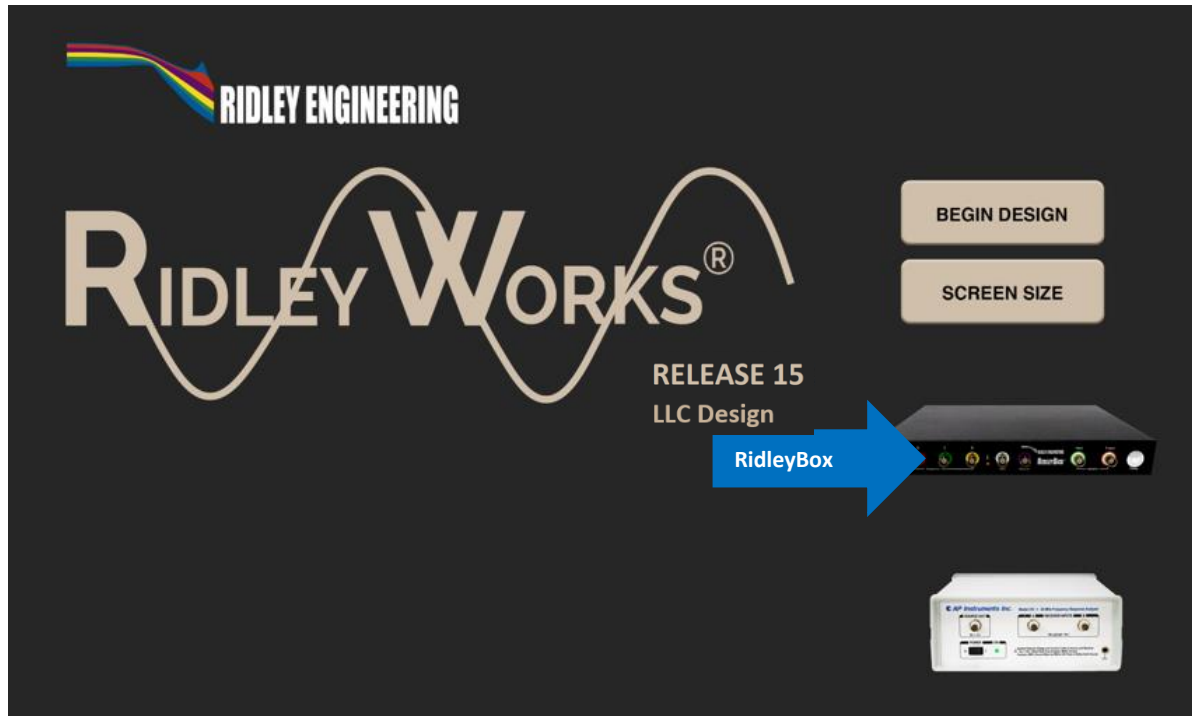
**Fig. 7.4: Magnetism impedance measurement setup and test results**

There are multiple setups for automating the settings on the AP310. This includes measuring the power stage plant, loop gains, magnetic impedances, PSRR and output impedance.

Each of these setups will adjust the source, bandwidth, and frequency range for the AP analyzer. You can still adjust these quantities individually within the APfra software as needed.

Full details of these measurement setups are provided in the AP310 analyzer user manual.

# 8 Using RidleyWorks® with the RidleyBox®



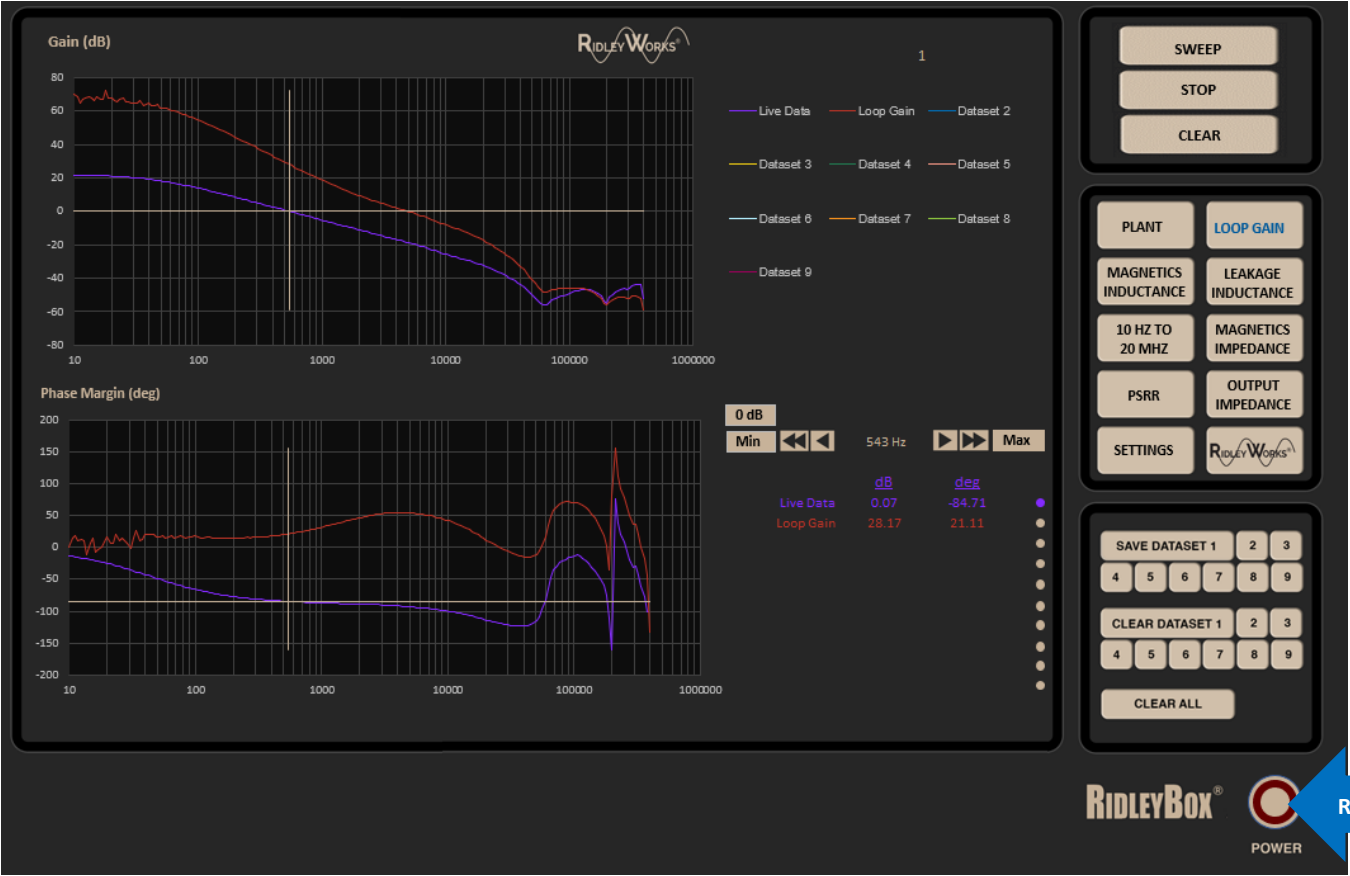
## 8.1 Connecting RidleyWorks® to the RidleyBox®

RidleyWorks® has a powerful and user-friendly interface to the RidleyBox®. This enables you to control complex analyzer functions and settings with a single button click inside RidleyWorks®.

Open RidleyWorks® on your computer. You will then see the opening screen shown here. Click on the RidleyBox® picture to start controlling your instrument. (You can also click on the RidleyBox® on any page that you see it within RidleyWorks®.)

RidleyWorks® will connect to the RidleyBox® measurements page shown below.

**Fig. 8.1: RidleyWorks® Opening page with AP310 and RidleyBox® Icons**



## 8.2 RidleyBox® Interface

This page provides a powerful interface for driving the RidleyBox® and collecting multiple data sweeps. There are 8 preset settings on this page for magnetics measurements, control measurements, impedance measurements, and PSRR measurements.

The first step is to turn on the RidleyBox® from within the software. No physical power button is needed. When you click on the button, you will soon hear a message that the connection is complete, and the power indicator will turn green.

You can then select the measurement that you want to make. Parameters will be taken from RidleyWorks® to set up each individual measurement properly. Please refer to the RidleyBox® manual for more details.

Fig. 8.2: RidleyBox® Interface Page

# 9

## Designing LLC Converters with RidleyWorks®

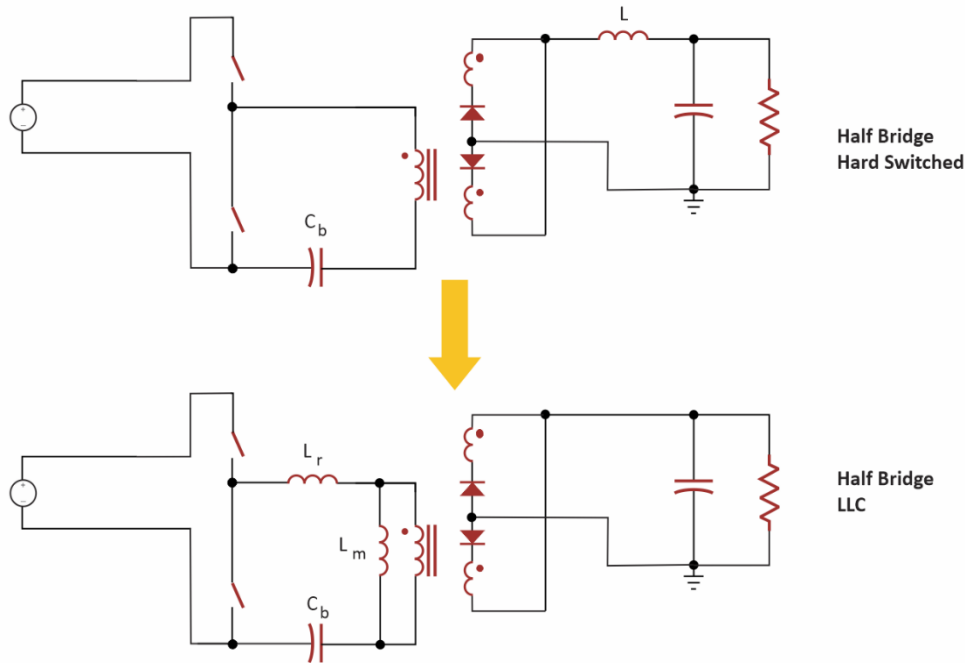


Figure 10.1: Comparison of Half-Bridge and LLC Circuits

### 9.1 Why LLC?

The half-bridge converter has always held a place in the world of practical power conversion. Now imagine a technology that:

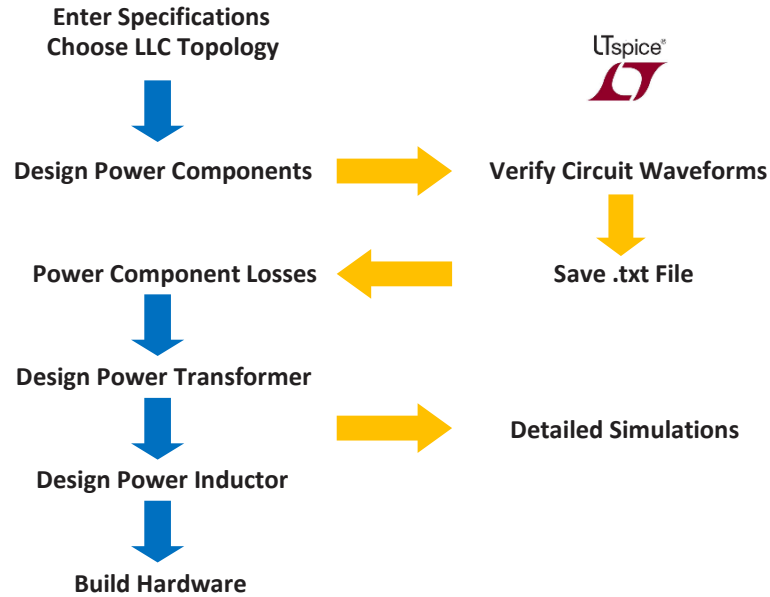
- 1) Eliminates the output inductor
- 2) Reduces the AC coupling cap by an order of magnitude
- 3) Achieves zero-voltage (lossless) switching in the primary
- 4) Achieves zero-current switching in the secondary

That is what the LLC (Inductor-Inductor-Capacitor) converter achieves. It has been around for decades, but only really became widespread after regulations mandated power factor correction at the input of the power converter.

The LLC is not a panacea for all applications. Drawbacks include:

- 1) Poorly-understood circuit design process (solved by RidleyWorks®)
- 2) Higher peak and RMS currents
- 3) Complex and widely-varying control characteristics
- 4) Light-load regulation challenges
- 5) Complex magnetics design process
- 6) Unsuitable for wide input ranges





**Figure 10.2: RidleyWorks® LLC Design Process Overview**

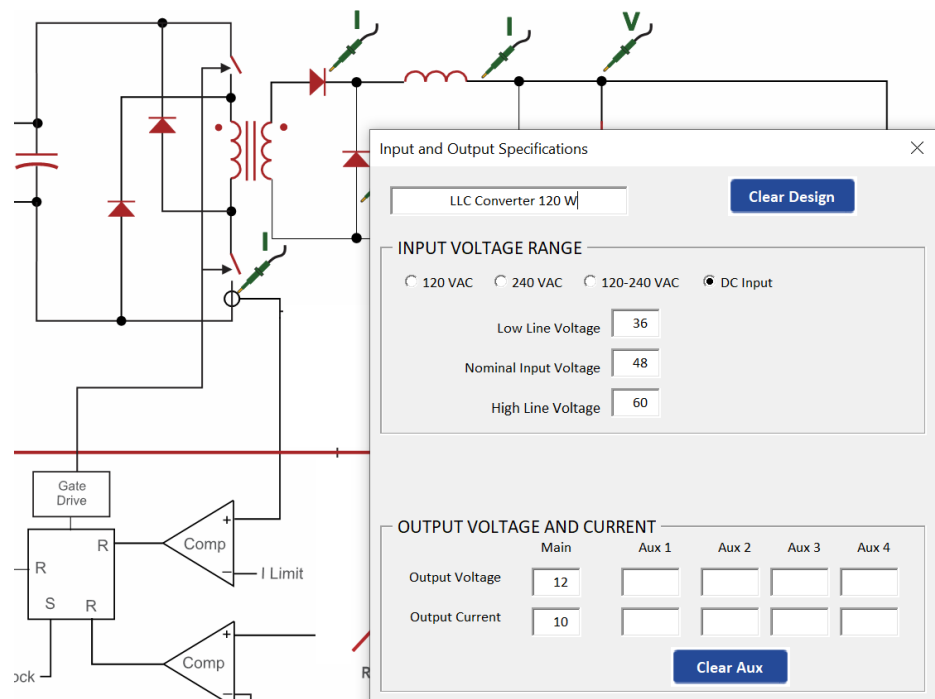
## 9.2 RidleyWorks® LLC Design Overview

RidleyWorks® has a fully automated power stage design algorithm built into the program. The design process is based upon the superb work of Sanjaya Maniktala, and the follow-on work by Nicola Rosano. Their basic philosophy is that you only need one design optimized for an LLC converter, and every other converter specification will take this optimized design and scale it according to voltage and current specifications. There is no need to reinvent the wheel every time you design an LLC converter.

The overview of the design flow is shown in Figure 9.2. Once you have designed the power components you can then run a circuit simulation in LTspice® to verify the circuit waveforms under different operating conditions. These waveforms can then be brought back into RidleyWorks® for more detailed component design work.

Less experienced designers may want to let their preferred magnetics vendor design the custom parts needed for your converter. You just need to provide the the specified component values and simulated waveforms.

Experienced designers usually like to have full control over the magnetics design and parts specification. RidleyWorks® will help you with this process. It can work at a basic level, or deep into issues of magnetics proximity core losses if you prefer.



**Fig. 9.3: RidleyWorks® Specification Form**

## 9.3 Entering Specifications

The first step of the design is to enter the input line range and the output load specifications. A narrow range of input voltage will lead to a more optimal design for the LLC converter. The best application of LLC is when you have a pre-regulated rail, as would be provided by a power factor correction circuit on the input to the converter.

When you click OK, you will see the form on the next page allowing you to specify the switching frequency for your design.

Frequency Selection

INPUT VOLTAGE

Maximum Voltage	60	VDC
Nominal Voltage	48	VDC
Minimum Voltage	36	VDC

POWER SUPPLY OUTPUT

Main Output Voltage	12	V
Main Output Current	10	A
Total Output Power	120	W

SWITCHING FREQUENCY

Swi

100

kHz

144

kHz Default

Clear Design

OK

Specifications

Frequency

Topology

LLC Power Design

Waveforms

Control Design

Efficiency

Inductor Design

Transformer Design

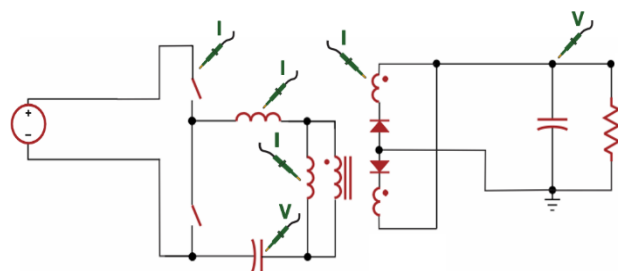
Design Summary

## 9.4 LLC Frequency Selection

This form should appear automatically after entering the power specifications. You can also pull up the form by clicking on the Frequency button.

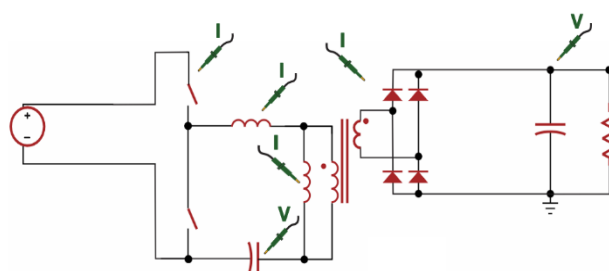
For the LLC converter, the switching frequency entered on this form is the target frequency for full-load, low-line operation. At higher line input, or lighter load, the regulated frequency of operation will be higher than this target frequency. This is a natural characteristic of the LLC converter.

**Fig. 9.4: Frequency Selection for the LLC Converter**



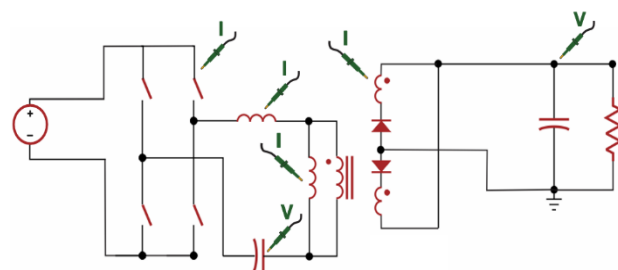
Half-bridge Inverter

Half-wave Rectifier



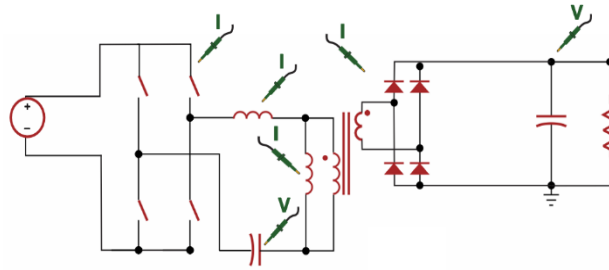
Half-bridge Inverter

Full-wave Rectifier



Full-bridge Inverter

Half-wave Rectifier

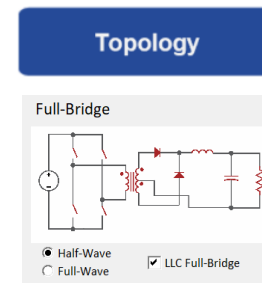


Full-bridge Inverter

Full-wave Rectifier

## 9.5 LLC Topology Choices

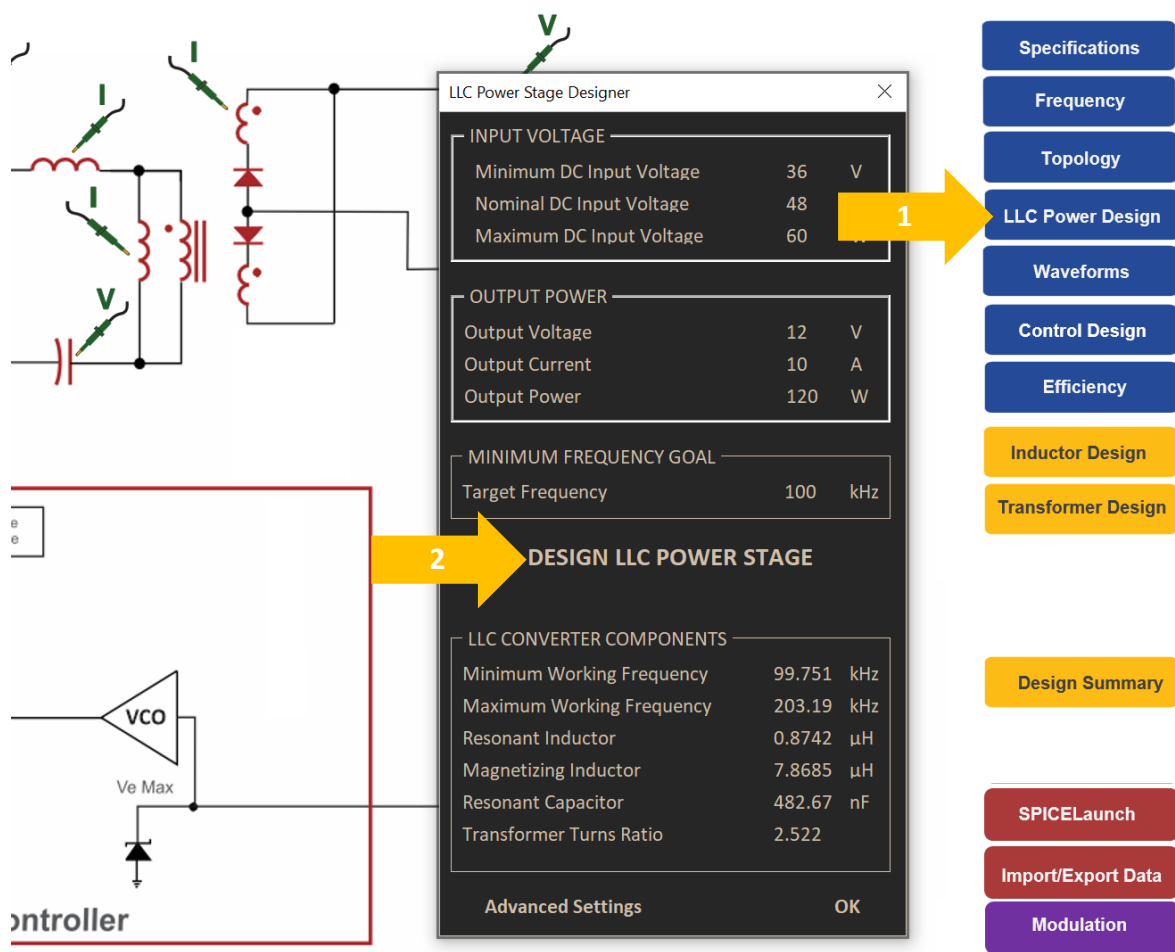
The first step is to click on the Topology button and select either the full-bridge or half-bridge converter.



There is a check box to select either full or half-wave rectification, and to enable the LLC selection. This is shown above.

(Please note that the small thumbnail schematic will always show just a half-wave rectifier, even when the full-wave is selected.)

**Fig. 9.5: LLC Circuit Variation Available in RidleyWorks®**



## 9.6 LLC Circuit Optimization

Once you have selected the desired LLC topology, you will see the appropriate schematic appear, along with the design button for choosing the LLC circuit component values. Click on this button to pull up the LLC Power Stage Designer form.

The design process is initiated by clicking on the text highlighted with the second arrow.

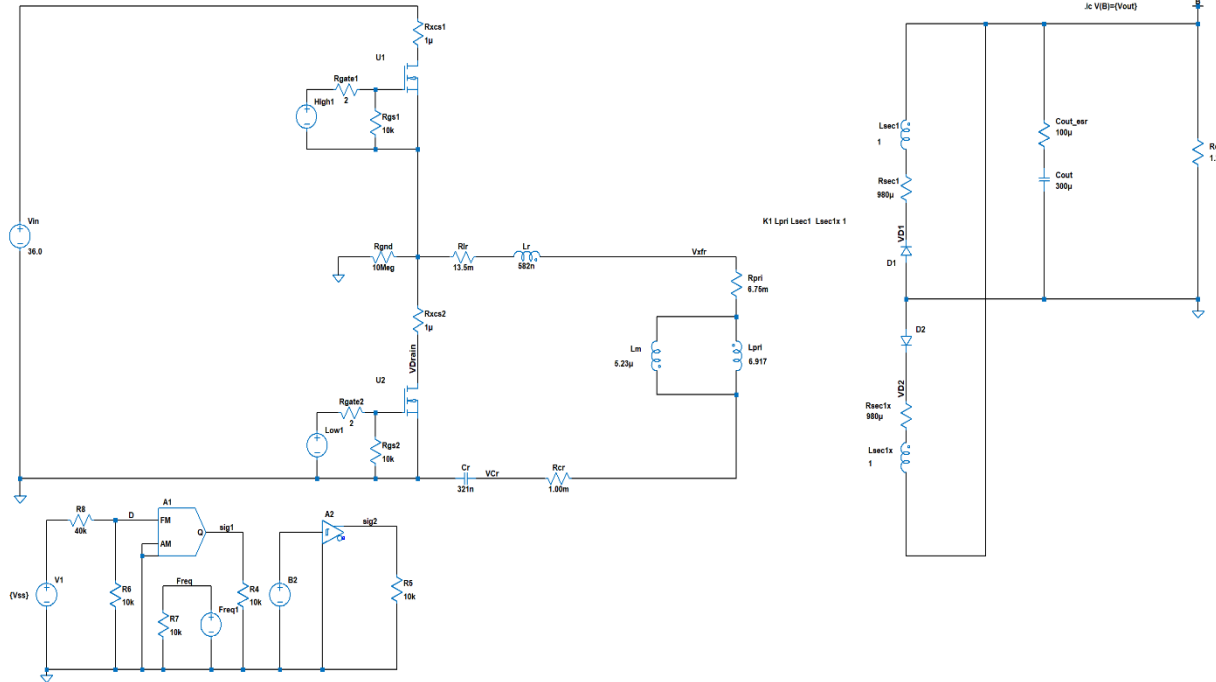
Once you do this, you will see RidleyWorks® search for the optimum design point, changing the component values and frequencies at the same time.

You will then see the message that the design is complete, and simulation files with all the component values will automatically be created, ready to run in LTspice®. A total of 5 simulation files will be created and placed in a folder on called C:\RidleyWorks\RidleyWorks LTspice.

Fig. 9.6: LLC Power Stage Design Form

RidleyWorks LTspice Schematic  
 LLC CONVERTER 120 W  
 Vin = 36.0, Vout1 = 12.0V @ 10.0A, llc steady state, voltage-mode

150 kHz



## 9.7 LLC Circuit Simulation

Before you can view waveforms and proceed further with the design in RidleyWorks®, a simulation of the LLC converter in steady state should be run. The file created for this simulation is:

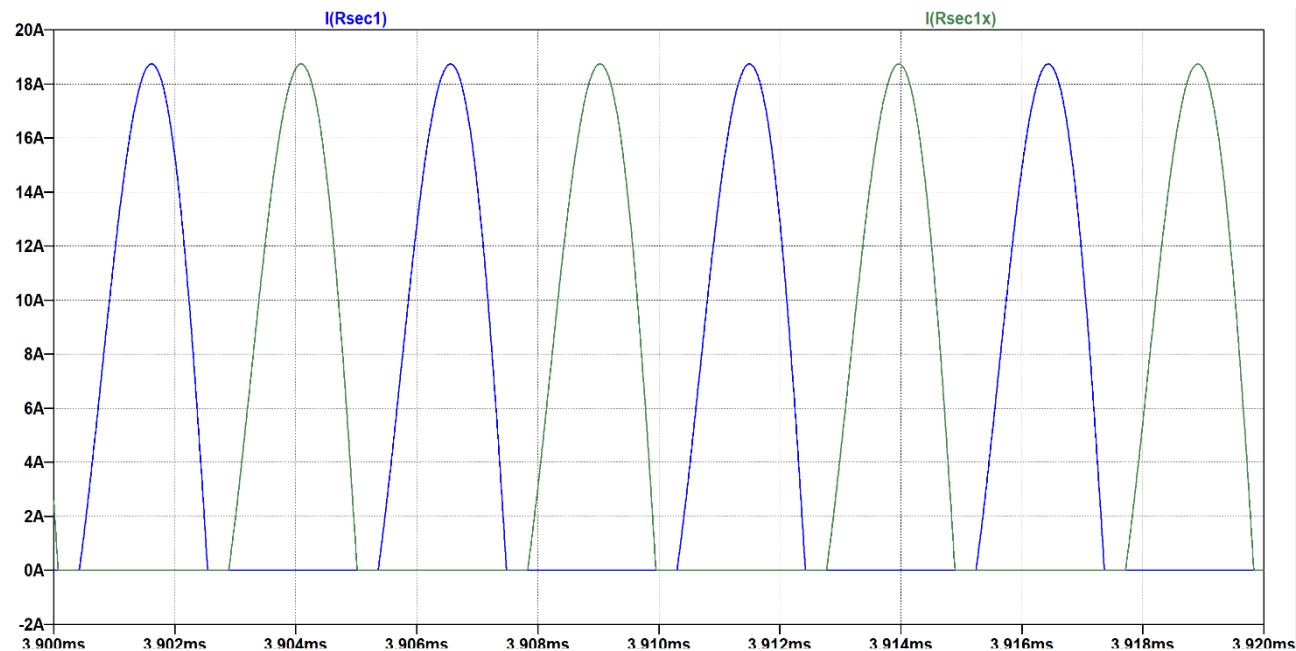
1 LLC STEADY STATE.asc

The default operating condition for the LLC design is with the minimum input voltage and full load. At this operating point, the converter should be close to the specified output voltage at the desired switching frequency selected in RidleyWorks.

If you see that the output voltage is too high, you can increase the value of **Vss**. That will increase the switching frequency and decrease the output voltage. If the output voltage is too high, decrease the value of **Vss** to lower the switching frequency.

You can look at the waveforms inside LTspice® to verify proper operation. Important waveforms are listed in the .save lines of LTspice®, and these can be imported into RidleyWorks®.

Fig. 9.7: LLC Power Stage Steady-State Simulation Circuit



**Fig. 9.8: LLC Secondary Currents at High-Line Condition. Zero -Current Switching is Maintained**

## 9.8 High Line Operation

It is a good idea to run the steady-state simulation again with high-line input to the converter. When you do this, you will see the output voltage will be higher than the desired value. This can be reduced by raising the switching frequency of the converter by increasing **Vss**.

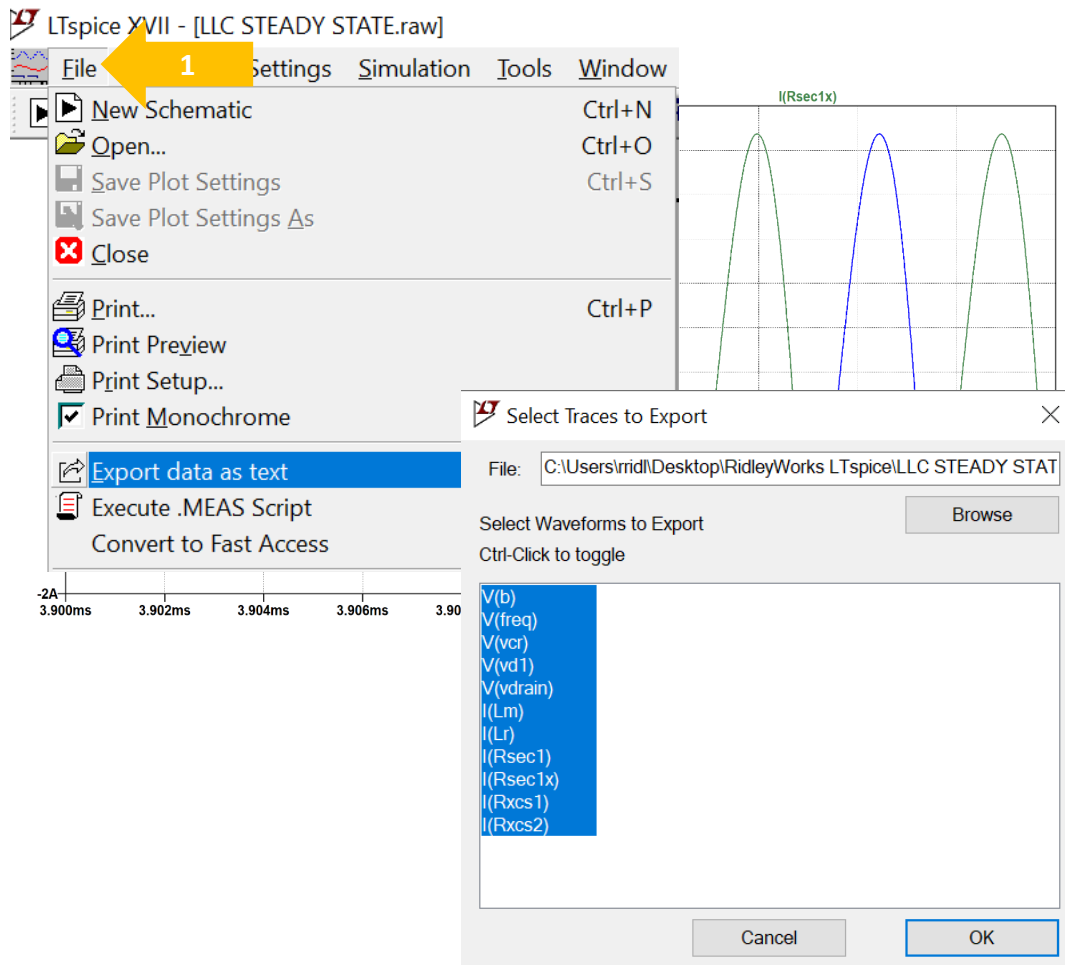
### CONTROLLER PARAMETERS

```
.param Vss=0.761
.param FreqMax=243K
.param FreqMin=73.89K
```

The parameter **Vss** controls the switching frequency. A value of 5 V will set it to the maximum value, and a value of 0 V will set it to the minimum value.

When the output is regulated to the desired value, you should see that both secondary currents complete a full half-cycle and zero-current switching is maintained. This indicates proper design of the power stage components.

When you are finished verifying the waveforms in LTspice®, you can proceed with the component design in RidleyWorks®.



## 9.9 Exporting Waveforms from LTspice®

In the waveform window of LTspice®, click on **File** and **Export data as text**. Highlight all of the listed waveforms and click OK to save the simulation results as a .txt file. RidleyWorks® has an automated reader to import all of these values.

Note: this text file will be quite large due to the long simulation time of the steady-state circuit. You can reduce the final simulation time in the following statement if you see that steady state is reached well before the end of the simulation.

```
.param sim_time=8.000m
```

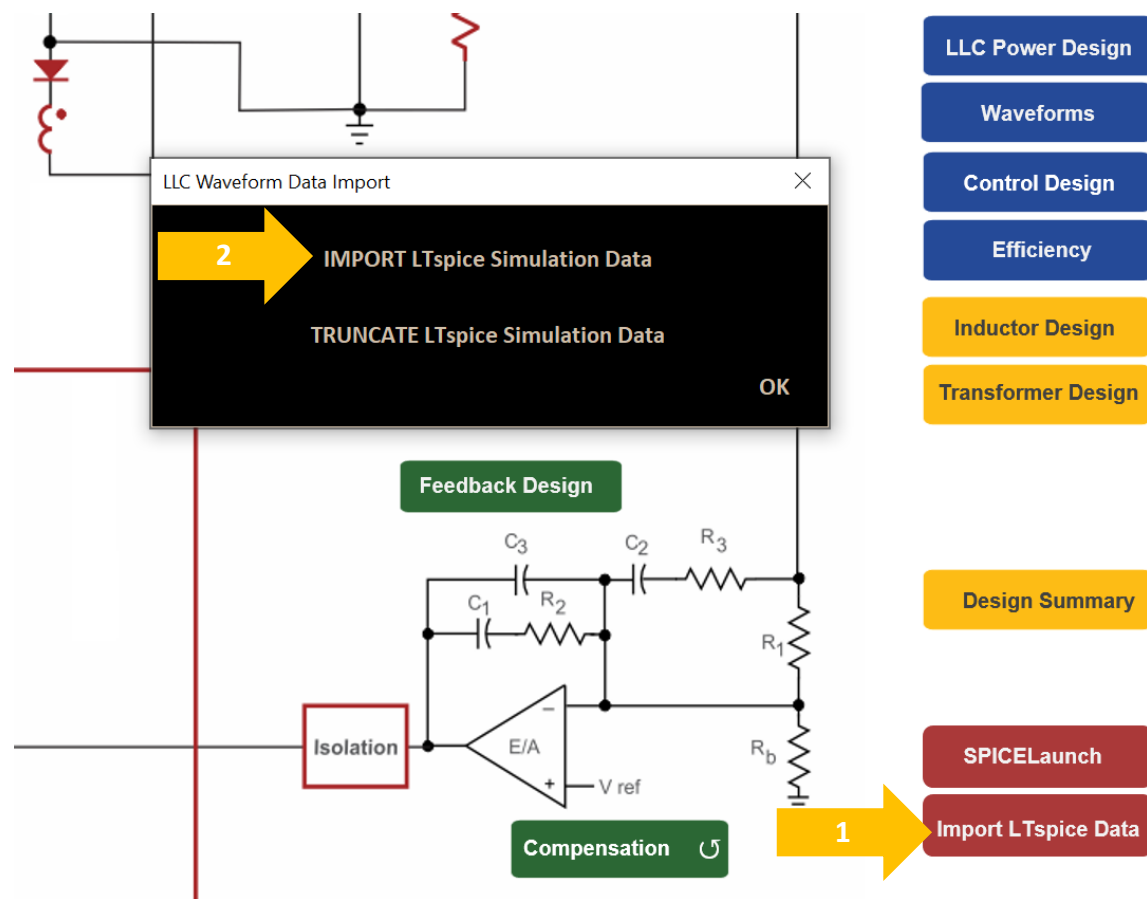
By default, the waveform data will be stored in the file named:

**1 LLC STEADY STATE.TXT**

Do not change this file name as RidleyWorks® will search for this file name when trying to import the data.

Fig. 9.9: Exporting Data from LTspice® Simulation





## 9.10 Importing LLC Waveforms

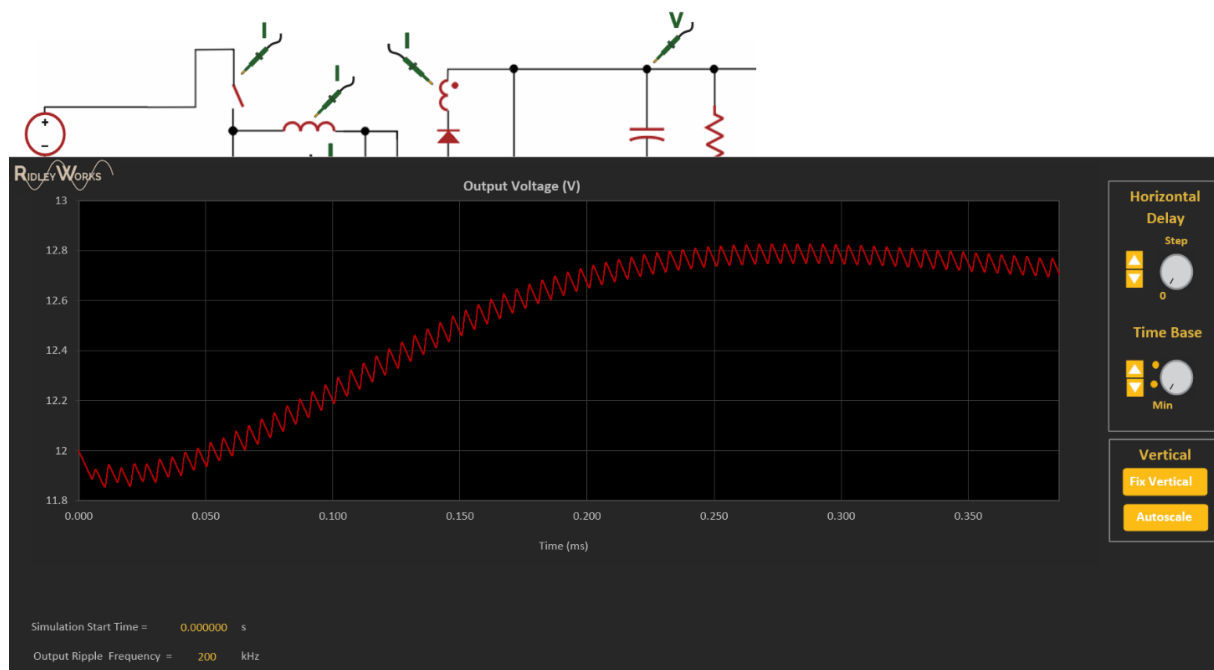
The waveform data that has been saved to the file 1 LLC STEADY STATE.TXT can now be imported into RidleyWorks®.

Click on the button to import the LTspice® data. You will then see a form with two options. The first option will import the large data file saved by LTspice®. You can bring in up to 1 million lines of data, but these will be truncated to the first 500 k lines. It will typically take a minute or so to import the large data set.

You also have the option to truncate the imported data to just the first 50 k lines in order to minimize the file size of RidleyWorks®. This step will always be automated when you exit RidleyWorks® and select the option to save the file.

The main purpose of importing the waveforms into RidleyWorks® is to be able to design the components of the power stage. In general, you will find it more useful to work on the waveform viewing inside LTspice®.

Fig. 9.10: Importing Data from LTspice® Simulation into RidleyWorks®



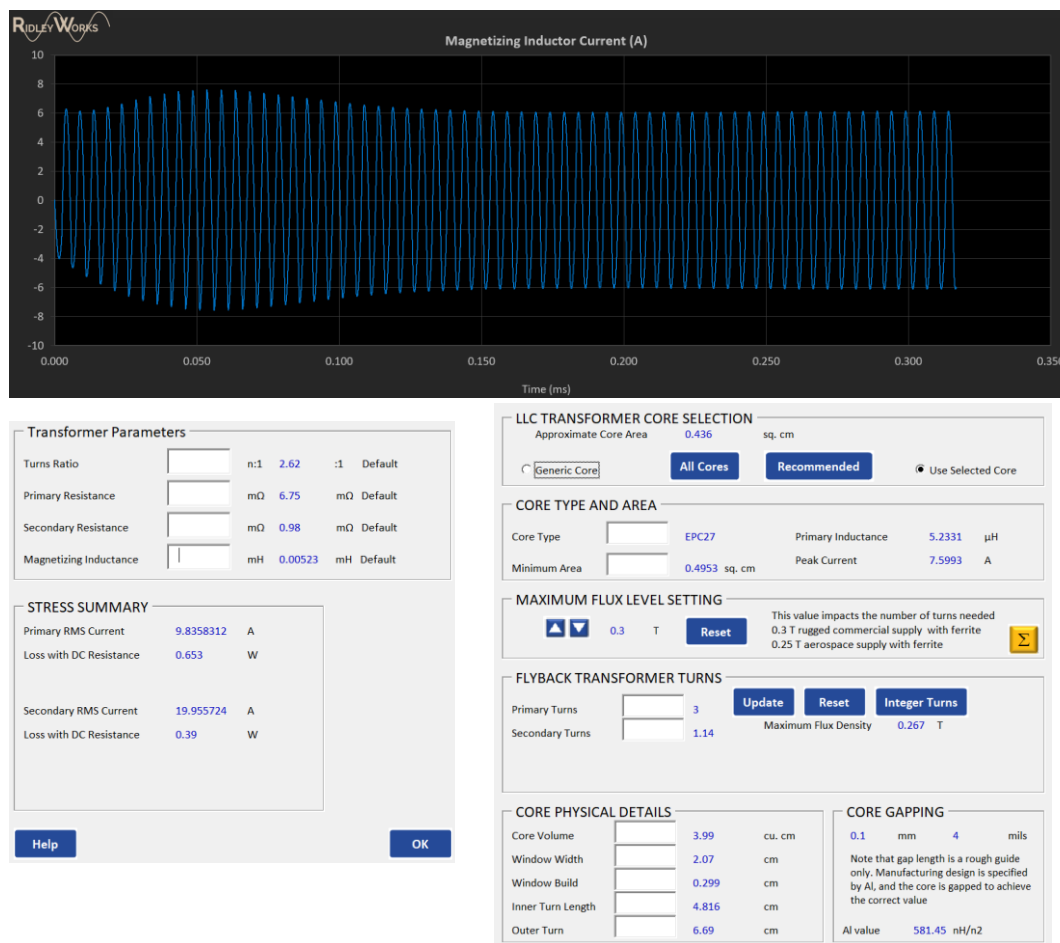
## 9.11 Viewing LLC Waveforms

Click on the of the green probes on the circuit schematic to see current or voltage waveforms. You can zoom in on the waveforms inside RidleyWorks® in the normal way using the timebase controls on the screen.

Once you have imported the waveforms, you can begin the transformer design. For the experienced designer, this process is straightforward, especially when assisted with the magnetics design capability of RidleyWorks®.

If you don't have magnetics design experience, you can learn more by coming to our [workshops](#) or you can work with your magnetics vendor to help you. Please note that LLC design is not familiar to many magnetics companies, and it may be better to learn the process yourself.

**Fig. 9.11: Viewing Waveforms in RidleyWorks®**



## 9.12 LLC Transformer Design

RidleyWorks® treats the LLC transformer like an inductor design for the selection of the turns needed on the transformer core. It is important to simulate the converter with LTspice to see the needed energy storage in the transformer. The peak current and the magnetizing inductance value are used to make sure that the maximum flux level of the core is not exceeded.

Notice that there is an additional constraint on the turns count on the primary since there cannot be less than one turn on the secondary. The number of primary turns must be at least equal to the turns ratio of the transformer.

As frequencies are raised above 150 kHz, you may find that you increase the minimum turns count to reduce the observed core loss in the transformer. This is usually an iterative design process to trade off core and copper losses. For example, with three turns on the primary, using a PC95 material, there are 3 W of core loss.

PC95	New	New	New
SELECTION MATERIAL			
Magnetizing Inductance	0.0108 mH	Core Loss	2.827 W

With 6 turns on the primary, the losses in the core are reduced

Core Loss	0.563 W
-----------	---------

This is a better design choice.

Fig. 9.12: LLC Transformer Parameters

Transformer Primary Winding Design

**PRIMARY WINDING ALLOCATION**

Window Allocation: 0.5  
Number of Turns: 6

**AVAILABLE WINDOW**

Margin Required: 0 mm each end, 2.07 cm Available Window  
Insulation Required: 0 mm total, 0.1495 cm

**WINDING STRUCTURE**

☒ Litz Wire ☐ Magnet Wire ☐ Triple Insulated Wire ☐ Foil

Number of Layers: 1 **Optimize Layers** Number of Parallel Wires: 1  
Maximum Conductor Size That Will Fit Exactly is: 18 awg  
Your Choice of Conductor Size: 18 awg **Update** Optimize Gauge  
Leakage Inductance: 0.0664  $\mu$ H

**WINDING LOSS DETAILS**

RMS Current	9.285	A	Winding Loss (Including Proximity)	0.73	W
DC Current	0	A	Winding Loss (without Proximity Loss)	0.58	W
AC Current	9.285	A	Winding Surface Area	12	sq.cm
DC Resistance	0.00673	$\Omega$			
AC Resistance	0.008477	$\Omega$	<b>Proximity Loss</b>	<b>Sweep AC Resistance</b>	$\Sigma$

**Fig. 9.13a: LLC Transformer Primary**

**WINDING STRUCTURE**

☐ Magnet Wire ☐ Triple Insulated Wire ☒ Foil

Number of Layers: 2 Minimize Loss Number of Parallel Wires: 1  
Maximum Conductor Size That Will Fit Exactly is: 0.902 mm = 35.53 mills  
Your Choice of Conductor Size: 0.2301 mm = 9.05 mills **Update** **Minimize Loss**

**WINDING LOSS DETAILS**

RMS Current	13.526	A	Winding Loss (Including Proximity)	0.203	W
DC Current	5.373	A	Winding Loss (without Proximity Loss)	0.168	W
AC Current	12.413	A	Winding Surface Area	12	sq.cm
DC Resistance	0.0004621	$\Omega$			
AC Resistance	0.000618	$\Omega$	<b>Proximity Loss</b>	<b>Sweep AC Resistance</b>	$\Sigma$

**Fig. 9.13b: LLC Transformer Secondary**

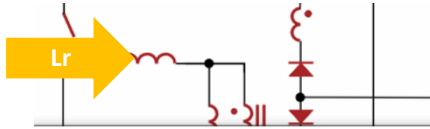
## 9.13 LLC Transformer Winding Design

Once the turns count is established for a specific core and turns ratio, we turn our attention to the detailed winding layout in the transformer. For this design, six turns are used on the primary of the transformer. If you try to use solid wire, the proximity losses are significant due to the high ac content of the primary transformer waveform. Switching to Litz wire reduces the losses substantially.

As with standard transformers, RidleyWorks estimates the loss of the winding including proximity effects. For experienced designers, you can go deeper and ask RidleyWorks to sweep the AC resistance versus frequency in order to build an LTspice model. This step is extremely powerful when estimating the losses in the transformer at different operating points.

You could also choose to use a foil for the primary. The disadvantage of this approach is that it will increase the capacitance of the transformer and affect the operation of the LLC converter. As with any design, it is crucial to measure the actual performance on prototype hardware (and later on production hardware) to ensure that the parameters of the transformer are close to the specified values. Engineers with either the AP300 or the RidleyBox will be aware of how to easily do this.

After you have designed both the primary and secondary windings, you can use the leakage inductance estimation of the primary winding to begin the design of the resonant inductor of the LLC.



Inductor Design

INDUCTOR VALUES

Inductance:   $\mu\text{H}$   $\mu\text{H}$  Default

Winding Resistance:   $\text{m}\Omega$   $\text{m}\Omega$  Default

$R_1$   $L$

☒ Subtract Measured Leakage   $\mu\text{H}$

Inductor Design

LLC RESONANT INDUCTOR CORE SELECTION

Approximate Core Area   $\text{sq. cm}$

☐ Generic Core   ☒ Use Selected Core

CORE TYPE AND AREA

Core Type  Inductance   $\mu\text{H}$

Minimum Area   $\text{sq. cm}$  Peak Current   $\text{A}$

MAXIMUM FLUX LEVEL SETTING

$\text{T}$

This value will determine the number of turns needed

INDUCTOR TURNS REQUIRED

Turns

Maximum Flux Density at Peak Current Limit:   $\text{T}$

CORE PHYSICAL DETAILS

Core Volume   $\text{cu. cm}$

Window Width   $\text{cm}$

Window Build   $\text{cm}$

Inner Turn Length   $\text{cm}$

Outer Turn Length   $\text{cm}$

CORE GAPPING

$\text{mm}$    $\text{mills}$

Note that gap length is a rough guide only. Manufacturing design is specified by  $Al$ , and the core is gapped to achieve the correct value

$Al$  value   $\text{nH/n}^2$

Fig. 9.14: Resonant Inductor Core Selection

## 9.14 LLC Inductor Core and Turns Design

The first step in designing the LLC series inductor is to determine the value of the discrete inductor to be added. The transformer leakage inductance will provide a portion of the resonant inductor value. You can account for this contribution by clicking on the schematic of the converter. Enter the value of the leakage inductance (make sure you measure this with your AP300 or RidleyBox to confirm the value at the switching frequency) on the primary side of the transformer. You will see the required amount of extra inductance needed changing as this leakage number is changed.

Some LLC transformers are deliberately designed with very high leakage so that an additional inductance is not needed. In our experience with industry LLC converters, the high-efficiency products add a discrete external inductor in most cases.

As with the transformer design, you may see that the minimum number of turns suggested (2 in this case) results in high core loss. Increasing the number of turns to 3 will produce a more optimal design.

Inductor Winding Design

**INDUCTOR DESIGN**

Inductor Value: 0.5156 uH      Number of Turns: 3

**AVAILABLE WINDOW**

Margin Required: 0 mm each end      0.67 cm winding length

Insulation Required: 0 mm total      0.24 cm build

**WINDING STRUCTURE**

☐ Use Magnet Wire    ☒ Use Foil    ☐ Helical Foil

Number of Layers: 3      Optimize Layers      Number of Parallel Wires: 1

Maximum Conductor Size That Will Fit Exactly is: 0.78 mm = 30.7 mills

Your Choice of Conductor Size: 0.3975 mm = 15.64 mills           

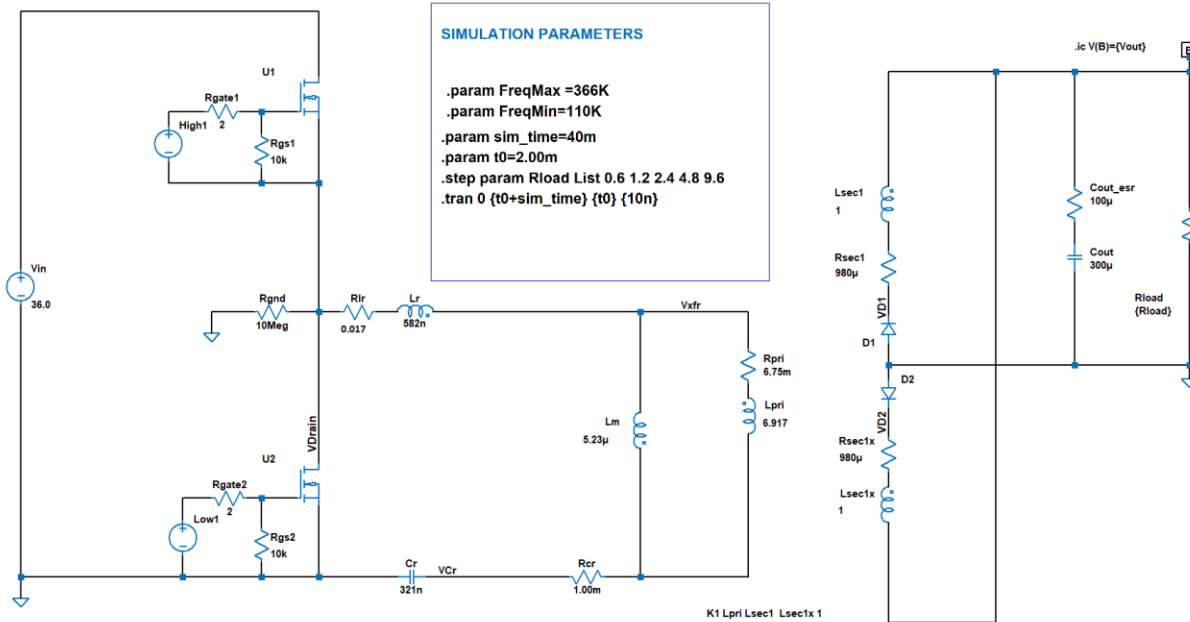
**WINDING LOSS DETAILS**

RMS Current	9.285786 A	Winding Loss (Including Proximity)	0.06 W
DC Current	0 A	Winding Loss (without Proximity Loss)	0.045 W
AC Current	18.136708 A	Winding Surface Area	2.07 sq.cm
DC Resistance	5.22E-04 Ω	<input type="button" value="Proximity Loss"/> <input type="button" value="Sweep AC Resistance"/> <input type="button" value="Σ"/>	
AC Resistance	7.05E-04 Ω		

## 9.15 LLC Inductor Winding Design

For this design, we only need three turns on the inductor. The losses are minimized by using a foil structure with a 15 mil thickness foil. Notice that it is possible to use a thicker foil, but the high AC currents would result in higher proximity losses in the winding.

**Fig. 9.15: Resonant Inductor Winding**



## 9.16 LLC Gain Curves Schematic

RidleyWorks® generates a second LTspice simulation file for the LLC converter named:

2 LLC CURVE TRACER.asc

When you run this file in LTspice, the frequency will be increased from the minimum value to the maximum value over the duration of the simulation.

```

.param FreqMax =366K
.param FreqMin=110K

```

The minimum and maximum values from RidleyWorks should cover the entire range of operation of the converter from high line down to low line.

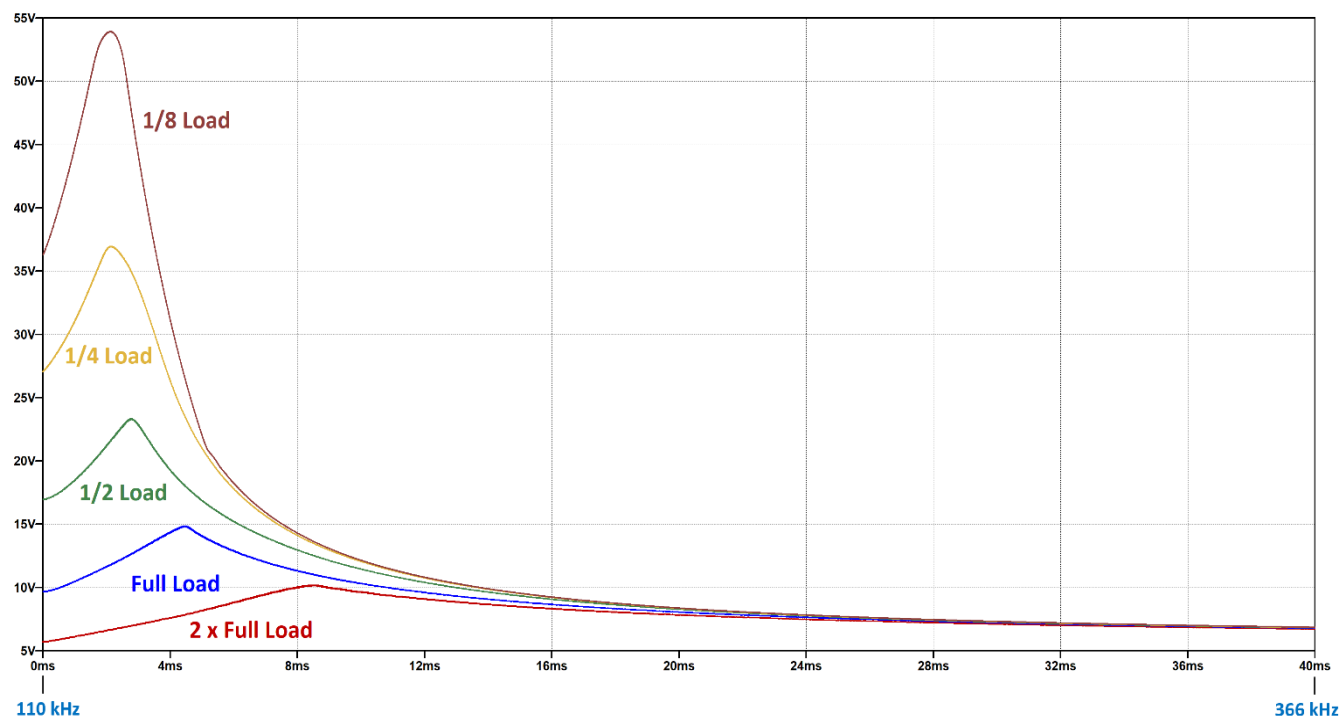
Please note that the accuracy of the curves will be impacted by the total simulation time. If the time is too short, there will be portions of the curve that have sharp changes in them. If you increase the time, and the curves do not change, then you can be confident the curves are accurate. In this example, the default simulation time of about 6 ms is increased to 40 ms for optimal results.

```

.param sim_time=40m

```

Fig. 9.16: Schematic for Generating LLC Gain Curves



**Fig. 9.17: LLC Gain Curves for a 12-V 10-A Output with a 36-V Input**

## 9.17 LLC Gain Curves

The LLC gain curves show the output voltage for your specific converter design at different load levels. For the lowest curve in red, the load is set at twice the specified value. In this case, output voltage cannot reach the specified value of 12 V and the converter cannot regulate.

At full load, the curve shown in blue shows that the output voltage reaches the desired regulation value just above the resonant peak.

The curves generated are the actual curves including all the losses of your particular circuit. This is much more useful than working with theoretical curves created by first harmonic analysis methods.

You can spend a lot of time working with these curves to understand how your converter is working. For example, you will want to redraw them for high-line input to determine the actual limit to the switching frequency needed.

























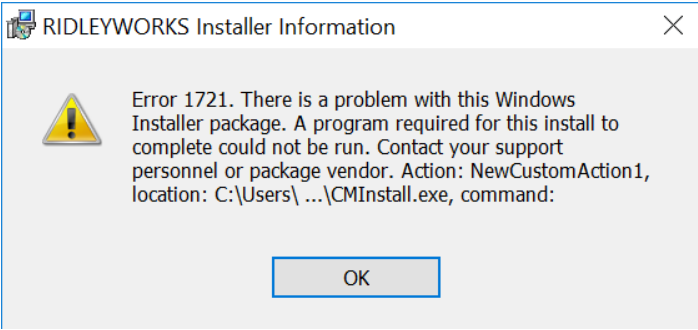




# Appendix A    Error Messages

Error Message	Page Number
Installer Error 1721	101
Cannot find RIDLEYWORKS.DLL	101
Protection Error 717 Incorrect Passkey Entered	102
Protection Error 758 License Expired	102
Protection Error 923	102
Trouble Closing RidleyWorks®	103
Run-Time Error 1004	104
Data Entry Format Error – Run-Time Error 13	105
Microsoft Excel Stopped Working (EMET problem)	105
Object Doesn’t Support Property or Method	106
LTSpice Cannot Find Sub or Sym Files	107

## Installer Error 1721

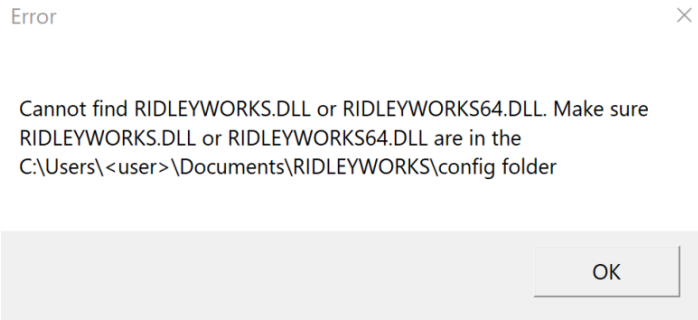


This message indicates that the CMInstall program must be run with Administrator privileges on your computer, and you will have to manually finish the installation. In this case, right click CMInstall.exe, and select Run As Administrator. Answer OK to any questions, and at the end of the process, the needed files will be created. More details of files are on Page 8 of this manual.

You can find this file in the folder C:\ProgramData\AFY\RIDLEYWORKS\cm

Please note that ProgramData is a Hidden file, and it will only show up if you click on View, Show and check Hidden Items.

## Cannot find RIDLEYWORKS.DLL

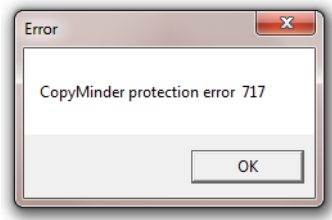


Some operating environments look in the SysWOW64 or the System32 directory or AFY directory for the proper .dll permission files. If you get this message, please do the following:

Copy all 6 of the .dll files from C:\ProgramData\AFY\RIDLEYWORKS\cm  
to: C:\Windows\SysWOW64

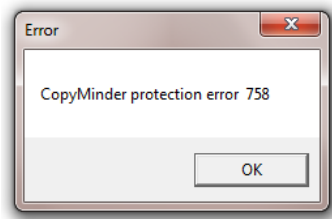
Copy all 6 of the .dll files from C:\ProgramData\AFY\RIDLEYWORKS\cm  
to: C:\Windows\System32

**Protection Error 717 Incorrect Passkey Entered**



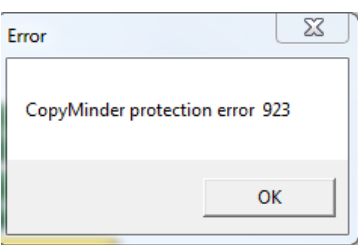
You will see this message if you enter an invalid passkey for the program. Check for the proper number, and contact Ridley Engineering if you cannot find it.

**Protection Error 758 License Expired**



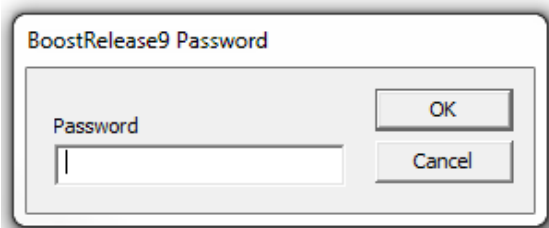
You will receive this error when your license has reached its expiration date. Contact us to verify the date or to renew your registration.

**Protection Error 923 Installation Incomplete**

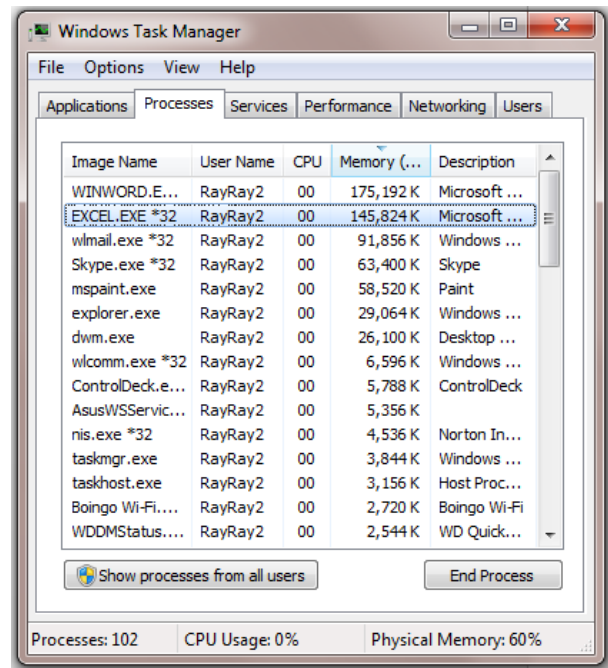


You may receive this error if the software installation did not finish generating all of the necessary files. Please go to Section 1.4, item number 2 of this manual to learn how to complete the installation.

## Trouble Closing RidleyWorks®

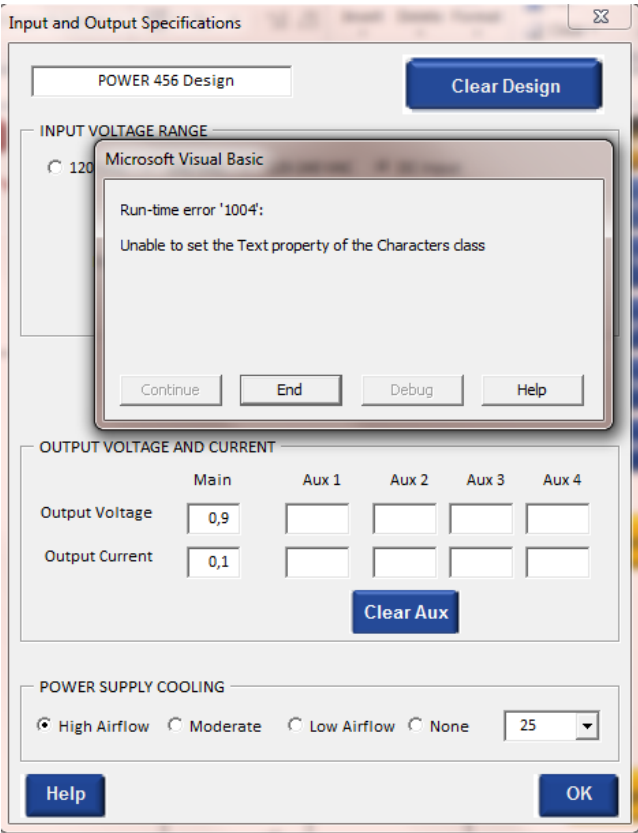


Very rarely, you may see this screen pop up when you try to close the program. Clicking OK or Cancel will not get rid of it. If this happens, type `ctl-alt-delete` and select Start Task Manager. Select Processes, highlight EXCEL.EXE\*32, and click on End Process. This should close the program.



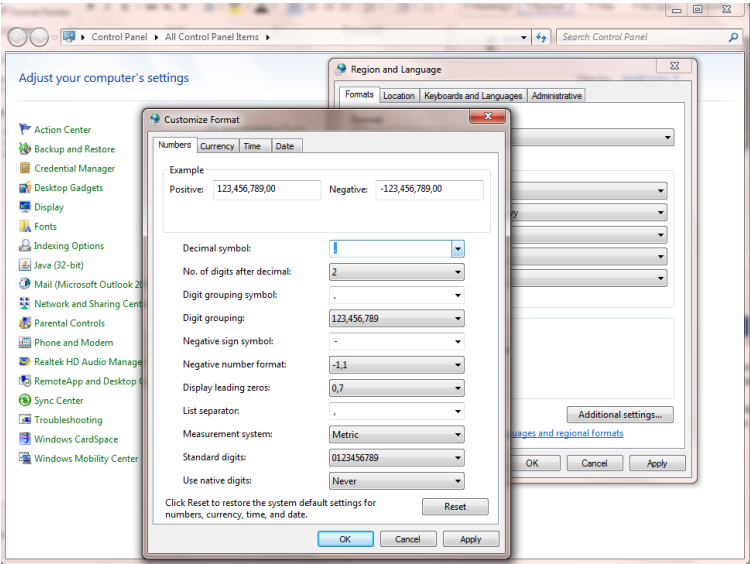


Run-Time Error 1004



This error will occur if you have an old version of Excel prior to 2013. Please upgrade to the latest version.

In some cases, the error will occur if you do not have the decimal separator character set to a period “.”. You can change the setting of the decimal separator in your Windows system as described below.



In the windows search bar pull up the Control Panel, Region and Language. Under the Format tab, click on Additional Settings, and you will be able to change the character to a period as shown.

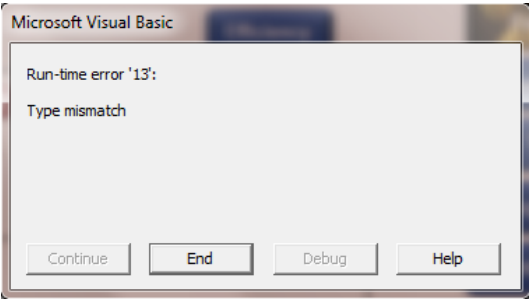
Also make sure that Excel is using the system separators. Go to File Options Advanced in Excel and check the box shown here:

☒ Use system separators

Decimal separator:

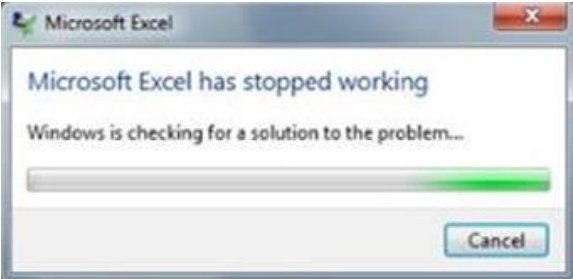
Thousands separator:

### Data Entry Format Error – Run-Time Error 13



Most mistakes in data entry will be flagged by RidleyWorks® and will give you an error message. However, if you see the message here, you may have the decimal separator set incorrectly in your system. This is changed as described in the above section. Or, you may have used some other invalid numerical character. If you see this happen, please try to describe the sequence of events to us or send us the file. We are always working to improve and update the program.

### Microsoft Excel Stopped Working (EMET problem)



If you get the message that Excel has stopped working while verifying the license, it may be because your IT department has installed Microsoft EMET (Enhanced Mitigation Experience Toolkit). This will block access to the required website to verify the license code.

It is necessary to disable the following EMET options for RidleyWorks® to register itself properly without triggering EMET to shut down the program.

- Export Address Table Filtering
- Simulate the execution flow

There is no security risk in implementing these changes.

## Object Doesn't Support Property or Method

You may see messages such as this when you click on one of the buttons. Microsoft did some security updates recently, and it causes problems in excel. You can find it discussed at:

<https://social.technet.microsoft.com/Forums/exchange/en-US/3f29c84b-97ad-46f7-9bc0-8648d69336ce/kb2553154-breaks-our-excel-macros?forum=officeitproprevious>

fix is to simply delete the file MSForms.exe from any Temp subfolder in the user's profile. For instance:

```
C:\Users\[user.name]\AppData\Local\Temp\Excel8.0\MSForms.exe
```

```
C:\Users\[user.name]\AppData\Local\Temp\VBE\MSForms.exe
```

```
C:\Users\[user.name]\AppData\Local\Temp\Word8.0\MSForms.exe
```

You can search for any files ending in .exe and delete them while excel is closed. Then start excel again and it should work properly. Microsoft are aware of this issue and are trying to fix it, apparently.

More information is at <http://stackoverflow.com/questions/27411399/microsoft-excel-activex-controls-disabled>

## LTSpice Cannot Find Sub or Sym Files

On installation, the special symbols and subcircuits used by RidleyWorks to simulate fast and efficiency are placed in the

**C:\RidleyWorks\LTspice** folder by default.

The folders are also placed in the directory:

**C:\Users\<username>\AppData\Local\LTspice\lib** folders, consistent with the current installation of LTspice (April 2024).





# Appendix B License Agreement

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